



OPA350 OPA2350 OPA4350

SBOS099C - SEPTEMBER 2000 - REVISED JANUARY 2005

High-Speed, Single-Supply, Rail-to-Rail OPERATIONAL AMPLIFIERS

MicroAmplifier™ Series

FEATURES

- RAIL-TO-RAIL INPUT
- RAIL-TO-RAIL OUTPUT (within 10mV)
- WIDE BANDWIDTH: 38MHz
 HIGH SLEW RATE: 22V/us
- LOW NOISE: 5nV/√Hz
- LOW THD+NOISE: 0.0006%
- UNITY-GAIN STABLE
- MicroSIZE PACKAGES
- SINGLE, DUAL, AND QUAD

APPLICATIONS

- CELL PHONE PA CONTROL LOOPS
- DRIVING A/D CONVERTERS
- VIDEO PROCESSING
- DATA ACQUISITION
- PROCESS CONTROL
- AUDIO PROCESSING
- COMMUNICATIONS
- ACTIVE FILTERS
- TEST EQUIPMENT

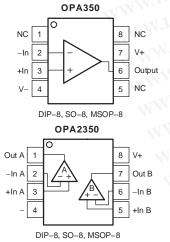
DESCRIPTION

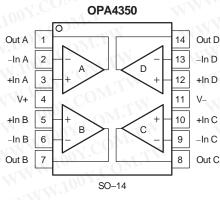
The OPA350 series rail-to-rail CMOS operational amplifiers are optimized for low voltage, single-supply operation. Rail-to-rail input/output, low noise ($5nV/\sqrt{Hz}$), and high speed operation (38MHz, $22V/\mu s$) make them ideal for driving sampling Analog-to-Digital (A/D) converters. They are also well suited for cell phone PA control loops and video processing (75Ω drive capability) as well as audio and general purpose applications. Single, dual, and quad versions have identical specifications for maximum design flexibility.

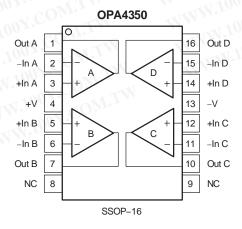
The OPA350 series operates on a single supply as low as 2.5V with an input common-mode voltage range that extends 300mV below ground and 300mV above the positive supply. Output voltage swing is to within 10mV of the supply rails with a $10k\Omega$ load. Dual and quad designs feature completely independent circuitry for lowest crosstalk and freedom from interaction.

The single (OPA350) and dual (OPA2350) come in the miniature MSOP-8 surface mount, SO-8 surface mount, and DIP-8 packages. The quad (OPA4350) packages are the space-saving SSOP-16 surface mount and SO-14 surface mount. All are specified from –40°C to +85°C and operate from –55°C to +150°C.

SPICE model available at www.ti.com







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ABSOLUTE MAXIMUM RATINGS(1)

Supply Voltage	7.0V
Signal Input Terminals ⁽²⁾ , Voltage	$(V-) - 0.3V$ to $(V+) + 0.3V$
Current	10mA
Open Short-Circuit Current(3)	Continuous
Operating Temperature Range	55°C to +150°C
Storage Temperature Range	55°C to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C

- (1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.3V beyond the supply rails should be current limited to 10mA or less.
- (3) Short-circuit to ground, one amplifier per package.

ELECTROSTATIC DISCHARGE SENSITIVITY



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

勝 特 力 材 料 886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787 Http://www.100y.com.tw

PACKAGE/ORDERING INFORMATION(1)

PRODUCT	PACKAGE-LEAD	PACKAGE DESIGNATOR	SPECIFIED TEMPERATURE RANGE	PACKAGE MARKING	ORDERING NUMBER	TRANSPORT MEDIA, QUANTITY
SINGLE	MM. CO	TW	MMA. 1002	I.CO.	N WW	11007.00
ODAGEGEA	MOOD o	DOK	1000 1 - 0500	COE	OPA350EA/250	Tape and Reel, 250
OPA350EA	MSOP-8	DGK	-40°C to +85°C	C50	OPA350EA/2K5	Tape and Reel, 2500
ODAGEGUA	(V) 00 0 00 V.	TW	4000 to 10000	OBASEOUA	OPA350UA	Rails
OPA350UA	SO-8	COMPD	-40°C to +85°C	OPA350UA	OPA350UA/2K5	Tape and Reel, 2500
OPA350PA	DIP-8	P	-40°C to +85°C	OPA350PA	OPA350PA	Rails
DUAL	WW. 100	Y.C.	W	1007.	M.T.V	W. 1001.
004005054	MOOD	V.COPOK TW	1000 1- 10500	1057	OPA2350EA/250	Tape and Reel, 250
OPA2350EA	MSOP-8	DGK	-40°C to +85°C	D50	OPA2350EA/2K5	Tape and Reel, 2500
ODAGGEGLIA	00.0	DMT	4000 to 10000	ODAGOCOLIA	OPA2350UA	Rails
OPA2350UA	SO-8	ON CONTRACT	-40°C to +85°C	OPA2350UA	OPA2350UA/2K5	Tape and Reel, 2500
OPA2350PA	DIP-8	POM	-40°C to +85°C	OPA2350PA	OPA2350PA	Rails
QUAD	W. T.	100 r. COM	11.	XXXV.100	COM	WW.
ODA 40505A	0000.40	110000	1000 1- 10500	ODA 40505A	OPA4350EA/250	Tape and Reel, 250
OPA4350EA	SSOP-16	DBQ	-40°C to +85°C	OPA4350EA	OPA4350EA/2K5	Tape and Reel, 2500
ODA 425011A	00.44	W.Tuo	4000 to 10000	ODA 425011A	OPA4350UA	Rails
OPA4350UA	SO-14	D	-40°C to +85°C	OPA4350UA	OPA4350UA/2K5	Tape and Reel, 2500

⁽¹⁾ For the most current package and ordering information, see the Package Option Addendum located at the end of this data sheet.

