Features

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330MHz, Gain of +1/Gain of +2 **Closed-Loop Buffers**

General Description

The MAX4178/MAX4278 are ±5V, wide-bandwidth, fastsettling, closed-loop buffers featuring high slew rate, high precision, high output current, low noise, and low differential gain and phase errors. The MAX4178, with a -3dB bandwidth of 330MHz, is preset for unity voltage gain (0dB). The MAX4278 is preset for a voltage gain of +2 (6dB) and has a 310MHz -3dB bandwidth.

The MAX4178/MAX4278 feature the high slew rate and low power that are characteristic of current-mode feedback amplifiers. However, unlike conventional currentmode feedback amplifiers, these devices have a unique input stage that combines the benefits of current-feedback topology with those of the traditional voltage-feedback topology. This combination results in low input offset voltage and bias current, low noise, and high gain precision and power-supply rejection.

The MAX4178/MAX4278 are ideally suited for driving 50Ω or 75Ω loads. They are the perfect choice for highspeed cable-driving applications, such as video routing. The MAX4178/MAX4278 are available in DIP, SO, space-saving µMAX, and SOT23 packages.

Applications

Broadcast and High-Definition TV Systems Video Switching and Routing High-Speed Cable Drivers Communications Medical Imaging Precision High-Speed DAC/ADC Buffers

♦ High Speed

330MHz -3dB Bandwidth (MAX4178) 310MHz -3dB Bandwidth (MAX4278) 250MHz Full-Power Bandwidth (Vout = 2Vp-p) 150MHz 0.1dB Flatness Bandwidth 1300V/µs Slew Rate (MAX4178) 1600V/µs Slew Rate (MAX4278)

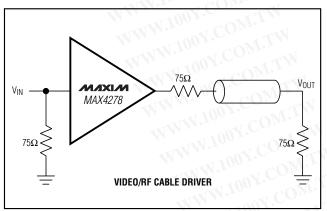
- ♦ Low Differential Phase/Gain Error: 0.01°/0.04%
- ♦ 8mA Supply Current
- ♦ 1µA Input Bias Current
- ♦ 0.5mV Input Offset Voltage
- ♦ 5nV/√Hz Input-Referred Voltage Noise
- ♦ 2pA/√Hz Input-Referred Current Noise
- ♦ 1.0% Max Gain Error with 100Ω Load
- ♦ Short-Circuit Protected
- ♦ 8000V ESD Protection
- **Available in Space-Saving SOT23 Package**

Ordering Information

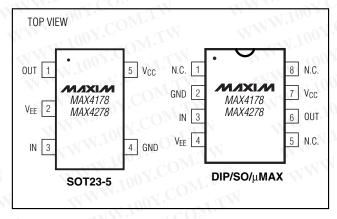
PART	TEMP. RANGE	PIN- PACKAGE	SOT TOP MARK	
MAX4178EPA	-40°C to +85°C	8 Plastic DIP	Y.Co.	
MAX4178ESA	-40°C to +85°C	8 SO	< COM	
MAX4178EUA	-40°C to +85°C	8 µMAX	MON.	
MAX4178EUK-T	-40°C to +85°C	5 SOT23-5	ABYX	
MAX4178MJA	-55°C to +125°C	8 CERDIP	CO.	

Ordering Information continued at end of data sheet.

Typical Operating Circuit



Pin Configurations



NIXIN

Maxim Integrated Products 1

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ABSOLUTE MAXIMUM RATINGS

Supply Voltage (VCC to VEE)		12V
Input Voltage		
Output Short-Circuit Duration	(to GND)	Continuous
Continuous Power Dissipation	$I(T_A = +70^{\circ}C)$	
SOT23 (derate 7.10mW/°C	above +70°C)	571mW
Plastic DIP (derate 9.09mV	V/°C above +70°C)	727mW
SO (derate 5.88mW/°C abo	ove +70°C)	471mW
μMAX (derate 4.10mW/°C	above +70°C)	330mW
CERDIP (derate 8.00mW/°	C above +70°C)	640mW

Operating Temperature Ranges (No	
MAX4178E_A/MAX4278E_A	40°C to +85°C
MAX4178EUK/MAX4278EUK	40°C to +85°C
MAX4178MJA/MAX4278MJA	55°C to +125°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Note 1: Specifications for the MAX4_78EUK (SOT23 packages) are 100% tested at T_A = +25°C, and guaranteed by design over temperature.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +5V, V_{EE} = -5V, V_{OUT} = 0, R_L = \infty, T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25$ °C.)

