## SLIC KIT OPTIMIZED FOR APPLICATIONS WITH BOTH FIRST AND SECOND GENERATION COMBOS

－PROGRAMMABLE DC FEED RESISTANCE AND LIMITING CURRENT（ $25 / 40 / 60 \mathrm{~mA}$ ）
－LOW ON－HOOK POWER DISSIPATION （50mW typ）
－SIGNALLING FUNCTION（off－hook／GND－Key）
－QUICK OFF－HOOK DETECTION IN CVS FOR LOW DISTORTION（＜ 1 \％）DIAL PULSE DE－ TECTION
－HYBRID FUNCTION
－RINGING GENERATION WITH QUASI ZERO OUTPUT IMPEDANCE，ZERO CROSSING IN－ JECTION（no ext．relay needed）AND RING TRIP DETECTION
－ABSOLUTELY NO NOISE INJECTED ON ADIACENT LINES DURING RINGING SE－ QUENCE
－AUTOMATIC RINGING STOP WHEN OFF－ HOOK IS DETECTED
－TEST MODE ALLOWS LINE LENGHT MEAS－ UREMENT
－PARALLEL LATCHED DIGITAL INTERFACE
－LOW NUMBER OF EXTERNAL COMPO－ NENTS WITH STANDARD TOLERANCE ONLY： 9 1\％RESISTORS AND 5 10－20\％CA－ PACITORS（for 600 ohm appl．）
－POSSIBILITY TO WORK ALSO WITH HIGH COMMON MODE CURRENTS
－GOOD REJECTION OF THE NOISE ON BAT－ TERY VOLTAGE（ 20 dB at 10 Hz ；35dB at 1 KHz ）
－INTEGRATED THERMAL PROTECTION
－SURFACE MOUNT PACKAGE（PLCC28＋ PowerSO－20）
－ $0^{\circ} \mathrm{C}$ TO $70^{\circ} \mathrm{C}$ ：L3000N／L3092
－$-40^{\circ} \mathrm{C}$ TO $+85^{\circ} \mathrm{C}$ ： $\mathrm{L} 3000 \mathrm{NT} / \mathrm{L} 3092 \mathrm{~T}$

## DESCRIPTION

The SLIC KIT（L3000N／L3092）is a set of solid state devices designed to integrate many of the functions needed to interface a telephone line．It consists of 2 integrated devices ；the L3000N line interface circuit and the L3092 control unit．
The kit implements the main features of the BORSHT functions：
－Battery feed（balance mode）
－Ringing Injection

PRELIMINARY DATA

－Signalling Detection
－Hybrid Function
The SLIC KIT injects the ringing signal in bal－ anced mode and requires a positive supply volt－ age of typically +72 V to be available on the sub－ scriber card．
The L3000N／L3092 kit generates the ringing sig－ nal internally，avoiding the requirement for expen－ sive external circuitry．A low level 1.5 V rms input is required．（This can be provided by the combo）．
A special operating mode limits the SLIC KIT power dissipation to 50 mW in on－hook condition keeping the on／off hook detection circuit active．
Through the Digital Interface it is also possible to set an operating mode that allows measurements of loop resistance and therefore of line lenght．
This kit is fabricated using a 140 V Bipolar tech－ nology for L3000N and a 12 V Bipolar $I^{2} \mathrm{~L}$ technol－ ogy for L3092．
Both devices are available PTH application （FLEXIWATT15 and DIP28）or SMD application （PowerSO－20 and PLCC28）．
This kit is specially suitable to Private Automatic Branch Exchange（PABX）and Low Range C．O． Applications．

PIN CONNECTIONS


## ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{b}}-$ | Negative Battery Voltage | -80 | V |
| $\mathrm{~V}_{\mathrm{b}}+$ | Positive Battery Voltage | 80 | V |
| $\left\|\mathrm{~V}_{\mathrm{b}}-\left\|+\left\|\mathrm{V}_{\mathrm{b}+}\right\|\right.\right.$ | Total Battery Voltage | 140 | V |
| $\mathrm{~V}_{\mathrm{dd}}$ | Positive Supply Voltage | +6 | V |
| $\mathrm{~V}_{\mathrm{ss}}$ | Negative Supply Voltage | -6 | V |
| $\mathrm{~V}_{\mathrm{agnd}}-\mathrm{V}_{\mathrm{bgnd}}$ | Max. Voltage between Analog Ground and Battery Ground | 5 | V |
| $\mathrm{~T}_{\mathrm{j}}$ | Max. Junction Temperature | +150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{stg}}$ | Storage Temperature | -55 to +150 | ${ }^{\circ} \mathrm{C}$ |

## THERMAL DATA

| Symbol | Parameter |  | Value | Unit |
| :---: | :---: | :---: | :---: | :---: |
| L3000N HIGH VOLTAGE |  |  |  |  |
| $\mathrm{R}_{\text {th }}$-case | Thermal Resistance Junction to case (FLEXIWATT15) | Max. | 4 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{R}_{\text {th }} \mathrm{j}$-amb | Thermal Resistance Junction to ambient (FLEXIWATT15) | Max. | 50 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{R}_{\text {th }}$-case | Thermal Resistance Junction to case (PowerSO-20) | Typ. | 2 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\mathrm{R}_{\text {th }}$-amb | Thermal Resistance Junction to ambient (PowerSO-20) | Max. | 60 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| L3092 LOW VOLTAGE |  |  |  |  |
| $\mathrm{R}_{\text {th }} \mathrm{j}$-amb | Thermal Resistance Junction to ambient | Max. | 80 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## OPERATING RANGE

| Symbol | Parameter | Min. | Typ. | Max. | Unit |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{T}_{\text {oper }}$ | Operating Temperature Range for L3000N/L3092 |  |  |  |  |
|  |  | 0 <br> L3000NT/L3092T |  | 70 <br> +85 | ${ }^{\circ} \mathrm{C}$ <br> ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{V}_{\mathrm{b}-}$ | Negative Battery Voltage | -40 |  | -70 | -48 |
| $\mathrm{~V}_{\mathrm{b}}+$ | Positive Battery Voltage | 0 | +72 | +75 | V |
| $\left\|\mathrm{~V}_{\mathrm{b}}-\left\|+\left\|\mathrm{V}_{\mathrm{b}}+\right\|\right.\right.$ | Total Battery Voltage |  | 120 | 130 | V |
| $\mathrm{~V}_{\mathrm{dd}}$ | Positive Supply Voltage | +4.5 |  | +5.5 | V |
| $\mathrm{~V}_{\mathrm{ss}}$ | Negative Supply Voltage | -5.5 |  | -4.5 | V |