OPA350 OPA2350



ELECTRICAL CHARACTERISTICS: $V_S = 2.7V$ to 5.5V Boldface limits apply over the temperature range, $T_A = -40$ °C to +85°C. $V_S = 5V$.

All specifications at $T_A = +25$ °C, $R_L = 1$ k Ω connected to $V_S/2$ and $V_{OUT} = V_S/2$, unless otherwise noted.

OFFSET VOLTAGE Input Offset Voltage VOS VS = 5V ±150 ± TA = -40°C to +85°C vs Temperature TA = -40°C to +85°C ±4 ±5 ±5 ±5 ±5 ±5 ±5 ±5 ±6	### UNIT ### ±500 μV ### μV/°C ### 150 μV/V ### 175 μV/V ### μV/V
Input Offset Voltage VOS VS = 5V ±150 ± TA = -40°C to +85°C vs Temperature TA = -40°C to +85°C ±4 40	±1 mV μV/°C 150 μV/V μV/V
$ T_{A} = -40^{\circ} C \text{ to } +85^{\circ} C \\ \text{vs Temperature} \\ \text{vs Power-Supply Rejection Ratio} \\ T_{A} = -40^{\circ} C \text{ to } +85^{\circ} C \\ \text{VS} = 2.77 \text{ to } 5.5 \text{V, V}_{CM} = 0 \text{V} \\ \text{VS} = 2.77 \text{ to } 5.5 \text{V, V}_{CM} = 0 \text{V} \\ \text{VS} = 2.77 \text{ to } 5.5 \text{V, V}_{CM} = 0 \text{V} \\ \text{Channel Separation (dual, quad)} \\ \text{INPUT BIAS CURRENT} \\ \text{Input Bias Current} \\ \text{Input Bias Current} \\ \text{Input Offset Current} \\ \text{Input Offset Current} \\ \text{Input Voltage Noise, } f = 100 \text{Hz to } 400 \text{kHz} \\ \text{Input Voltage Noise Density, } f = 10 \text{kHz} \\ \text{Input Voltage Noise Density, } f = 10 \text{kHz} \\ \text{Current Noise Density, } f = 10 \text{kHz} \\ \text{Current Noise Density, } f = 10 \text{kHz} \\ \text{Common-Mode Voltage Range} \\ \text{Common-Mode Voltage Range} \\ \text{VCM} \\ \text{CMRR} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} \\ \text{SONUSE} \\ \text{Input Voltage Noise Density, } f = 10 \text{kHz} \\ \text{Input Foltage Range} \\ \text{Common-Mode Rejection Ratio} \\ \text{CMRR} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\ \text{VS} = 5.5 \text{V, } -0.1 \text{V } \times \text{VCM} < 5.6 \text{V} \\$	±1 mV μV/°C 150 μV/V μV/V
vs Temperature $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ ± 4 vs Power-Supply Rejection Ratio PSRR $V_S = 2.7\text{V}$ to 5.5V , $V_{CM} = 0\text{V}$ 40 $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ $V_S = 2.7\text{V}$ to 5.5V , $V_{CM} = 0\text{V}$ 40 Channel Separation (dual, quad) dc 0.15 INPUT BIAS CURRENT Input Bias Current Input Bias Current 0.5 0.5 vs Temperature See Typical Characterist Input Offset Current Input Voltage Noise, f = 100Hz to 400kHz 0.5 0.5 0.5 Input Voltage Noise, f = 100Hz to 400kHz 0.5 <td>μV/°C 150 μV/V 175 μV/V</td>	μV/°C 150 μV/V 175 μV/V
vs Power-Supply Rejection Ratio PSRR $V_S = 2.7V \text{ to } 5.5V, V_{CM} = 0V$ 40 47 T _A = -40°C to +85°C V _S = 2.7V to 5.5V, V _{CM} = 0V 40	150 μV/V 175 μV/V
T _A = -40°C to +85°C V _S = 2.7V to 5.5V, V _{CM} = 0V 1 Channel Separation (dual, quad) dc 0.15 INPUT BIAS CURRENT Input Bias Current Ipput Bias Current ±0.5 ±0.2 ±0.2 ±0.2 ±0.2 ±0.2 ±0.2<	175 μV/V
Channel Separation (dual, quad) dc 0.15 INPUT BIAS CURRENT Input Bias Current Input Bias Current Input Offset	
Channel Separation (dual, quad) dc 0.15 INPUT BIAS CURRENT Input Bias Current Input Bias Current ± 0.5 $\pm $	μV/V
Input Bias Current	
vs Temperature See Typical Characterist Input Offset Current ± 0.5 NOISE ± 0.5 Input Voltage Noise, f = 100Hz to 400kHz 4 Input Voltage Noise Density, f = 10kHz 6 Input Current Noise Density, f = 10kHz 7 Input Voltage Range 4 Current Noise Density, f = 10kHz 1 Input Voltage Range V _{CM} Common-Mode Voltage Range V _{CM} Common-Mode Rejection Ratio CMRR V _S = 2.7V, -0.1V < V _{CM} < 2.8V	W
Input Offset Current IOS ±0.5 ±0.0 ±0.2	±10 pA
Input Offset Current IOS ±0.5 ±0.0 ±0.2 ±0.4 ±0.2 ±0.4 ±0.2	tics
NOISE Input Voltage Noise, f = 100Hz to 400kHz 4 Input Voltage Noise Density, f = 10kHz en 7 Input Current Noise Density, f = 10kHz 5 Current Noise Density, f = 10kHz in 4 INPUT VOLTAGE RANGE 4 Common-Mode Voltage Range V _{CM} T _A = -40°C to +85°C -0.1 (V+ Common-Mode Rejection Ratio CMRR V _S = 2.7V, -0.1V < V _{CM} < 2.8V	±10 pA