PARAMETER	SYMBOL	WW.IO	CONDITIONS	N v	MIN	TYP	MAX	UNITS
Innut Valtage Denge	Max	MAX4178		±2.5	±3.0	COM	- V	
Input Voltage Range	VIN	MAX4278		±1.25	±1.5	Mo	TV	
COM.	LTW	T 0500	MAX4_78ESA/EPA/EUA/MJA		MWW	0.5	2.0	TW
land Offer Weller		$T_A = +25^{\circ}C$	MAX4_78EUK			0.5	3.0	
Input Offset Voltage	Vos	T _A =	MAX4_78ESA/EPA/EUA/MJA		111	W.10	3.0	mV
MAN. TO OA. CO		T _{MIN} to T _{MAX}	MAX4_78EUK		MA	-11	5.0	
Input Offset Voltage Drift	TCVos	-1111	M. In Co)NI.	N.	2	ov.C	μV/°C
L IB: 0 30 100 1	OMITY	T _A = +25°C	JW.100	$O_{M',T}$		1N.	3	μА
Input Bias Current	IB	T _A = T _{MIN} to T	MAX	WI.M	V	1	5	
Input Resistance	RIN	N W	MANN ON COM		-	NA	1007	ΜΩ
Power-Supply Rejection Ratio	PSRR	$V_S = \pm 4.5 \text{V to } \pm 5.5 \text{V}$		70	90	N.100	dB	
MAN	Av	MAX4178 (Note 2) $\frac{R_L = 100\Omega}{R_L = 50\Omega}$		+0.990	MA	+1.000	V/V	
Valta va Oalia				+0.985	XXI"	+1.000		
Voltage Gain				+1.98	7	+2.01		
MMM				+1.97	V	+2.01		
Gain Linearity	A _{V(LIN)}	$V_{OUT} = \pm 1 \text{mV}$	to ±2V	COM.	TW	0.01	MAN.	%
Output Resistance	Rout	f = DC		1.1	0.1	TIVY	Ω	
Minimum Output Current	lout	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$		70	100	-T	mA	
Short-Circuit Output Current	Isc	Short to GND		TV	150	MW	mA	
Outrot Waltana Outra	W.100	$R_L = 100\Omega$		±2.5	±3.0	VIV	111	
Output Voltage Swing VouT		$R_L = 50\Omega$			±2.0	±2.5		V
W	t Isy	T _A = +25°C		717	8	10	-311	
Quiescent Supply Current		$T_A = T_{MIN}$ to T_{MAX} $MAX4_78E_{_}$ $MAX4_78MJA$		COM	TW	12	mA	
				COM		14		

Note 2: Voltage Gain = $(V_{OUT} - V_{OS}) / V_{IN}$ measured at $V_{IN} = \pm 2.5 V$.

Note 3: Voltage Gain = $(V_{OUT} - V_{OS}) / V_{IN}$ measured at $V_{IN} = \pm 1.25 V$.