## PIN DESCRIPTION (L3000N)

| $\underset{\mathbf{N}^{\circ}}{\text { FLEX. }}$ | $\begin{gathered} \text { SO-P. } \\ \mathrm{N}^{\circ} . \end{gathered}$ | Name | Description |
| :---: | :---: | :---: | :---: |
| 1 | 3 | TIP | A line termination output with current capability up to 100 mA ( $\mathrm{I}_{\mathrm{s}}$ is the current sourced from this pin). |
| 2 | 4 | MNT | Positive Supply Voltage Monitor. |
| 3 | 5 | $\mathrm{V}_{\mathrm{B}+}$ | Positive Battery Supply Voltage. |
| 4 | 6 | BGND | Battery ground relative to the $\mathrm{V}_{\mathrm{S}_{+}}$and the $\mathrm{V}_{\mathrm{B}}$ supply voltages. It is also the reference ground for TIP and RING signals. |
| 5 | 7 | $\mathrm{V}_{\mathrm{DD}}$ | Positive Power Supply +5V. |
| 6 | 8 | $\mathrm{V}_{\text {IN }}$ | 2 wire unbalanced voltage input. |
| 7 | 9 | VBIM | Output voltage without current capability, with the following functions: give an image of the total battery voltage scaled by 40 to the low voltage part. - filter by an external capacitor the noise on . |
| 8 | $\begin{gathered} \hline 1,10,11, \\ 20 \\ \hline \end{gathered}$ | $\mathrm{V}_{\mathrm{B}}$ - | Negative Battery Supply Voltage. |
| 9 | 12 | AGND | Analog Ground. All input signals and the V $\mathrm{V}_{\text {DD }}$ supply voltage must be referred to this pin. |
| 10 | 13 | REF | Voltage reference output with very low temperature coefficient. The connected resistor sets Internal circuit bias current. |
| 11 | 14 | C1 | Digital signal input (3 levels) that defines device status with pin 12. |
| 12 | 15 | C2 | Digital signal input (3 levels) that defines device status with pin 11. |
| 13 | 16 | ${ }_{\text {IT }}$ | High precision scaled transversal line current signal. $I_{T}=\frac{I_{a}+I_{b}}{100}$ |
| 14 | 17 | IL | Scaled longitudinal line current signal. $\mathrm{I}_{\mathrm{L}}=\frac{\mathrm{I}_{\mathrm{a}}-\mathrm{I}_{\mathrm{b}}}{100}$ |
| 15 | 19 | RING | B line termination output with current capability up to 100 mA ( $\mathrm{l}_{\mathrm{b}}$ is the current sunk into this pin). |
| - | 2, 18 | N.C. | Not connected. |

Notes: 1) Unless otherwise specified all the diagrams in this datasheet refers to the FLEXIWATT15 pin connection.
2) All information relative to the PowerSO-20 package option should be considered as advanced information on a new product now in developement or undergoing evaluation. Details are subject to change without notice.

## PIN DESCRIPTION (L3092)

| $\mathrm{N}^{\circ}$ | Name | Description |
| :---: | :---: | :---: |
| 1 | VOUT | Two wire unbalanced output carryng out the following signals reduced by 40: <br> 1) DC voltage to perform the proper DC characteristic. <br> 2) Ringing Signal <br> 3) Voice Signal |
| 2 | RPC | AC line Impedance Adjustment Protection Resistances Compensation |
| 3 | TX | Transmit Amplifier Output |
| 4 | COMP | Comparator Input. This is the input comparator that senses the line voltage in power down and in automatic stand-by, allowing off hook detection in this mode. |
| 5 | AUT | Aut. Input. It is a part of the digital interface. Loaded when CS is low. |
| 6 | MR | Master Reset Input. When it is connected to ground the SLIC is forced in power down. It has an internal pull-up. (typ. 200K $\Omega$ ) (*) |
| 7 | PWON | Power on/power off input. This input is part of digital interface. Loaded when CS is low. |
| 8 | RING | Ring Enable Input. This input is part of the digital interface. Loaded when CS is low. |
| 9 | CS | Chip Select Input. |
| 10 | GDK | Ground Key Output Enabled by CS Low. |
| 11 | ONHK | On Hook/off Hook Output Enabled by CS Low. |
| 12 | C2 | State control Signal 2. |
| 13 | C1 | State Control Signal 1. Combination of C1 and C2 define operating mode of the high voltage part. |
| 14 | RGIN | Low Level Ringing Signal Input. |
| 15 | CRT | Ring Trip Detection |
| 16 | IL | Longitudinal Line Current Input $\mathrm{IL}=\frac{\mathrm{I}_{\mathrm{b}}-\mathrm{I}_{\mathrm{a}}}{100}$ |
| 17 | RDC | DC Feeding System |
| 18 | IT | Transversal Line Current Input $I T=\frac{I_{a}+I_{b}}{100}$ |
| 19 | ACDC | AC - DC Feedback Input. |
| 20 | VDD | Positive Supply Voltage, +5 V . |
| 21 | REF | Bias Setting Pin. |
| 22 | VSS | Negative Supply Voltage, -5V. |
| 23 | GND | Analog and Digital Ground. |
| 24 | LIM | Limiting Current Selection Input. Loaded when CS is low. |
| 25 | PDO | Power Down Output. Driving the high voltage part L3000N through the bias resistor RH. |
| 26 | ZB | TX Amplifier Negative Input performig the two to four wire conversion. In case of application with 2nd Generation COMBO performing also the echo cancellation (ex TS5070/5071), this pin must be connected to GND. |
| 27 | CAC | AC Feedback Input. |
| 28 | ZAC | AC Line Impedance Synthesis. |



## L3000N BLOCK DIAGRAM



## L3092 BLOCK DIAGRAM



FUNCTIONAL DIAGRAM


## FUNCTIONAL DESCRIPTION

## L3000N - HIGH VOLTAGE CIRCUIT

The L3000N line interface provides battery feed for telephone lines and ringing injection. Both these operations are done in Balance Mode. This is very important in order to avoid the generation of common mode signals in particular during the pulse dialling operation of the telephone set connected to the SLIC. The IC contains a state decoder that under external control can force the following operational modes : stand-by, conversation and ringing.
In addition Power down mode can be forced connecting the bias current resistor to $V_{D D}$ or leaving it open.
Two pins, IL and IT, carry out the information concerning line status which is detected by sensing the line current into the output stage.
The L3000N amplifies both the AC and DC signals entering at pin 6 (VIN) by a factor equal to 40 .
Separategrounds are provided:

- Analog ground as reference for analog signals
- Battery ground as a reference for the output stages
The two ground should be shorted together at a low impedance point.


## L3092- LOW VOLTAGE CIRCUIT

The L3092 Low Voltage Control Unit controls the L3000N line interface module providing set up data to set line feed characteristics and to inject ringing. An on chip digital parallel interface allows a microprocessor or a second generation COMBO as the TS5070 to control all the operations.
L3092 defines working states of Line Interface Circuit and also informs the card controller about line status.

## L3000N WORKING STATES

In order to carry out the different possible operations, the L3000N has several different working states. Each state is defined by the voltage respectively applied by pin 12 and 13 of L3092 to the pins 12 and 11 of L3000N.
Three different voltage levels $(-3,0,+3)$ are available at each connection, so defining nine possible states as listed in tab. 1.
Appropriate combinations of two pins define four of the five possible L3000N working states that are:
a) Stand-by (SBY)
b) Conversation (CVS)

Table 1.

|  |  | Pin 12 of L3092 Pin 12 of L3000N (C2) |  |  |
| :--- | :---: | :---: | :---: | :---: |
|  |  | $\mathbf{+ 3}$ | $\mathbf{0}$ | $\boldsymbol{- 3}$ |
| Pin 13 of L3092 | +3 | Stand-by | Conversation | Not Used |
| (C1) | 0 | Not Used | B.B | Not Used |
| Pin 11 of L3000N | -3 | Not Used | Ringing | Not Used |

c) Ringing (RING)
d) Boost Battery (BB),(see Appendix B).

The fifth status, Power down (PD), is set by the output pin PDO of the L3092 that disconnect the Bias Resistor, RH, of L3000N from ground.
The main difference between Stand-by and Power down is that in SBY the power consumption on the voltage battery $\mathrm{VB}-(-48 \mathrm{~V})$ is reduced but the L3000N DC Feeding and monitoring circuits are still active, in PD the power consumption on VB- is reduced to zero, and the L3000N is completely switched off.