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	ONT.
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	μVrm
Input Current Noise Density, f = 100kHz	nV/√H
Current Noise Density, $f = 10kHz$ i_{D} 4 $10PUT VOLTAGE RANGE Common-Mode Voltage Range V_{CM} V_{CM} V_{CM} = -40^{\circ}C to +85^{\circ}C -0.1 (V+COM) = -40^{\circ}C to +85^{\circ}C V_{CM} = -40^{\circ}C t$	nV/√H
INPUT VOLTAGE RANGE Common-Mode Voltage Range V _{CM} T _A = -40°C to +85°C -0.1 (V+ Common-Mode Rejection Ratio CMRR V _S = 2.7V, -0.1V < V _{CM} < 2.8V	fA/√H
$ \begin{array}{cccccccccccccccccccccccccccccccccccc$	Y
Common-Mode Rejection Ratio CMRR $V_S = 2.7V, -0.1V < V_{CM} < 2.8V$ 66 84 $V_S = 5.5V, -0.1V < V_{CM} < 5.6V$ 74 90 TA = -40°C to +85°C $V_S = 5.5V, -0.1V < V_{CM} < 5.6V$ 74 INPUT IMPEDANCE Differential $10^{13} \parallel 2.5$ Common-Mode $10^{13} \parallel 2.5$ OPEN-LOOP GAIN Open-Loop Voltage Gain AOL $R_L = 10k\Omega, 50mV < V_O < (V_+) -50mV$ 100 122 $R_L = 10k\Omega, 50mV < V_O < (V_+) -50mV$ 100 $R_L = 10k\Omega, 50mV < V_O < (V_+) -200mV$ 100 120 $T_A = -40$ °C to +85°C $R_L = 1k\Omega, 200mV < V_O < (V_+) -200mV$ 100 120	+) + 0.1 V
$V_{S} = 5.5 \text{V}, -0.1 \text{V} < \text{V}_{CM} < 5.6 \text{V} \qquad 74 \qquad 90$ $V_{S} = 5.5 \text{V}, -0.1 \text{V} < \text{V}_{CM} < 5.6 \text{V} \qquad 74$ INPUT IMPEDANCE Differential $Common\text{-Mode} \qquad 10^{13} \parallel 2.5$ $Common\text{-Mode} \qquad 10^{13} \parallel 6.5$ OPEN-LOOP GAIN $Copen\text{-Loop Voltage Gain} \qquad A_{OL} \qquad R_{L} = 10 \text{k}\Omega, 50 \text{mV} < \text{V}_{O} < (\text{V+}) - 50 \text{mV} \qquad 100$ $R_{L} = 10 \text{k}\Omega, 50 \text{mV} < \text{V}_{O} < (\text{V+}) - 50 \text{mV} \qquad 100$ $R_{L} = 1 \text{k}\Omega, 200 \text{mV} < \text{V}_{O} < (\text{V+}) - 200 \text{mV} \qquad 100$ $R_{L} = 1 \text{k}\Omega, 200 \text{mV} < \text{V}_{O} < (\text{V+}) - 200 \text{mV} \qquad 100$	dB
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Common-Mode 10^{13} 6.5 OPEN-LOOP GAIN AOL Open-Loop Voltage Gain AOL $T_A = -40^{\circ}C$ to +85°C $R_L = 10k\Omega$, $50mV < V_O < (V+) -50mV$ $R_L = 10k\Omega$, $50mV < V_O < (V+) -50mV$ 100 $R_L = 10k\Omega$, $200mV < V_O < (V+) -200mV$ 100 $R_L = 1k\Omega$, $200mV < V_O < (V+) -200mV$ 100 $R_L = 1k\Omega$, $200mV < V_O < (V+) -200mV$ 100	· CO
Common-Mode 10^{13} 6.5 OPEN-LOOP GAIN AOL R _L = 10kΩ, 50mV < V _O < (V+) -50mV 100 122 T _A = -40°C to +85°C R _L = 10kΩ, 50mV < V _O < (V+) -50mV 100 100 R _L = 1kΩ, 200mV < V _O < (V+) -200mV 100 120 R _L = 1kΩ, 200mV < V _O < (V+) -200mV 100 100	$\Omega \parallel p$
OPEN-LOOP GAIN Open-Loop Voltage Gain A_{OL} $R_{L} = 10k\Omega$, $50mV < V_{O} < (V+) -50mV$ 100 122 $T_{A} = -40^{\circ}C$ to +85°C $R_{L} = 10k\Omega$, $50mV < V_{O} < (V+) -50mV$ 100 120 $R_{L} = 1k\Omega$, $200mV < V_{O} < (V+) -200mV$ 100 120 $R_{L} = 1k\Omega$, $200mV < V_{O} < (V+) -200mV$ 100 100	$\Omega \parallel p$
$ \begin{array}{llllllllllllllllllllllllllllllllllll$	4.
$ \begin{array}{llllllllllllllllllllllllllllllllllll$	dB
$R_L = 1kΩ, 200mV < V_O < (V+) -200mV$ 100 120 $R_L = 1kΩ, 200mV < V_O < (V+) -200mV$ 100 100	dB
$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C}$ $R_L = 1\text{k}\Omega, 200\text{mV} < \text{V}_O < (\text{V+}) -200\text{mV}$ 100	dB
	dB
I TENOLINO I TEOLONOE	N 1 10
Gain-Bandwidth Product GBW G = 1 38	MHz
Slew Rate	V/µs
Settling Time: 0.1% $G = \pm 1$, 2V Step 0.22	μs
0.01% $G = \pm 1, 2V \text{ Step}$ 0.5	μs
Overload Recovery Time $V_{IN} \bullet G = V_S$ 0.1	μs
Total Harmonic Distortion + Noise THD+N $R_L = 600\Omega$, $V_O = 2.5V_{PP}(2)$, $G = 1$, $f = 1$ kHz 0.0006	%
Differential Gain Error $G = 2$, $R_L = 600\Omega$, $V_O = 1.4V(3)$ 0.17	%
Differential Phase Error $G = 2$, $R_L = 600\Omega$, $V_O = 1.4V(3)$ 0.17	deg