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PARAMETER	SYMBOL	CONDITIONS		MIN TYP MAX	UNITS	
Small-Signal, -3dB Bandwidth	BW	V _{OUT} ≤ 0.1Vp-p	MAX4178	330	MHz	
Small-Signal, -Sub Bandwidth	DVV		MAX4278	310		
Small-Signal, ±0.1dB Bandwidth	DM/(a / ID)	Vour < 0.1\/p.p.	MAX4178	150	MHz	
Siriali-Sigrial, ±0. 105 Baridwidti	BW _(0.1dB)	V _{OUT} ≤ 0.1Vp-p	MAX4278	150	- IVITIZ	
Full-Power Bandwidth	FPBW	V _{OUT} = 2Vp-p	MAX4178	250	MHz	
Tull-Lower Bandwidth	N I I BV		MAX4278	250		
Slew Rate	SR	VOLUE - 12\/0 p	MAX4178	1300	1//110	
Siew hate	Sh	$V_{OUT} = \pm 2V_{p-p}$	MAX4278	1600	V/µs	
Settling Time	ts	V _{OUT} = 2V step	to 0.1%	10	ns	
Setting Time			to 0.01%	12		
Rise/Fall Times	t _R , t _F	V _{OUT} = 2V step		2	ns	
Input Capacitance	CIN	M. To. COMP.		MM. TOOM	pF	
Input Voltage Noise Density	en	f = 10MHz		5 COM.	nV/√Hz	
Input Current Noise Density	in	f = 10MHz		2	pA/√Hz	
Differential Gain	DO 1	f = 3.58MHz	MAX4178	0.04	- %	
(Note 4)	DG		MAX4278	0.04		
Differential Phase	DD	f 0.50MH-	MAX4178	0.01	degree	
(Note 4)	DP	f = 3.58MHz	MAX4278	0.01		
Total Harmonic Distortion	TUD	THD f _C = 10MHz, V _{OUT} = 2Vp-p	MAX4178	-58	dB	
Total Harmonic Distortion	UHI		MAX4278	-59		
On the Design of	OFFD	f = 5MHz, V _{OUT} = 2Vp-p	MAX4178	-81	dBC	
Spurious-Free Dynamic Range	SFDR		MAX4278	-74		
Third Order Intercent	IP3	f _C = 10MHz, Vout = 2Vp-p	MAX4178	36	dBm	
Third-Order Intercept			MAX4278	31		

Note 4: Tested with a 3.58MHz video test signal with an amplitude of 40IRE superimposed on a linear ramp (0 to 100IRE). An IRE is a unit of video signal amplitude developed by the Institute of Radio Engineers; 140IRE = 1V in color systems.

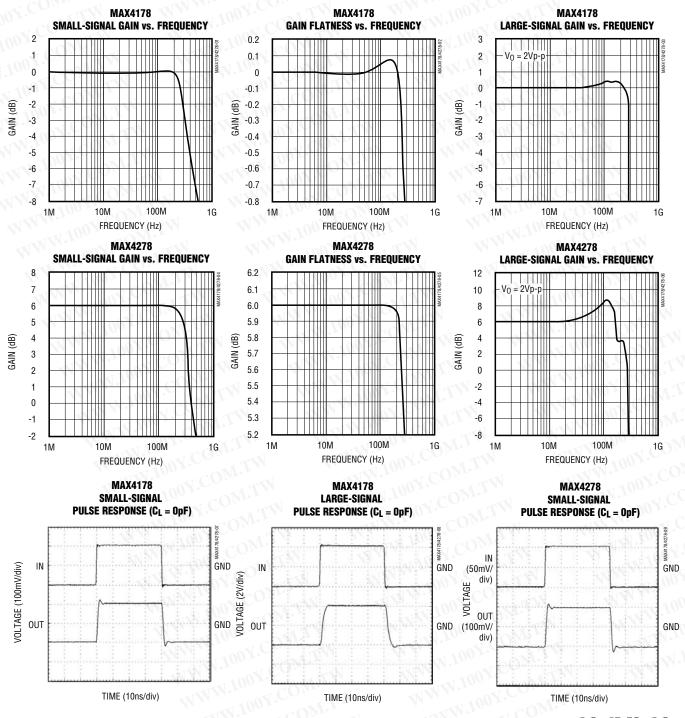
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330MHz, Gain of +1/Gain of +2 Closed-Loop Buffers