## SLIC OPERATING MODES

Through the L3092 Digital Interface it is possible to select six different SLIC OPERATING MODES :

1) Conversation or Active Mode (CVS)
2) Stand - By Mode (SBY)
3) Power - Down Mode (PD)
4) Automatic Stand - By Mode (ASBY)
5) Test Mode (TS)
6) Ringing Mode (RNG)

## 1) CONVERSATION (CVS) OR ACTIVE MODE

This operating mode is set by the control processor when the Off hook condition has been recognized,
As far as the DC Characteristic is concerned two different feeding conditions are present:
a) Current limiting region : the DC impedance of the SLIC is very high ( $>20 \mathrm{~K} \Omega$ ) and therefore the system works like a current generator. By the L3092 Digital Interface it is possible to selects the value of the limiting current.:

$$
60 \mathrm{~mA}, 40 \mathrm{~mA} \text { or } 25 \mathrm{~mA} \text {. }
$$

b) A standard resistive feeding mode : the characteristic is equal to a battery voltage (VB-) minus 5 V , in series with a resistor, whose value is set by external components (see external component list of L3092).
Switching between the two regions is automatic without discontinuity, and depends on the loop resistance. The SLIC AC characteristics are guaranteed in both regions.
Fig. 1 shows the DC characteristic in conversa-
tion mode.
Fig. 2 shows the line current versus loop resistance for two different battery values and RFS = $200 \Omega$.
The allowed maximum loop resistance depends on the values of the battery voltage (VB), on the RFS and on the value of the longitudinal current (IGDK). With a battery voltage of 48V, RFS $=200 \Omega$ and $\operatorname{IGDK}=0 \mathrm{~mA}$, the maximum loop resistance is over $3000 \Omega$ and with IGDK $=20 \mathrm{~mA}$ is about $2000 \Omega$ (see Application Note on maximum loop resistance for L3000N/L3092 SLIC KIT).
In conversation mode the AC impedance at the line terminals is synthetized by the external components ZAC and RP, according to the following formula:

$$
\mathrm{ZML}=\frac{\mathrm{ZAC}}{25}+2 \cdot \mathrm{RP}
$$

Depending the characteristic of the ZAC network, ZML can be either a pure resistance or a complex impedance. This allows for ST SLIC to meet different standards as far as the return loss is concerned. The capacitor CCOMP guarantees stability to the system.
The two to four wire conversion is achieved by means of a circuit that can be represented as a Wheatstone bridge, the branches of which being:

1) The line impedance (Zline).
2) The SLIC impedance at line terminals (ZML).
3) The balancing network ZA connected between RX input and ZB pin of L3092.
4) The network ZB between ZB pin and ground that shall copy the line impedance.
It is important to underline that ZA and ZB are not equal to ZML and to Zline. They both must be multiplied by a factor in the range of 10 to 25 , allowing use of smaller capacitors.
In case the L3000N/L3092 kit is used with a second generation programmable COMBO (EG TS5070FN) which is able to perform the two to four wire conversion, the two impedances ZA and ZB can be removed and the ZB pin connected to GND. The -6dB Tx gain of the L3000N/L3092 SLIC kit in fact allows to keep the echo signal always within the COMBO Hybrid Balance Filter dynamic range. In conversation mode, the L3000N dissipates about 250 mW for its own operation. The dissipation related to the current supplied to the line shall be added, in order to get the total dissipation.

In the same condition the power dissipation of L3092 is typically 100 mW .

Figure 1: DC Characteristics in Conversation Mode


Figure 2: Line Current versus Loop Resistance -
RFS = 200 ; Limiting Currents: 25/40/60mA


## 2) STAND-BY (SBY) MODE

In this mode the bias currents of both L3000N and L3092 are reduced as only some parts of the two circuits are completely active, control interface and current sensors among them. The current supplied to the line is limited at 10 mA , and the slope of the DC characteristic corresponds to $2 \times$ RFS.
The AC characteristic in Stand-by corresponds to a low impedance ( $2 \times \mathrm{RP}$ )
In Stand-by mode the line voltage polarity is just in direct condition, that is the TIP wire more positive than the RING one as in Conversation Mode.

When the SLIC is in Stand-by mode, the power dissipation of L3000N does not exceed 120 mW from - 48V) eventually increased of a certain amount if some current is flowing into the line.
The power dissipation of the L3092 in the same condition is typically 50 mW .
SBY Mode is usually selected when the telephone is in on-hook. It allows a proper off-hook detection also in presence of high common mode line current or with telephone set sinking few milliAmpere of line current in on hook condition.

Figure 3: DC Characteristics in Stand-by Mode


## 3) POWER DOWN (PD) MODE

In this mode the L3000N present a high impedance ( $>1$ Mohm) to the line and cannot feed any line current.
The L3092 forces L3000N in Power Down disconnecting its bias Resistor, RH, from the ground through the output pin PDO.
The power dissipation from the battery voltage $(-\mathrm{VB})$ is almost equal to zero and the power dissipation of L3092 is typically 50 mW .
The PD mode is normally used in emergency condition but can be used also in normal on-hook condition.
In this case the off-hook detection is performed using the line sense comparator integrated in the L3092.
The fig. 4 shows the functional circuit to perform the off hook detection in Power down mode.
The resistor RR and RT feed the line current. The voltage at the terminal of the resistor RS connected to RING wire is normally -48 V .
When there is a loop resistor between TIP and

RING wires the voltage will increases to -24 V .
The comparator C1 will change its output voltage from low to high level.
If the Chip Select input (CS) is low the ONHK output pin will be set to low level (+ OV) indicating that the off hook condition is present.
This off-hook detection circuit can be influenced by common mode signal present on RING Terminal. The capacitor Cs is used to filter this common mode signal.
In the case of very high common mode signal af-
Figure 4: Off-hook Detection Circuit in Power Down Mode


Figure 5: Off-hook Detection Circuit in Automatic Standby Mode

4) AUTOMATIC STAND - BY (ASB) MODE

This is an operating mode similar to the Power Down Mode, but with the software procedure to detect off-hook condition integrated in hardware on chip.
Fig. 5 shows the functional circuit activated in this mode.
When the off-hook condition occurs RING wire voltage goes high (from -48 V to -24 V ).
The output of the comparator C 1 will go high setting the output of the flip - flop FF high.
Therefore L3092 will set L3000N in Stand-by providing a ground signal at pin PDO.
At the same time the external capacitor CINT will be slowly charged.
In Stand-by the internal off-hook Detection circuit will be activated and will check if the off-hook condition detected by the comparator C 1 was true or not true.
If the off-hook condition is confirmed the SLIC will be kept in Stand-by Mode and the output ONHK will go low when CS is low.
If the off-hook condition is not confirmed the SLIC will be kept in Stand - By only for a few seconds. (typ. 5 sec ). When the voltage at CRT out put will reach the VREF value the C2 comparator will reset the FF Flip - Flop and therefore the SLIC will be set again in Power Down.
The Automatic Stand-by (ASBY) Mode combine the key characteristics of Power Down (PD) and

Stand-by (SBY) Modes in particular it is characterized by a very low power consumption (as the Power Down mode) and a sophisticated off hook detection circuit (as the Stand-By mode).
The card controller will receive the off-hook information from the pin ONHK only after that it is checked and confirmed by the internal off-hook detector that is not sensitive to spikes and common mode line signal. Therefore the software required to manage the SLIC will be very simple.