 $⁽¹⁾ V_S = +5V.$

⁽²⁾ $V_{OUT} = 0.25V$ to 2.75V.

⁽³⁾ NTSC signal generator used. See Figure 6 for test circuit.

⁽⁴⁾ Output voltage swings are measured between the output and power supply rails.

⁽⁵⁾ See typical characteristic curve, Output Voltage Swing vs Output Current.



ELECTRICAL CHARACTERISTICS: $V_S = 2.7V$ to 5.5V (continued)

Boldface limits apply over the temperature range, $T_A = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$. $V_S = 5\text{V}$.

All specifications at $T_A = +25^{\circ}C$, $R_I = 1k\Omega$ connected to $V_S/2$ and $V_{OUT} = V_S/2$, unless otherwise noted.

V.1001. CONT.			OPA350	OPA350, OPA2350, OPA4350				
PARAMETER	WW	TEST CONDITIONS	MIN	TYP(1)	MAX	UNIT		
OUTPUT	WWW	ON CONTRACT	N. N. YO	M.Co.	TW			
Voltage Output Swing from Rail ⁽⁴⁾	VOUT	$R_L = 10k\Omega$, $A_{OL} \ge 100dB$	MW.Iu	10	50	mV		
$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	MAL	$R_L = 10k\Omega$, $A_{OL} \ge 100dB$	W TAN J	00 λ	50	mV		
	WWW	$R_L = 1k\Omega$, $A_{OL} \ge 100dB$		25	200	mV		
$T_A = -40^{\circ}C \text{ to } +85^{\circ}C$	-131	$R_L = 1k\Omega$, $A_{OL} \ge 100dB$	THE WAY	V.C	200	√ mV		
Output Current	IOUT		N	±40(5)	OM.	mA		
Short-Circuit Current	Isc		MM	±80		mA		
Capacitive Load Drive	C _{LOAD}		See Ty	pical Charac	teristics	W		
POWER SUPPLY		ZIW.Inv		M. Inc	41 COM	-XX		
Operating Voltage Range	V _S	$T_A = -40^{\circ}C$ to $+85^{\circ}C$	2.7	100	5.5	V		
Minimum Operating Voltage	W.		V	2.5	ON.CO.	V		
Quiescent Current (per amplifier)	IQ	I _O = 0		5.2	7.5	mA		
$T_A = -40^{\circ}C$ to $+85^{\circ}C$	TW	I _O = 0			8.5	mA		
TEMPERATURE RANGE	WTS	WW TOOY.CO TIT		M. M.	1007.0			
Specified Range	VI.		-40	WWW	+85	°C		
Operating Range	Mil		-55		+150	°C		
Storage Range	WILL		-55		+150	°C		
Thermal Resistance	$\theta_{\sf JA}$		WT.	WW	100	Y.Co.		
MSOP-8 Surface Mount	OM.I		M	150	MM.To	°C/W		
SO-8 Surface Mount	MITW		TITY	150	TW 10	°C/W		
DIP-8	CON		WILL	100	11	°C/W		
SO-14 Surface Mount	COM.		OM.	100	MW.	°C/W		
SSOP-16 Surface Mount	Y. OM.T.		COM.	100	TATE OF THE PARTY	°C/W		

⁽¹⁾ $V_S = +5V$.

⁽²⁾ $V_{OUT} = 0.25V$ to 2.75V.

⁽³⁾ NTSC signal generator used. See Figure 6 for test circuit.

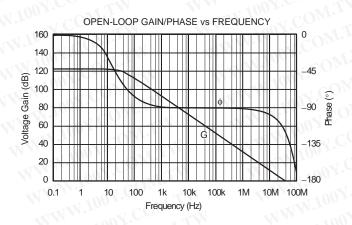
⁽⁴⁾ Output voltage swings are measured between the output and power supply rails.

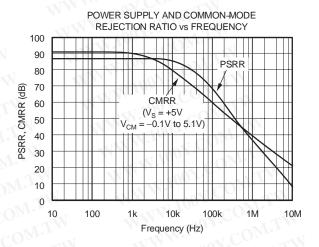
⁽⁵⁾ See typical characteristic curve, Output Voltage Swing vs Output Current.

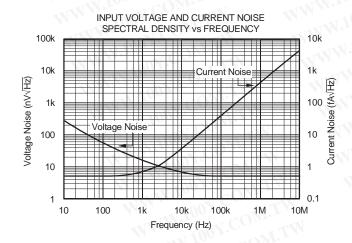


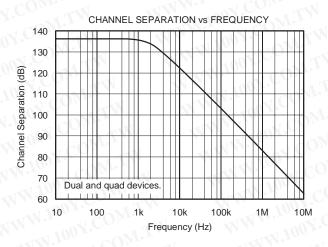
TYPICAL CHARACTERISTICS

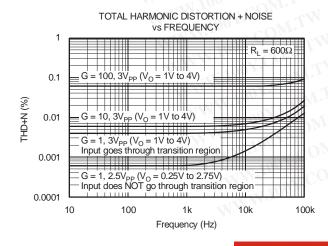
All specifications at $T_A = +25$ °C, $V_S = +5V$, and $R_L = 1k\Omega$ connected to $V_S/2$, unless otherwise noted.