Typical Operating Characteristics

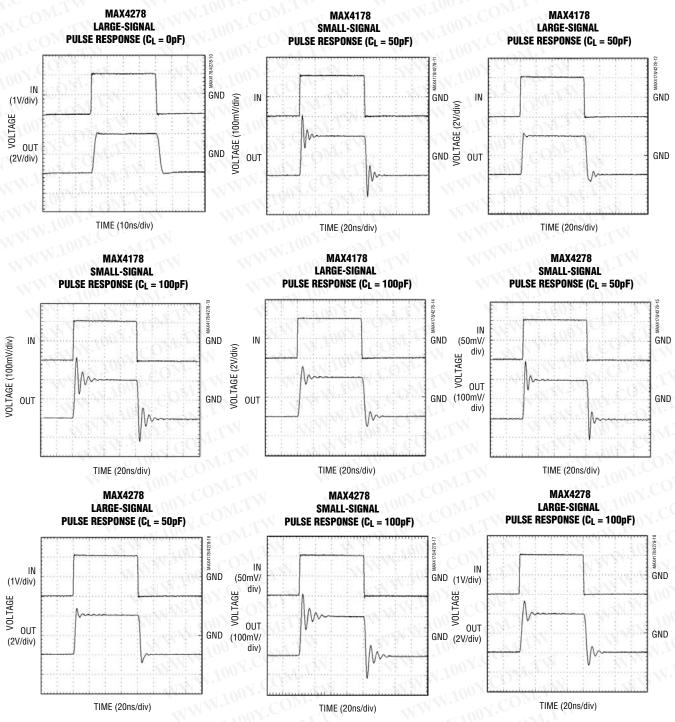
 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 100\Omega, C_L = 0pF, T_A = +25^{\circ}C, unless otherwise noted.)$



330MHz, Gain of +1/Gain of +2 Closed-Loop Buffers

Typical Operating Characteristics (continued)

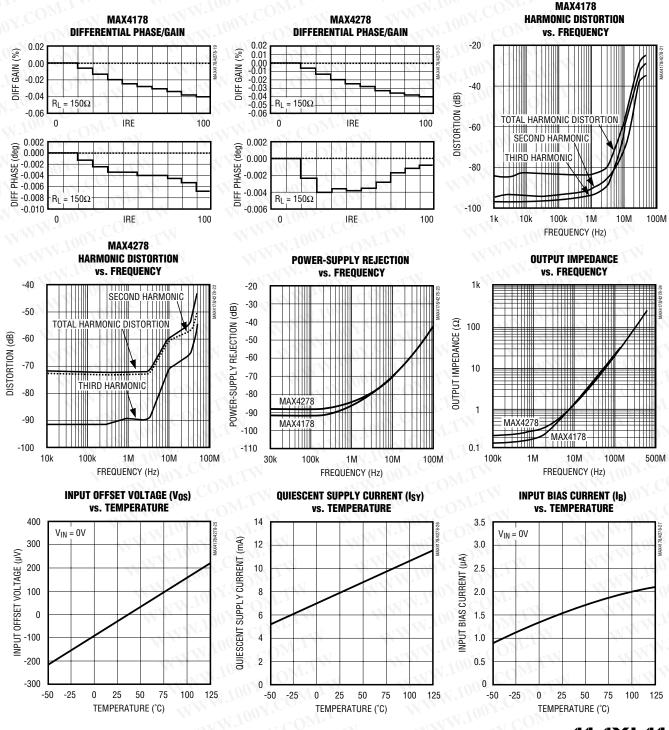
 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 100\Omega, C_L = 0pF, T_A = +25^{\circ}C$, unless otherwise noted.)