## 5) TEST (TS) MODE

When this mode is activated the SLIC will be set in conversation mode keeping the initial value of limiting current.
The GDK output pin of L3092 Digital Interface will be set to "0" if the SLIC is operating in the limiting current region of the DC characteristic, see fig. 1 and 2. GDK output will be set to 1 if the SLIC is operating in the resistive region.
The SLIC will work in one of the two region depending on the loop resistance and the programmed limiting current value.
By changing the liming current value selected in conversation mode it is possible to measure the Loop Resistance and therefore the line lenght connected to the SLIC.
The following table shows the ranges of the loop resistance that set the GDK output pin to high and low level in correspondance of all the possible limiting current values ( $25 / 40 / 60 \mathrm{~mA}$ ) with RFS $=200 \Omega$.

| Limiting Current | GDK $=\mathbf{0}$ | GDK $=\mathbf{1}$ |
| :---: | :---: | :---: |
| 60 mA | $(0-300)$ ohm | $>300 \mathrm{ohm}$ |
| 40 mA | $(0-650) \mathrm{hm}$ | $>650 \mathrm{ohm}$ |
| 25 mA | $(0-1300) \mathrm{ohm}$ | $>1300 \mathrm{ohm}$ |

If, for example, the loop resistance is $400 \Omega$ the GDK output will be 0 only when the limiting current value is 40 or 25 mA .
The card controller can program consecutive Test Mode and Conversation Mode with different limiting current in order to individuate the range of loop resistance as shown in the flow chart of fig. 6.
The information of the Loop Resistance Range
can be very useful to optimize the transmission characteristics of the Line Card to each line.
For example, if a second generation COMBO like TS5070 is used the Card Controller can use this information to change the Tx, RX Gains and echo cancellation characteristics into the programmable COMBO improving the quality of the system.

Figure 6: Procedure for Loop Resistance Evaluation.


## 6) RINGING MODE

When the ringing function is selected by the control processor a low level signal ( 1.5 V rms) with a frequency in the range from 16 to 70 Hz , permanently applied to the L3092 (pin RGIN), is amplified and injected in balanced mode into the line through the L3000N with a super imposed DC voltage of 24 V typ.
This low level sinewave can be obtained also from COMBO connecting RGIN pin to RX COMBO output with a decoupling capacitor.
The first and the last ringing cycles are synchronized by the L3092 so that the ringing signal always starts and stops when the line voltage crosses zero.
When this mode is activated, the L3000N operates between the negative and the positive battery voltages typically -48 V and +72 V . The impedance to the line is just equal to the two external resistors (typ. 100 $\Omega$ ).
Ring trip detection is performed autonomously by the SLIC, without waiting for a command from the control processor, using a patented system which allows detection during a ringing burst ; when the off-hook condition is detected, the SLIC stops the ringing signal and forces the Conversation Mode.
In this condition, if $\mathrm{CS}=0 \mathrm{~V}$, the output pin ONHK goes to 0 V .
After the detection of the ONHK $=0$, the Card Controller must set the SLIC in Conversation Mode to remove the internal latching of the

On/Off hook information.

## CONTROL INTERFACE BETWEEN THE SLIC AND THE CARD CONTROLLER

The SLIC states and functions are controlled by microprocessor or interface latches of a second generation combo through seven wires that define a parallel digital interface.
The seven pins of the digital interface have the following functions :

- Chip select input (CS)
- Power on/off input (PWON)
- Ring enable input (RNG)
- Automatic SBY input (AUT)
- Limiting current input (LIM)
- On hook/Off hook detection output (ONHK)
- Ground Key detection output (GDK)

The four input pins PWON, RNG, AUT and LIM, set the status of the SLIC as shown in the following table.
The output pin ONHK is equals to 0 V when the line is in OFF hook condition (line $>7,5 \mathrm{~mA}$ ) and is equal to +5 V when the line is in On hook condition (line $<5,5 \mathrm{~mA}$ ).
The output pin GDK monitors the ground key function when the SLIC is in Conversation (CVS) Mode and the DC operating region (limiting or resistive) in Test (TS) Mode. When the SLIC is in Conversation (CVS) Mode and IGDK (longitudinal current) $>12 \mathrm{~mA}$, pin GDK is set to 0 V ;

| Operating Mode | Input Pin |  |  |  | Output Pin |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | RNG | PWON | AUT | LIM | ONHK | GDK |
| Conversation 25 mA Conversation 40 mA Conversation 60 mA | $\begin{aligned} & 0 \\ & 0 \\ & 0 \end{aligned}$ | $\begin{aligned} & 1 \\ & 1 \\ & 1 \end{aligned}$ | $\begin{aligned} & 1 \\ & 0 \\ & 0 \end{aligned}$ | $\begin{aligned} & X \\ & 1 \\ & 0 \end{aligned}$ | 1 on-hook 0 off-hook | 1 Ground key not detected. <br> 0 Ground key detected. |
| Stand-by | 0 | 0 | 0 | X |  | Disable |
| Automatic Stand-by | 1 | 0 | 1 | X |  |  |
| Power-down | 1 | 0 | 0 | X | C1 Comparator Output | Disable |
| Test Mode | 0 | 0 | 1 | X | 1 on-hook | 0 Limiting Region |
| Ringing (CVS 25mA) <br> Ringing (CVS 40mA) <br> Ringing (CVS 60mA) | 1 1 1 | 1 1 1 | $\begin{aligned} & 1 \\ & 0 \\ & 0 \end{aligned}$ | $\begin{aligned} & \hline X \\ & 1 \\ & 0 \end{aligned}$ | 0 off-hook | 1 Resistive Region Disable |

N.B. : When Ringing Mode is selected, you must choose also which of the three possible Conversation Modes. The SLIC will automatically select if Off-Hook condition will be detected during ringing.

When $\mathrm{I}_{\text {GDk }}<8 \mathrm{~mA}$, pin GDK set to +5 V
The longitudinal current (IGDK) is defined as follows:

$$
I_{G D K}=\frac{I_{b}-I_{a}}{2}
$$

Where $I_{a}$ is the current sourced from pin TIP and $\mathrm{l}_{\mathrm{b}}$ is the current sunk into pin RING.
The CS input pin allows to connect the I/O pins of the digital interfaces of many SLIC together.

It is possible to do it because :
When the $\mathrm{CS}=+5 \mathrm{~V}$ the output pins (ONHK, GDK) are in high impedance condition (> $100 \mathrm{~K} \Omega$ ). The signals present at the input pins are not transfered into the SLIC.
When the $\mathrm{CS}=0 \mathrm{~V}$ the output pins change in function of the values of the line current (line) and the longitudinal current (lgdk). The operating status of the SLIC are set by the voltage applied to the input pins.
The rising edge of the CS signal latches the signal applied to the input pins. The status of the SLIC will not change until the CS signal will be again equal to zero.
See timings fig 7 \& 8 .
An additional input pin MR (Master Reset) can be useful during the system start up phase or in emergency condition.
In fact when this pin is set to " 0 " the SLIC will be set in POWER DOWN MODE. This pin has an internal pull-up resistor of about $200 \mathrm{~K} \Omega$

## EXTERNAL COMPONENTS LIST

To set up the SLIC kit into operation, the following parameters have to be defined:

- The DC feeding resistance RFS, defined as the resistance of each side of the traditional feeding system (most common value for RFS are 200, 400 or 500).
- The AC input/output SLIC impedance at line terminals, ZML, to which the return loss measurement is refered. It can be real (typically $600 \Omega$ ) or complex.
- The equivalent $A C$ impedance of the line Zline used for evaluation of the trans-hybrid loss ( $2 / 4$ wire conversion). It is usually a complex impedance.
- The frequency of the ringing signal Fr (SLIC can work with this frequency ranging from 16 to 68 Hz ).
- The value of the two resistors RP in series with the line terminals ; main purpose of the a.m. resistors is to allow primary protection to fire..
With these assumptions the following components list is defined:

Figure 7: Typical Application Circuit


## EXTERNAL COMPONENT LIST FOR THE L3000N

| Component |  | Value |  | Involved Parameter or Function |
| :---: | :---: | :--- | :---: | :---: |
| Ref | $22.5 \mathrm{~K} \Omega \pm 2 \%$ | Bias Resistor |  |  |
| RH | 30 to $100 \Omega$ | Lines Series Resistor |  |  |
| RP | $47 \mu \mathrm{~F}-20 \mathrm{~V} \pm 20 \%$ | Battery Voltage Rejection |  |  |
| CDVB | $0.1 \mu \mathrm{~F}-100 \mathrm{~V} \pm 20 \%$ | Positive Battery Filter |  |  |
| CVB + | $0.1 \mu \mathrm{~F}-100 \mathrm{~V} \pm 20 \%$ (note 1) | Negative Battery Filter |  |  |
| CVB- | BAT49X (note 2) | Protective Shottky Diode |  |  |
| DS |  |  |  |  |

EXTERNAL COMPONENT LIST FOR THE L3092

| CVSS | $0.1 \mu \mathrm{~F}-15 \mathrm{~V}$ | Negative Supply Voltage Filter |
| :---: | :---: | :---: |
| CVDD | $0.1 \mu \mathrm{~F}-15 \mathrm{~V}$ | Positive Supply Voltage Filter |
| CAC | $47 \mu \mathrm{~F}-10 \mathrm{~V} \pm 20 \%$ | AC Path Decoupling |
| ZAC | $25 \times$ (ZML - 2xRP) | 2 Wire AC Impedance |
| CCOMP | $\frac{1}{2 \Pi f_{0}\left(50 R_{P}\right)} \text { with } f_{0}=200 \mathrm{KHz}$ | AC Loop Compensation |
| RPC | $25 \times(2 \times R P)$ | R Insertion Loss Compensation |
| RDC | $2 \times$ (RFS - RP) | DC Feeding Resistor (RDC > 200 2 ) |
| RL | $63.4 \mathrm{~K} \Omega \pm 1 \%$ | Bias Resistor |
| ZA | $\mathrm{K} \times \mathrm{Z}_{\mathrm{ML}}$ (note 3) | SLIC Impedance Balancing Network |
| ZB | $(\mathrm{K} \times \mathrm{Z} \text { line }) / /\left(\frac{25}{\mathrm{~K}} \times \mathrm{CCOMP}\right)(\text { note } 4)$ | Line Impedance Balancing Network |
| CINT | see Table 2 (note 5) | Ring Trip Detection Time Constant |
| RT | $47 \mathrm{~K} \Omega$ |  |
| RR | $47 \mathrm{~K} \Omega$ | Resistors used only in the automatic stand-by mode. |
| RS | $1.5 \mathrm{M} \Omega$ (note 6) |  |
| CS | 47 nF | To be used only if high common mode rejection in Aut. SBY mode and in Power Down mode is requested (note 7) |
| CMR | 100nF | To be used only if Power on reset requested. The capacitor value depends on $V_{D D}$ rise time. |

## Notes:

1) In case line cards with less than 7 subscribers are implemented CVB- capacitor should be equal to $680 \mathrm{nF} / \mathrm{N}$ where N is the number of subscriber per card.
2) This shottky diode or equivalent is necessary to avoid damage to the device during hot insertion or in all those cases when a proper power up sequenœ cannot be guaranteed. In case the Shottky diode is not implemented the power sequence should guarantee that VB+ is always the last supply applied at power on and the first removed at power off.
In case an other shottky diode type is adopted it must fulfill the following characteristics:
$V_{F}<450 \mathrm{mV} @ l_{\mathrm{F}}=\mathrm{n} \cdot 15 \mathrm{~mA}, \mathrm{~T}_{\text {amb }}=25^{\circ} \mathrm{C}$
$V_{F}<350 \mathrm{mV} @ \mathrm{I}_{\mathrm{F}}=\mathrm{n} \cdot 15 \mathrm{~mA}, \mathrm{~T}_{\mathrm{amb}}=50^{\circ} \mathrm{C}\left(\mathrm{T}_{\mathrm{jL} 3000}=90^{\circ} \mathrm{C}\right)$
$V_{F}<245 \mathrm{mV} @ \mathrm{I}_{\mathrm{F}}=\mathrm{n} \cdot 15 \mathrm{~mA}, \mathrm{~T}_{\mathrm{amb}}=85^{\circ} \mathrm{C}\left(\mathrm{T}_{\mathrm{jL} 3000}=120^{\circ} \mathrm{C}\right)$
Where n is the number of line sharing the same diode.
3) The structure of this network shall copy the SLIC output impedance multiplexed by a factor $K=10$ to 25 . This network must be removed when $2 / 4$ wire conversion is implemented with 2nd generation COMBO (EG. TS5070).
4) The structure of this network shall copy the line impedance, Zline, multiplexed by a factor $\mathrm{K}=10$ to 25 and compensate the effect of CCOMP on transhybrid rejection. This network must be removed when $2 / 4$ wire conversion is implemented with 2nd generation COMBO (EG. TS5070).
5) The CINT value depends on the ringing frequency $F_{R}$.
6) Value related to $\mathrm{Vb}=48 \mathrm{~V}$ application, for application with different battery voltages should be properly dimensioned (see Fig.4).
7) Ex.: For line leakage resistance to GND equal to $500 \mathrm{~K} \Omega$, the common mode rejection is $5 \mathrm{~V}_{\mathrm{P}}$ without CS and about 10 Vp with CS -

Table 2

| $\operatorname{Fr}(\mathrm{Hz})$ | $\mathbf{1 6 / 1 8}$ | $\mathbf{1 9 / 2 1}$ | $\mathbf{2 2 / 2 7}$ | $\mathbf{2 8 / 3 2}$ | $\mathbf{3 3 / 3 8}$ | $\mathbf{3 9} / 46$ | $\mathbf{4 7 / 5 5}$ | $56 / 68$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CINT (nF) | 680 | 580 | 470 | 390 | 330 | 270 | 220 | 180 |

The CINT value can be optimized experimentally for each application choosing the lower value that in correspondance of the lower ringing frequency,
the minimum line lenght and the higher number of ringers doesn't produce false off-hook detection.