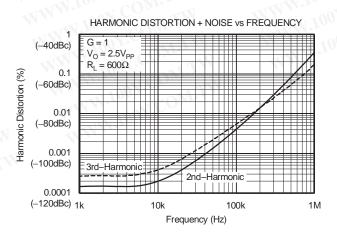












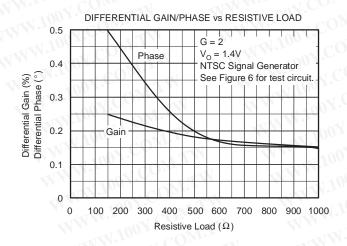
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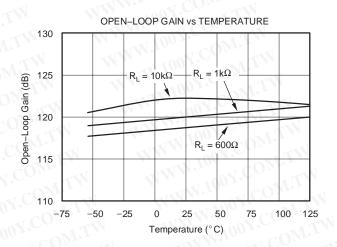
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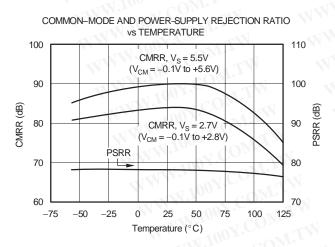


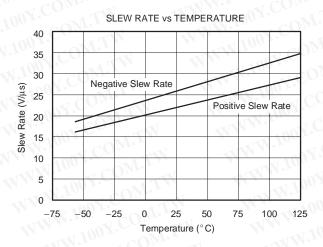
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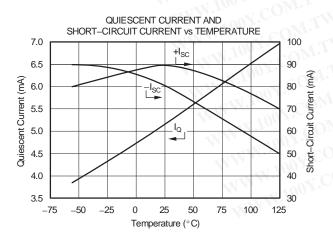
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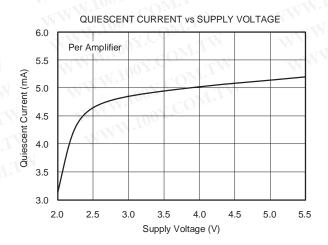










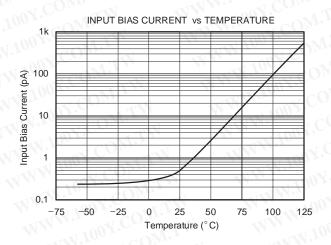


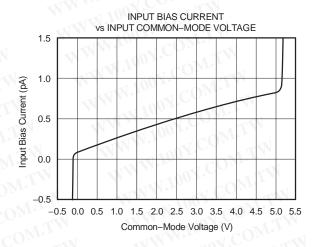
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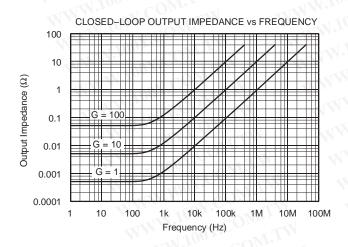


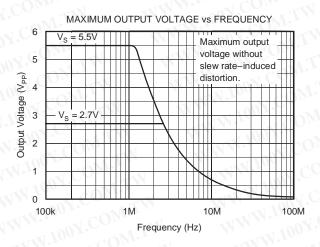
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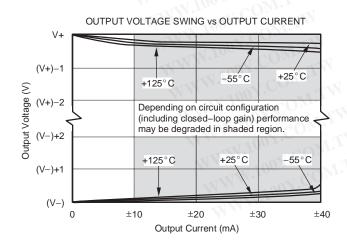
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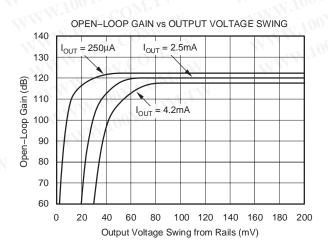






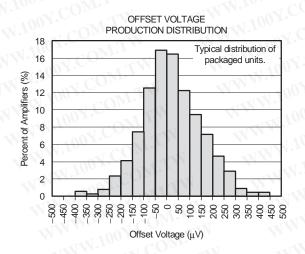


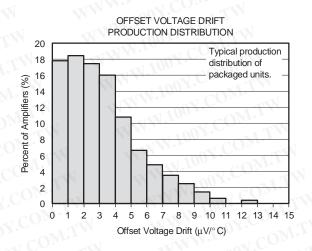


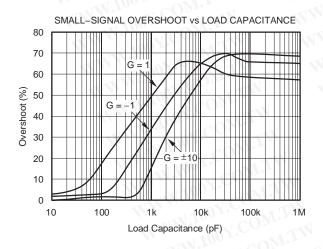


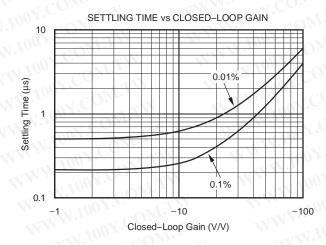
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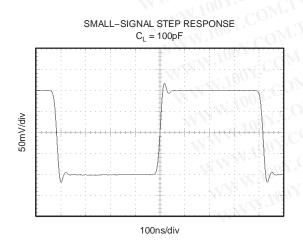
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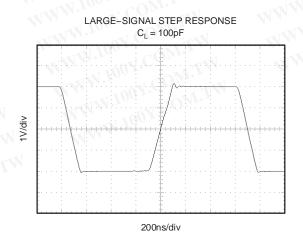














APPLICATIONS INFORMATION

OPA350 series op amps are fabricated on a state-of-the-art 0.6 micron CMOS process. They are unity-gain stable and suitable for a wide range of general-purpose applications. Rail-to-rail input/output make them ideal for driving sampling A/D converters. They are also well-suited for controlling the output power in cell phones. These applications often require high speed and low noise. In addition, the OPA350 series offers a low-cost solution for general-purpose and consumer video applications (75 Ω drive capability).