330MHz, Gain of +1/Gain of +2 Closed-Loop Buffers

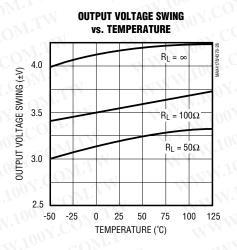
Typical Operating Characteristics (continued)

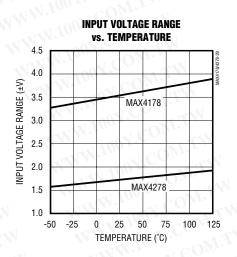
 $(V_{CC} = +5V, V_{EE} = -5V, R_L = 100\Omega, C_L = 0pF, T_A = +25^{\circ}C, unless otherwise noted.)$



Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = -5V, R_L = 100Ω , C_L = 0pF, T_A = +25°C, unless otherwise noted.)





Pin Description

PIN	001.	NAME	FUNCTION		
SO/µMAX/DIP	SOT23	NAME	FUNCTION		
1, 5, 8	<u>-0</u> V	N.C.	No Connection		
2	4	GND	Ground		
3	3	IN	Input		
4	2	VEE	Negative Power Supply. Connect to -5V.		
6	1.10	OUT	Output		
7	5	Vcc	Positive Power Supply. Connect to +5V.		

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Detailed Description

The MAX4178/MAX4278 are ±5V, wide-bandwidth, fast-settling, closed-loop buffers featuring high slew rate, high precision, high output current, low noise, and low differential gain and phase errors. The MAX4178, with a -3dB bandwidth of 330MHz, is preset for unity voltage gain (0dB). The MAX4278 is preset for a voltage gain of +2 (6dB) and has a 310MHz -3dB bandwidth.

These devices have a unique input stage that combines the benefits of a current-mode-feedback topology (high slew rate and low power) with those of a traditional voltage-feedback topology. This combination of architectures results in low input offset voltage and bias current, and high gain precision and power-supply rejection.

Under short-circuit conditions, the output current is typically limited to 150mA. This is low enough that a short to ground of any duration will not cause permanent damage to the chip. However, a short to either supply will create double the allowable power dissipation and may cause permanent damage if allowed to exist for longer than approximately 10 seconds. The high output-current capability is an advantage in systems that transmit a signal to several loads. See the *High-Performance Video Distribution Amplifier* section.

330MHz, Gain of +1/Gain of +2 Closed-Loop Buffers

Applications Information

Grounding, Bypassing, and PC Board Layout

In order to obtain the MAX4178/MAX4278s' full 330MHz/310MHz bandwidths, microstrip and stripline techniques are recommended in most cases. To ensure that the PC board does not degrade the amplifier's performance, it's a good idea to design the board for a frequency greater than 1GHz. Even with very short traces, it's good practice to use these techniques at critical points, such as inputs and outputs. Whether you use a constant-impedance board or not, observe the following guidelines when designing the board:

- Do not use wire-wrap boards. They are too inductive.
- Do not use IC sockets. They increase parasitic capacitance and inductance.
- In general, surface-mount components have shorter leads and lower parasitic reactance, giving better high-frequency performance than through-hole components.
- The PC board should have at least two layers, with one side a signal layer and the other a ground plane.
- Keep signal lines as short and straight as possible.
 Do not make 90° turns; round all corners.
- The ground plane should be as free from voids as possible.

On Maxim's evaluation kit, the ground plane has been removed from areas where keeping the trace capacitance to a minimum is more important than maintaining ground continuity.

Driving Capacitive Loads

The MAX4178/MAX4278 provide maximum AC performance with no output load capacitance. This is the case when the MAX4178/MAX4278 are driving a correctly terminated transmission line (e.g., a back-terminated 75 Ω cable). However, the MAX4178/MAX4278 are capable of driving capacitive loads up to 100pF without oscillations, but with reduced AC performance.

Driving large capacitive loads increases the chance of oscillations in most amplifier circuits. This is especially true for circuits with high loop gains, such as voltage followers. The amplifier's output resistance and the load

capacitor combine to add a pole and excess phase to the loop response. If the frequency of this pole is low enough and if phase margin is degraded sufficiently, oscillations may occur.