ELECTRICAL CHARACTERISTICS $\left(\mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V} ; \mathrm{V}_{\mathrm{SS}}=-5 \mathrm{~V} ; \mathrm{V}_{\mathrm{B}+}=+72 \mathrm{~V} ; \mathrm{V}_{\mathrm{B}-}=-48 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=+25^{\circ} \mathrm{C}{ }^{(1)}\right)$
STANDBY

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| V LS | Output Voltage at L3000N Terminals | I Line $=0 \mathrm{~mA}$ |  | 43 |  | V |
| ILCC | Short Circuit Current |  | 8.8 |  | 12.5 | mA |
| lot | Off-hook Detection Threshold |  | 5.3 |  | 8.8 | mA |
| Hys | Off-hook/on-hook Hysteresis |  | 1.5 |  | 2.5 | mA |
| VIs | Simmetry to Ground |  |  |  | .75 | V |

## CONVERSATION

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {LO }}$ | Output Voltage at L3000N Terminals | I Line $=$ OmA |  | 43 |  | V |
| Ilim | Current Programmed Through the <br> LIM and AUT Inputs |  | $\lim$ <br> $-10 \%$ |  | $\lim$ <br> $+10 \%$ | mA |
| Iot | Off-hook Detection Threshold |  | 5.6 |  | 9.8 | mA |
| Hys | Off-hook/on-hook Hysteresis |  | 1.5 |  | 2.5 | mA |
| Ilgk | Longitudinal Line Current with GDK <br> Detect |  | 6.5 |  | 15 | mA |

## POWER-DOWN

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :--- | :--- | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{CN}}$ | Input Voltage at Pin COMP to Set <br> the Output Pin ONHK $=1$ |  |  | -100 | mV |  |
| $\mathrm{V}_{\text {CF }}$ | Input Voltage at Pin COMP to Set <br> the Output Pin ONHK $=0$ |  | 100 |  |  | mV |
| $\mathrm{I}_{\text {Com }}$ | Output Current at Pin COMP | COMP $=$ GND |  | 20 |  | $\mu \mathrm{~A}$ |

## SUPPLY CURRENT

| Symbol | Parameter |  | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IDD | Positive Supply Current $C S=1$ | Power Down/aut. Stand-by Stand-by Conversation Ringing |  | $\begin{gathered} \hline 5.7 \\ 7.5 \\ 11.7 \\ 11.3 \\ \hline \end{gathered}$ |  | $\begin{aligned} & \mathrm{mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \end{aligned}$ |
| Iss | Negative Supply Current $C S=1$ | Power Down/aut. Stand-by Stand-by Conversation Ringing |  | $\begin{aligned} & \hline 4.2 \\ & 4.2 \\ & 8.2 \\ & 8.2 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & \mathrm{mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \end{aligned}$ |
| $\mathrm{I}_{\text {BAT- }}$ | Negative Battery Supply Current Line Current $=0 \mathrm{~mA}$ | Power Down/aut. Stand-by Stand-by Conversation Ringing |  | $\begin{gathered} 0 \\ 2 \\ 5 \\ 5 \\ 14 \\ \hline \end{gathered}$ | $\begin{aligned} & 2.5 \\ & 6.5 \\ & 17 \end{aligned}$ | $\begin{aligned} & \mathrm{mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \\ & \mathrm{~mA} \end{aligned}$ |
| $\mathrm{I}_{\text {BAT }+}$ | Positive Battery Supply Current Line Current $=0 \mathrm{~mA}$ | Power Down/aut. Stand-by <br> Stand-by Conversation Ringing |  | $\begin{gathered} \hline 0 \\ 10 \\ 10 \\ 12 \\ \hline \end{gathered}$ | $\begin{gathered} 15 \\ 15 \\ 13.5 \\ \hline \end{gathered}$ | $\mu \mathrm{A}$ <br> $\mu \mathrm{A}$ <br> $\mu \mathrm{A}$ <br> mA |

AC OPERATION

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ZIx | Sending Output Impedance on TX |  |  |  | 15 | $\Omega$ |
| THD | Signal Distortion at 2W and 4W Terminals | Vtx = 0dBm @ 1020Hz |  |  | 0.3 | \% |
| RI | 2W Return Loss | $\mathrm{f}=300$ to 3400 Hz | 22 |  |  | dB |
| Thl | Transhybrid Loss | $\begin{aligned} & \mathrm{f}=300 \text { to } 3400 \mathrm{~Hz} \\ & 20 \log _{10}\left\|\frac{\mathrm{~V}_{\mathrm{R}}}{\mathrm{~V}_{\mathrm{S}}}\right\| \end{aligned}$ | 30 |  |  | dB |
| Gs | Sending Gain ${ }^{(2)}$ | Vso = 0dBm; f = 1020Hz | -6.27 | -6.02 | -5.77 | dB |
| Gsf | Sending Gain Flatness vs. Frequency | $\begin{aligned} & f=300 \text { to } 3400 \mathrm{~Hz} \\ & \text { Respect to } 1020 \mathrm{~Hz} \end{aligned}$ | -0.1 |  | +0.1 | dB |
| GI | Sending Gain Linearity | $\begin{aligned} & \mathrm{fr}=1020 \mathrm{~Hz} \\ & \text { Vsoref }=-10 \mathrm{dBm} \\ & \text { Vso }=+4 /-40 \mathrm{dBm} \end{aligned}$ | -0.1 |  | +0.1 | dB |
| Gr | Receiving Gain | $\mathrm{Vrl}=0 \mathrm{dBm} ; \mathrm{f}=1020 \mathrm{~Hz}$ | -0.25 |  | +0.25 | dB |
| Grf | Receiving Gain Flatness | $\begin{aligned} & \mathrm{f}=300 \text { to } 3400 \mathrm{~Hz} \\ & \text { Respect to } 1020 \mathrm{~Hz} \\ & \hline \end{aligned}$ | -0.1 |  | +0.1 | dB |
| Grf | Receiving Gain Linearity | $\begin{aligned} & \mathrm{fr}=1020 \mathrm{~Hz} \\ & \mathrm{Vrlref}=-10 \mathrm{dBm} \\ & \mathrm{Vrl}=+4 /-40 \mathrm{dBm} \end{aligned}$ | -0.1 |  | +0.1 | dB |
| Np4W | Psophomet. Noise 4W - Tx Terminals |  |  | -79 | -74 | dBmp |
| Np4W | Psophomet. at Line Terminals |  |  | -75 | -70 | dBmp |
| SVRR | Supply Voltage Rejection Ratio Relative to VB- | $\mathrm{f}=10 \mathrm{~Hz} \mathrm{Vn=100mVrms}$ |  | -20 |  | dB |
|  |  | $\mathrm{f}=1 \mathrm{KHz} \mathrm{Vn=100mVrms}$ |  |  | -35 | dB |
|  |  | $\mathrm{f}=3.4 \mathrm{KHz}$ Vn $=100 \mathrm{mVrms}$ |  |  | -30 | dB |
| Ltc | Longitudinal to Transversal Conversion | $\begin{aligned} & f=300 \text { to } 3400 \mathrm{~Hz} \\ & I \text { line }=30 \mathrm{~mA} \\ & Z M L=600 \Omega \end{aligned}$ |  |  |  |  |
| TIc | Transversal to Longitudinal Conversion |  | 48 | 51 |  | dB |

Notes:
(*) 52 dB using selected L3000N
(1) The datasheet certifies the electrical characteristics at $25^{\circ} \mathrm{C}$. For applications requiring operations in the standard temperature range $\left(0^{\circ} \mathrm{C}\right.$ to $70^{\circ} \mathrm{C}$ ) use $\mathrm{L} 3000 \mathrm{~N} / \mathrm{L} 3092$. If operations are required in the extended temperature range $\left(-40^{\circ} \mathrm{C}\right.$ to $\left.+85^{\circ} \mathrm{C}\right)$, use the kit $\mathrm{L} 3000 \mathrm{NT} / \mathrm{L} 3092 \mathrm{~T}$.
(2) value optimized for programmable COMBO Hybrid Balance Filter