Excellent ac performance makes the OPA350 series well-suited for audio applications. Their bandwidth, slew rate, low noise (5nV/ $\sqrt{\text{Hz}}$), low THD (0.0006%), and small package options are ideal for these applications. The class AB output stage is capable of driving 600Ω loads connected to any point between V+ and ground.

Rail-to-rail input and output swing significantly increases dynamic range, especially in low voltage supply applications. Figure 1 shows the input and output waveforms for the OPA350 in unity-gain configuration. Operation is from a single +5V supply with a 1k Ω load connected to Vg/2. The input is a 5Vpp sinusoid. Output voltage swing is approximately 4.95Vpp.

Power supply pins should be bypassed with $0.01\mu F$ ceramic capacitors.

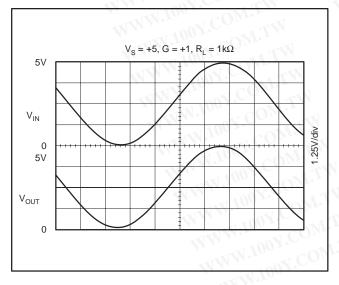


Figure 1. Rail-to-Rail Input and Output

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OPERATING VOLTAGE

OPA350 series op amps are fully specified from +2.7V to +5.5V. However, supply voltage may range from +2.5V to +5.5V. Parameters are tested over the specified supply range—a unique feature of the OPA350 series. In addition, many specifications apply from -40°C to +85°C. Most behavior remains virtually unchanged throughout the full operating voltage range. Parameters that vary significantly with operating voltage or temperature are shown in the typical characteristics.

RAIL-TO-RAIL INPUT

The tested input common-mode voltage range of the OPA350 series extends 100mV beyond the supply rails. This is achieved with a complementary input stage—an N-channel input differential pair in parallel with a P-channel differential pair, as shown in Figure 2. The N-channel pair is active for input voltages close to the positive rail, typically (V+) - 1.8V to 100mV above the positive supply, while the P-channel pair is on for inputs from 100mV below the negative supply approximately (V+) - 1.8V. There is a small transition region, typically (V+) - 2V to (V+) - 1.6V, in which both pairs are on. This 400mV transition region can vary ±400mV with process variation. Thus, the transition region (both input stages on) can range from (V+) -2.4V to (V+) - 2.0V on the low end, up to (V+) - 1.6Vto (V+) - 1.2V on the high end.

OPA350 series op amps are laser-trimmed to reduce offset voltage difference between the N-channel and P-channel input stages, resulting in improved common-mode rejection and a smooth transition between the N-channel pair and the P-channel pair. However, within the 400mV transition region PSRR, CMRR, offset voltage, offset drift, and THD may be degraded compared to operation outside this region.

A double-folded cascode adds the signal from the two input pairs and presents a differential signal to the class AB output stage. Normally, input bias current is approximately 500fA. However, large inputs (greater than 300mV beyond the supply rails) can turn on the OPA350's input protection diodes, causing excessive current to flow in or out of the input pins. Momentary voltages greater than 300mV beyond the power supply can be tolerated if the current on the input pins is limited to 10mA. This is easily accomplished with an input resistor, as shown in Figure 3. Many input signals are inherently current-limited to less than 10mA; therefore, a limiting resistor is not required.



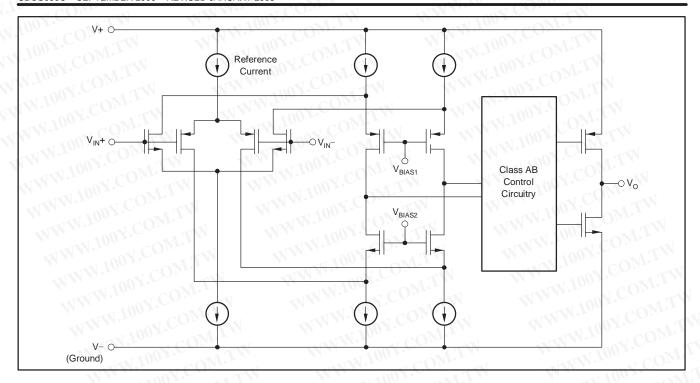


Figure 2. Simplified Schematic

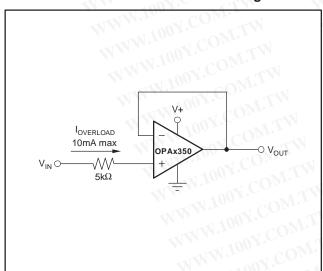


Figure 3. Input Current Protection for Voltages
Exceeding the Supply Voltage

RAIL-TO-RAIL OUTPUT

A class AB output stage with common-source transistors is used to achieve rail-to-rail output. For light resistive loads (>10k Ω), the output voltage swing is typically ten millivolts from the supply rails. With heavier resistive loads (600 Ω to 10k Ω), the output can swing to

within a few tens of millivolts from the supply rails and maintain high open-loop gain. See the typical characteristics *Output Voltage Swing vs Output Current* and *Open-Loop Gain vs Output Voltage*.