A second problem when driving capacitive loads results from the amplifier's output impedance, which looks inductive at high frequency. This inductance forms an L-C resonant circuit with the capacitive load, which causes peaking in the frequency response and degrades the amplifier's gain margin.

The MAX4178/MAX4278 drive capacitive loads up to 100pF without oscillation. However, some peaking (in the frequency domain) or ringing (in the time domain) may occur. This is shown in Figures 2a and 2b and the in the Small- and Large-Signal Pulse Response graphs in the *Typical Operating Characteristics*.

To drive larger-capacitance loads or to reduce ringing, add an isolation resistor between the amplifier's output and the load, as shown in Figure 1.

The value of RISO depends on the circuit's gain and the capacitive load. Figures 3a and 3b show the Bode plots that result when a 20Ω isolation resistor is used with a voltage follower driving a range of capacitive loads. At the higher capacitor values, the bandwidth is dominated by the RC network, formed by RISO and CL; the bandwidth of the amplifier itself is much higher. Note that adding an isolation resistor degrades gain accuracy. The load and isolation resistor form a divider that decreases the voltage delivered to the load.

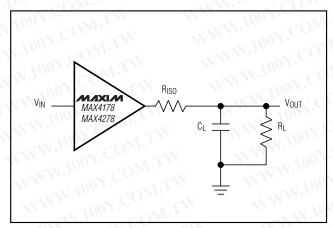


Figure 1. Capacitive-Load Driving Circuit

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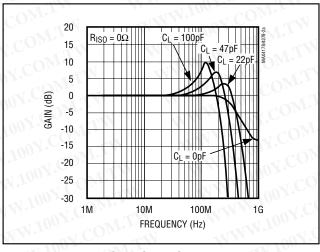


Figure 2a. MAX4178 Small-Signal Gain vs. Frequency with Capacitive Load

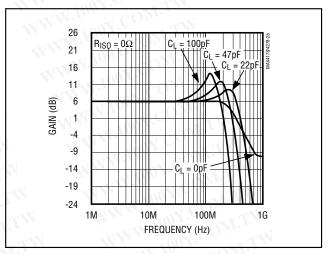


Figure 2b. MAX4278 Small-Signal Gain vs. Frequency with Capacitive Load

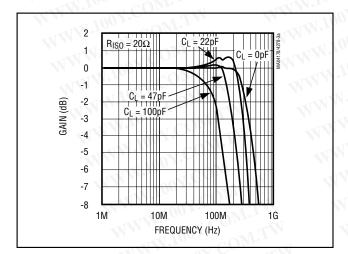


Figure 3a. MAX4178 Small-Signal Gain vs. Frequency with Capacitive Load and Isolation Resistor (R_{ISO})

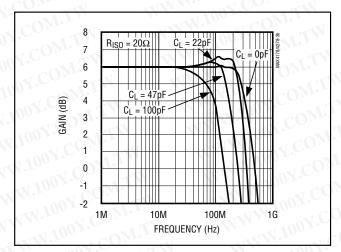


Figure 3b. MAX4278 Small-Signal Gain vs. Frequency with Capacitive Load and Isolation Resistor (R_{ISO})

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Flash ADC Preamp

The MAX4178/MAX4278s' high current-drive capability makes them well suited for buffering the low-impedance input of a high-speed flash ADC. With their low output impedance, these buffers can drive the inputs of the ADC with no loss of accuracy. Figure 4 shows a preamp for digitizing video, using the 250Msps MAX100 and the 500Msps MAX101 flash ADCs. Both of these ADCs have a 50Ω input resistance and a 1.2GHz input bandwidth.