RINGING PHASE

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vir | Superimposed DC Voltage | Rloop > $100 \mathrm{~K} \Omega$ | 19 |  | 27 | V |
|  |  | Rloop $=1 \mathrm{~K} \Omega$ | 17 |  | 25 | V |
| Vacr | Ringing Siganl at Line Terminal | $\begin{aligned} & \text { Rloop }>100 \mathrm{k} \Omega \\ & \mathrm{~V}_{\mathrm{RGN}}=1.5 \mathrm{Vrms} / 30 \mathrm{~Hz} \end{aligned}$ | 56.0 |  |  | Vrms |
|  |  | $\begin{aligned} & \text { Rloop }=1 \mathrm{~K} \Omega+1 \mu \mathrm{~F} \\ & \mathrm{~V} \text { RGN }=1.5 \mathrm{Vrms} / 30 \mathrm{~Hz} \end{aligned}$ | 56.0 |  |  | Vrms |
| If | DC Off-hook Del Threshold |  |  | 5.5 |  | mA |
| 1 lim | Output Current Capability |  | 85 |  | 130 | mA |
| Vrs | Ringing Symmetry |  |  |  | 2 | Vrms |
| THDr | Ringing Signal Distortion |  |  |  | 5 | \% |
| Zir | Ringing Amplicat. Input Impedance | L3092's Pin RGIN | 50 |  |  | $\mathrm{K} \Omega$ |
| Vrr | Residual of Ringing Signal at Tx Output |  |  |  | 100 | mVrms |
| Trt | Ring Trip Detection Time | $\begin{aligned} & \text { fring }=25 \mathrm{~Hz}(\mathrm{~T}=1 / \text { fring }) \\ & \mathrm{CINT}=470 \mathrm{nF} \end{aligned}$ |  | 80(3T) |  | ms |
| Toh | Off-hook Status Delay after the Ringing Stop |  |  |  | 50 | $\mu \mathrm{s}$ |

## DIGITAL INTERFACE ELECTRICAL CHARACTERISTICS

$\left(\mathrm{V}_{\mathrm{DD}}=+5 \mathrm{~V} ; \mathrm{V}_{\mathrm{SS}}=-5 \mathrm{~V} ; \mathrm{T}_{\mathrm{amb}}=25^{\circ} \mathrm{C}{ }^{(1)}\right)$
STATIC ELECTRICAL CHARACTERISTICS

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vil | Input Voltage at Logical "0" | Pins CS PWON LIM | 0 |  | 0.8 | V |
| Vih | Input Voltage at Logical "1" |  | 2 |  | 5 | V |
| Vil | Input Voltage at Logical "0" | Pins RNG-AUT | 0 |  | 0.5 | V |
| Vih | Input Voltage at Logical "1" |  | 2.3 |  | 5 | V |
| lil | Input Current at Logical "0" | All logic pins $\begin{gathered}\mathrm{Vil}=0 \mathrm{~V} \\ \mathrm{Vih}=5 \mathrm{~V}\end{gathered}$ |  |  | 15 | $\mu \mathrm{A}$ |
| lih | Input Current at Logical "1" |  |  |  | 25 | $\mu \mathrm{A}$ |
| Vol | Output Voltage at Logical "0" | $\begin{aligned} & \text { Pins ONHK GDK } \\ & \text { lout }=-1 \mathrm{~mA} \\ & \text { lout }=1 \mathrm{~mA} \end{aligned}$ |  |  | 0.4 | V |
| Voh | Output Volatge at Logical "1" |  | 2.4 |  |  | V |
| IIk | Tristate Leak Current | CS = "1" |  |  | 10 | $\mu \mathrm{A}$ |
| IMR | Pull-up MR Output Current | MR = "0" |  | 50 |  | $\mu \mathrm{A}$ |

## DYNAMIC ELECTRICAL CHARACTERISTICS

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :--- | :---: | :---: | :---: | :---: | :---: |
| Tsd | PWON, RING, AUT, LIM |  | 1500 |  |  | ns |
| Thd | PWON, RING, AUT, LIM |  | 0 |  |  | ns |
| Tww | CS Impulse Width (writing op.) |  | 1500 |  |  | ns |
| Thv | ONHK, GDK Data Out to "0" CS <br> Delay |  |  |  | 600 | ns |
| Tvh | ONHK, GDK High Imped. to "1 "CS <br> Delay |  |  | 600 | ns |  |
| Twr | CS Impulse Width (writing op.) |  | 800 |  |  | ns |

(1) The datasheet certifies the electrical characteristics at $25^{\circ} \mathrm{C}$. For applications requiring operations in the standard temperature range $\left(0^{\circ} \mathrm{C}\right.$ to $70^{\circ} \mathrm{C}$ ) use $\mathrm{L} 3000 \mathrm{~N} / \mathrm{L} 3092$. If operations are required in the extended temperature range $\left(-40^{\circ} \mathrm{C}\right.$ to $\left.+85^{\circ} \mathrm{C}\right)$, use the kit $\mathrm{L} 3000 \mathrm{NT} / \mathrm{L} 3092 \mathrm{~T}$.

Figure 8: Writing Operating Timing (controller to SLIC).


Figure 9: Reading Operating Timing (from SLIC to controller).


Figure 10: Test Circuit

$\mathrm{A}, \mathrm{B}, \mathrm{C}, \mathrm{D}$ are test reference points used during testing.

Figure 11: Typical Application Circuit with 2nd Generation COMBO for Complete Subscriber Circuit (Protection- SLIC - COMBO).


Figure 12: Typical Application Circuit with 1st Generation COMBO for Complete Subscriber Circuit (Protection - SLIC - COMBO).


## APPENDIX A

SLIC TEST CIRCUITS

Referring to the test circuit reported at the end of each SLIC data sheet here below you can find the proper configuration for each measurement. In particular:

A-B: Line terminals
C: Tx sending output on 4 W side
D: Rx receiving input on 4W Side
E : TTx teletaxe signal input
Rain: low level ringing signal input.

TEST CIRCUITS
Figure A1: Symmetry to Ground


Figure A2: 2W Returns Loss


TEST CIRCUITS (continued)
Figure A3: Trans-hybrid Loss.


Figure A4: Sending Gain


Figure A5: Receiving Gain


TEST CIRCUITS (continued)
Figure A6: PSRR Relative to Battery Voltage VB-


Figure A7: Longitudinalto Transversal Conversion


Figure A8: Longitudinalto Transversal Conversion


TEST CIRCUITS (continued)
Figure A9: TTX Level at Line Terminals


Figure A10: Ringing Simmetry


## APPENDIX B

## ADDITIONAL OPERATING FEATURES

Two further operating modes are provided on the L3092, boosted battery and ring pause. Both of these Modes are accessed by appllying a high impedance on inputs AUT and or RING of the digital interface.

## 1.Boosted Battery (BB)

This operating mode is equivalent to conversation mode with respect to $A C$ and signaling functions but with the following changes to the DC characteristics:
a) Current limiting value fixed at 25 mA .
b) Characteristic in the resistive feeding region corresponds to a battery voltage equal to $\left(-5+|V B-|+\mathrm{VB}_{+}\right)$Volt in series with the same feeding resistor utilized in the DC characteristic of conversation mode.
BB mode is typically used to feed long lines ( $20 \mathrm{~mA} / 4 \mathrm{Kohm}$ ) and to implement special functions such as message waiting where high voltage signals are required.

Further information about this opersating mode may be found by referring to the L3000/L3030 datasheet.