CAPACITIVE LOAD AND STABILITY

OPA350 series op amps can drive a wide range of capacitive loads. However, all op amps under certain conditions may become unstable. Op amp configuration, gain, and load value are just a few of the factors to consider when determining stability. An op amp in unity-gain configuration is the most susceptible to the effects of capacitive load. The capacitive load reacts with the op amp's output impedance, along with any additional load resistance, to create a pole in the small-signal response that degrades the phase margin.

In unity gain, OPA350 series op amps perform well with very large capacitive loads. Increasing gain enhances the amplifier's ability to drive more capacitance. The typical characteristic *Small-Signal Overshoot vs Capacitive Load* shows performance with a $1k\Omega$ resistive load. Increasing load resistance improves capacitive load drive capability.

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FEEDBACK CAPACITOR IMPROVES RESPONSE

For optimum settling time and stability with high-impedance feedback networks, it may be necessary to add a feedback capacitor across the feedback resistor, R_F, as shown in Figure 4. This capacitor compensates for the zero created by the feedback network impedance and the OPA350's input capacitance (and any parasitic layout capacitance). The effect becomes more significant with higher impedance networks.

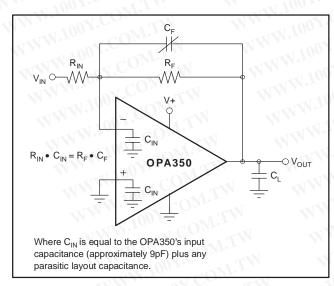


Figure 4. Feedback Capacitor Improves Dynamic Performance

It is suggested that a variable capacitor be used for the feedback capacitor since input capacitance may vary between op amps and layout capacitance is difficult to determine. For the circuit shown in Figure 4, the value of the variable feedback capacitor should be chosen so that the input resistance times the input capacitance of the OPA350 (typically 9pF) plus the estimated parasitic layout capacitance equals the feedback capacitor times the feedback resistor:

$$R_{IN} \cdot C_{IN} = R_F \cdot C_F$$

where C_{IN} is equal to the OPA350's input capacitance (sum of differential and common-mode) plus the layout capacitance. The capacitor can be varied until optimum performance is obtained.

DRIVING A/D CONVERTERS

OPA350 series op amps are optimized for driving medium speed (up to 500kHz) sampling A/D converters. However, they also offer excellent performance for higher speed converters. The OPA350

series provides an effective means of buffering the A/D's input capacitance and resulting charge injection while providing signal gain.

Figure 5 shows the OPA350 driving an ADS7861. The ADS7861 is a dual, 500kHz, 12-bit sampling converter in the tiny SSOP-24 package. When used with the miniature package options of the OPA350 series, the combination is ideal for space-limited applications. For further information, consult the ADS7861 data sheet (SBAS110A).

OUTPUT IMPEDANCE

The low frequency open-loop output impedance of the OPA350's common-source output stage is approximately $1k\Omega$. When the op amp is connected with feedback, this value is reduced significantly by the loop gain of the op amp. For example, with 122dB of open-loop gain, the output impedance is reduced in unity-gain to less than 0.001Ω . For each decade rise in the closed-loop gain, the loop gain is reduced by the same amount which results in a ten-fold increase in effective output impedance (see the typical characteristic, *Output Impedance vs Frequency*).

At higher frequencies, the output impedance will rise as the open-loop gain of the op amp drops. However, at these frequencies the output also becomes capacitive due to parasitic capacitance. This prevents the output impedance from becoming too high, which can cause stability problems when driving capacitive loads. As mentioned previously, the OPA350 has excellent capacitive load drive capability for an op amp with its bandwidth.

VIDEO LINE DRIVER

Figure 6 shows a circuit for a single supply, G=2 composite video line driver. The synchronized outputs of a composite video line driver extend below ground. As shown, the input to the op amp should be ac-coupled and shifted positively to provide adequate signal swing to account for these negative signals in a single-supply configuration.

The input is terminated with a 75Ω resistor and ac-coupled with a $47\mu F$ capacitor to a voltage divider that provides the dc bias point to the input. In Figure 6, this point is approximately (V–) + 1.7V. Setting the optimal bias point requires some understanding of the nature of composite video signals. For best performance, one should be careful to avoid the distortion caused by the transition region of the OPA350's complementary input stage. Refer to the discussion of rail-to-rail input.