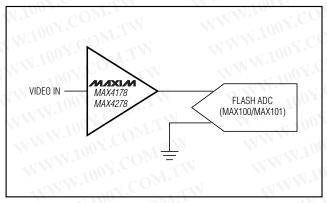


Figure 4. Preamp for Video Digitizer

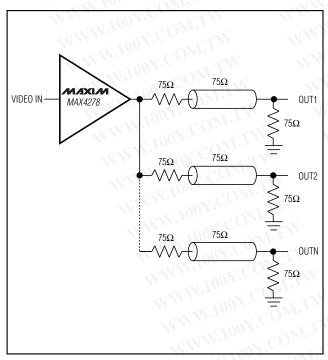


Figure 5. High-Performance Video Distribution Amplifier

High-Performance Video Distribution Amplifier

The MAX4278 (Ay = +2) makes an excellent driver for multiple back-terminated 75Ω video coaxial cables (Figure 5). The high current-output capability allows the attachment of up to six ± 2 Vp-p, 150Ω loads to the MAX4278 at ± 2 °C. With the output limited to ± 1 Vp-p, the number of loads may double. For multiple gain-of-2 video line drivers in a single package, refer to the MAX496/MAX497data sheet.

Ordering Information (continued)

PART	TEMP. RANGE	PIN- PACKAGE	SOT TOP MARK
MAX4278EPA	-40°C to +85°C	8 Plastic DIP	- W
MAX4278ESA	-40°C to +85°C	8 SO	- VIV
MAX4278EUA	-40°C to +85°C	8 µMAX	<u>-</u>
MAX4278EUK-T	-40°C to +85°C	5 SOT23-5	ABYY
MAX4278MJA	-55°C to +125°C	8 CERDIP	_ (T¥1)

Chip Information

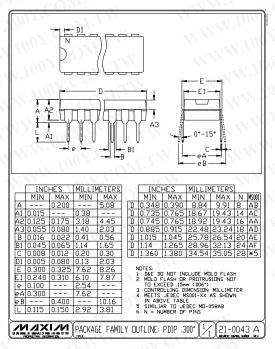
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SUBSTRATE CONNECTED TO VEE

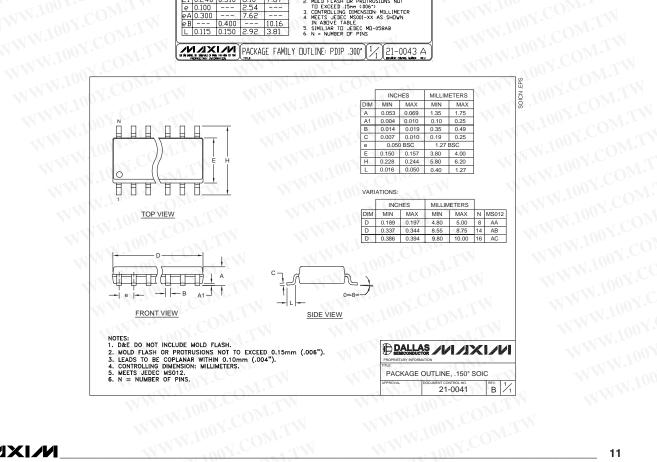
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WW.100Y.COM.TW 330MHz, Gain of +1/Gain of +2 **Closed-Loop Buffers**

Package Information

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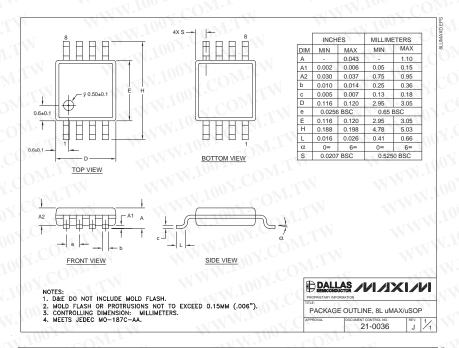
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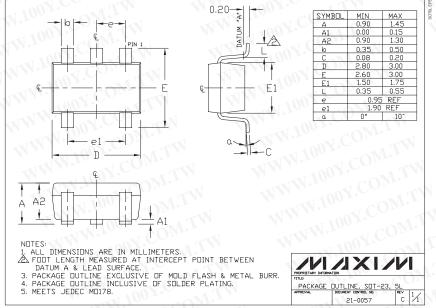
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Package Information (continued)

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