## 2.Ringing Pause Mode

During Ring Pause - Mode the SLIC is always in ringing mode but the AC ringing signal is not injected into the line. This mode allows to avoid any common mode voltage variation of TIP and RING wire during the transition bteween Ringing Burst and Ringing Pause. This feature is used in application where it is mandatory to avoid perturbations on adjacent lines during ringing injection. For example when in the same system analog lines are used both for speech and modem transmission.
The following table shows all operating modes of L3000/L3092 SLIC KIT. Boosted Battery or Ringing Pause Modes are selected by applying a high impedance (HI) to input pins RNG and/or AUT.
Included also in this table are the operating modes to which the SLIC defaults automatically during ringing mode when OFF HOOK is detected.

## CONTROL INTERFACE BETWEEN THE SLIC AND THE CARD CONTROLLER

| Operating Mode | Input Pin |  |  |  | Output Pin |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | RNG | PWON | AUT | LIM | ONHK | GDK |
| Conversation 25 mA | 0 | 1 | 1 | X | 1 on-hook 0 off-hook | 1 Ground key not detected <br> 0 Ground key detected |
| Conversation 40 mA | 0 | 1 | 0 | 1 |  |  |
| Conversation 60mA | 0 | 1 | 0 | 0 |  |  |
| Boosted Battery 25mA | 0 | 1 | HI | X |  |  |
| Stand-by | 0 | 0 | 0 | X |  | Disable |
| Automatic Stand-by | 1 | 0 | 1 | X |  |  |
| Power Down | 1 | 0 | 0 | X | C1 Comparator Output | Disable |
| Test Mode | 0 | 0 | 1 | X | 1 on-hook 0 off-hook | 0 Limiting Region <br> 1 Resistive Region |
| Ringing Inj. (CVS 25mA) | 1 | 1 | 1 | $x$ | 1 on-hook 0 off-hook | Disable |
| Ringing Inj. (CVS 40mA) | 1 | 1 | 0 | 1 |  |  |
| Ringing Inj. (CVS 60mA) | 1 | 1 | 0 | 0 |  |  |
| Ringing Inj. (BB 25mA) | 1 | 1 | HI | X |  |  |
| Ringing Pause (CVS | HI | 1 | 1 | X |  |  |
| 25mA) | HI | 1 | 0 | 1 |  |  |
| Ringing Pause (CVS | HI | 1 | 0 | 0 |  |  |
| 40mA) |  |  |  |  |  |  |
| Ringing Pause (CVS 60 mA ) |  |  |  |  |  |  |
| Ringing Pause (BB 25mA) | HI | 1 | HI | X |  |  |

NB:
$\mathrm{HI}=$ High Impedance
BB = Boosted Battery

## APPENDIX C

## LAYOUTSUGGESTIONS

Standard layout rules should be followewd in order to get the best system performances:

1) Use always 100 nF filtering capacitor close to the supply pins of each I.C.
2) Connect together BGND and AGND at a low impedance point. (e.g. on a ground
plane common to the line card).
3) The L3092 bias resistor (RL) should be connected close to the corresponding pins of L3092 (REF and GND).
Avoid any digital line to pass close to REF pin.
Eventually screen REF pin with a GND track.

PLCC28 PACKAGE MECHANICAL DATA

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | 12.32 |  | 12.57 | 0.485 |  | 0.495 |
| B | 11.43 |  | 11.58 | 0.450 |  | 0.456 |
| D | 4.2 |  | 4.57 | 0.165 |  | 0.180 |
| D1 | 2.29 |  | 3.04 | 0.090 |  | 0.120 |
| D2 | 0.51 |  |  | 0.020 |  |  |
| E | 9.91 |  | 10.92 | 0.390 |  | 0.430 |
| e |  | 1.27 |  |  | 0.050 |  |
| e3 |  | 7.62 |  |  | 0.018 |  |
| F |  | 0.46 |  |  | 0.028 |  |
| F1 |  | 0.71 |  |  |  |  |
| G |  |  |  |  |  | 0.04 |
| M |  | 1.24 |  |  |  |  |
| M1 |  |  |  |  |  |  |



PowerSO-20(SLUG UP) PACKAGE MECHANICAL DATA

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A |  |  | 3.70 |  |  | 0.145 |
| a1 | 0 |  | 0.25 | 0 |  | 0.01 |
| b | 0.40 |  | 0.53 | 0.016 |  | 0.021 |
| c | 0.23 |  | 0.32 | 0.009 |  | 0.012 |
| D | 15.80 |  | 16.00 | 0.622 |  | 0.63 |
| D1 | 9.4 |  | 9.80 | 0.37 |  | 0.385 |
| E | 13.90 |  | 14.50 | 0.547 |  | 0.57 |
| e |  | 1.27 |  |  | 0.05 |  |
| e3 |  | 11.43 |  |  | 0.45 |  |
| E1 | 10.90 |  | 11.10 | 0.429 |  | 0.437 |
| E2 |  |  | 2.90 |  |  | 0.114 |
| E3 | 5.80 |  | 6.20 | 0.228 |  | 0.244 |
| G | 0 |  | 0.10 | 0 |  | 0.004 |
| h |  |  | 1.10 |  |  | 0.043 |
| L | 0.80 |  | 1.10 | 0.031 |  | 0.043 |
| N | $10^{\circ}$ (Max.) |  |  |  |  |  |
| S | $8^{\circ}$ (Max.) |  |  |  |  |  |




PSO20DME

## DIP28 PACKAGE MECHANICAL DATA

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| a1 |  | 0.63 |  |  | 0.025 |  |
| b |  | 0.45 |  |  | 0.018 |  |
| b1 | 0.23 |  | 0.31 | 0.009 |  | 0.050 |
| b2 |  | 1.27 |  |  |  | 0.012 |
| D |  |  | 37.34 |  | 0.598 |  |
| E | 15.2 |  | 16.68 |  | 1.300 |  |
| e |  | 2.54 |  |  |  |  |
| e3 |  | 33.02 |  |  |  |  |
| F |  |  |  |  |  | 0.1750 |
| L |  |  |  |  |  |  |



FLEXIWATT 15 PACKAGE MECHANICAL DATA

| DIM. | mm |  |  | inch |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A |  |  | 5.00 |  |  | 0.196 |
| B |  |  | 1.90 |  |  | 0.074 |
| b1 |  |  | 0.1 |  |  | 0.004 |
| D | $4^{\circ}$ (typ.) |  |  |  |  |  |
| E |  | 0.30 |  |  | 0.012 |  |
| F |  | 0.90 |  |  | 0.035 |  |
| F1 |  |  | 0.57 |  |  | 0.022 |
| G | 1.77 | 1.9 | 2.03 | 0.070 | 0.075 | 0.080 |
| G1 |  | 26.77 |  |  | 1.054 |  |
| H1 |  | 29.00 |  |  | 1.142 |  |
| H2 |  | 28.00 |  |  | 1.102 |  |
| H3 |  | 17.00 |  |  | 0.669 |  |
| H4 |  | 0.80 |  |  | 0.031 |  |
| L | 19.05 |  | 19.95 | 0.75 |  | 0.785 |
| L1 | 1.10 |  | 1.40 | 0.043 |  | 0.055 |
| L2 | 2.60 |  | 2.90 | 0.102 |  | 0.114 |
| L3 | 15.35 |  | 15.65 | 0.604 |  | 0.616 |
| N1 |  | 10 |  |  | 0.394 |  |
| N3 |  | 6.8 |  |  | 0.268 |  |
| N4 |  | 3.8 |  |  | 0.15 |  |
| Dia1 |  | 13.00 |  |  | 0.511 |  |
| Dia2 |  | 14.00 |  |  | 0.551 |  |
| Dia3 |  | 2.50 |  |  | 0.098 |  |
| Dia4 |  | 12.00 |  |  | 0.472 |  |



## L3000N - L3092

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