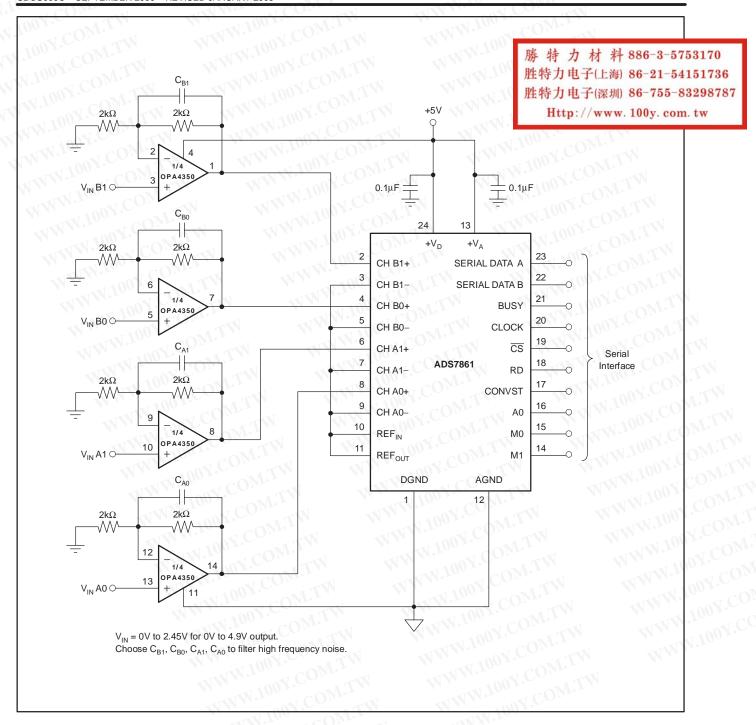


Figure 5. OPA4350 Driving Sampling A/D Converter



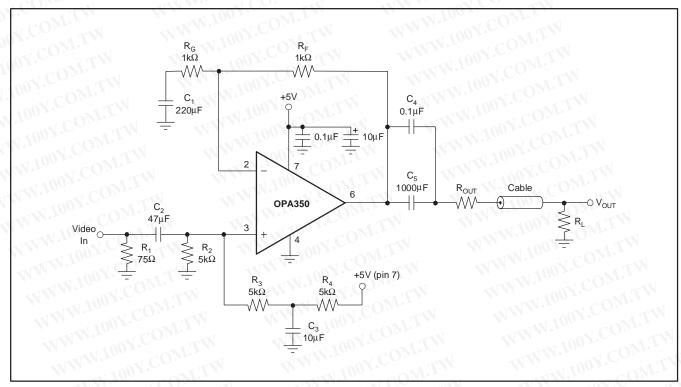


Figure 6. Single-Supply Video Line Driver

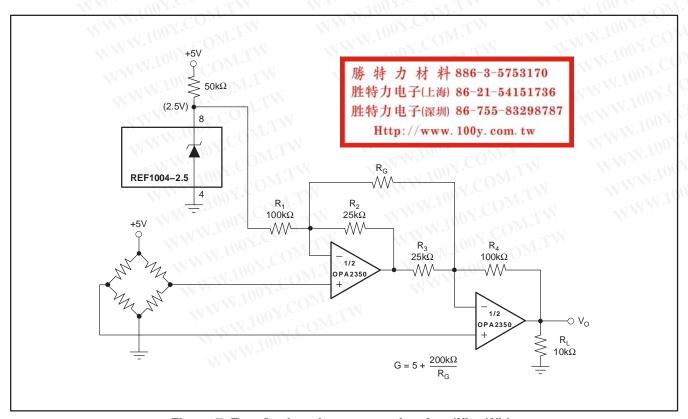
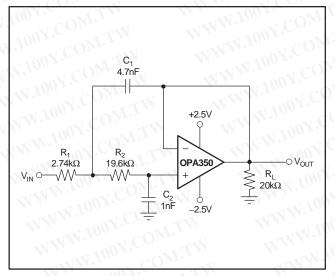
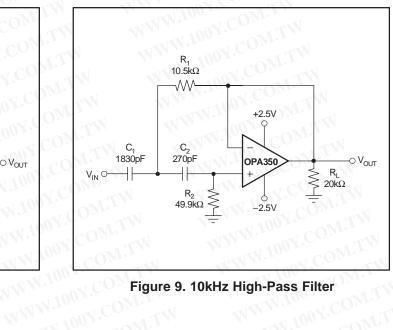


Figure 7. Two Op-Amp Instrumentation Amplifier With Improved High Frequency Common-Mode Rejection







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Figure 8. 10kHz Low-Pass Filter

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WWW.100Y.COM.TW Figure 9. 10kHz High-Pass Filter WWW.100Y.COM

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PACKAGE OPTION ADDENDUM

10-Dec-2007

Orderable Device	Status (1)	Package	Package	Pins	Packag	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp (3)
OPA2350EA/250	ACTIVE	Type MSOP	Drawing DGK	8	Qty 250	Green (RoHS &	CU NIPDAU	Level-2-260C-1 YEAR
100Y.Com.TVI		100	Moore	TW		no Sb/Br)	OM.	
OPA2350EA/250G4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350EA/2K5	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350EA/2K5G4	ACTIVE	MSOP	DGK	0 8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350PA	ACTIVE	PDIP	P 7.	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
OPA2350PAG4	ACTIVE	PDIP	Pooy	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
OPA2350UA	ACTIVE	SOIC	D100	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350UA/2K5	ACTIVE	SOIC	D . 10	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350UA/2K5G4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA2350UAG4	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350EA/250	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350EA/250G4	ACTIVE	MSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350EA/2K5	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350EA/2K5G4	ACTIVE	MSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350PA	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
OPA350PAG4	ACTIVE	PDIP	Р	8	50	Green (RoHS & no Sb/Br)	CU NIPDAU	N / A for Pkg Type
OPA350UA	ACTIVE	SOIC	D	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350UA/2K5	ACTIVE	SOIC	T.D.	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350UA/2K5G4	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA350UAG4	ACTIVE	SOIC	OND	8	100	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350EA/250	ACTIVE	SSOP/ QSOP	DBQ	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350EA/250G4	ACTIVE	SSOP/ QSOP	DBQ	16	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350EA/2K5	ACTIVE	SSOP/ QSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350EA/2K5G4	ACTIVE	SSOP/ QSOP	DBQ	16	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350UA	ACTIVE	SOIC	D	14	58	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR



PACKAGE OPTION ADDENDUM

10-Dec-2007

Orderable Device	Status (1)	Package Type	Package Drawing	Pins Packag Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp (3)
OPA4350UA/2K5	ACTIVE	SOIC	CODY	14 2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350UA/2K5G4	ACTIVE	SOIC	V.CD	14 2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
OPA4350UAG4	ACTIVE	SOIC	D D	14 58	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

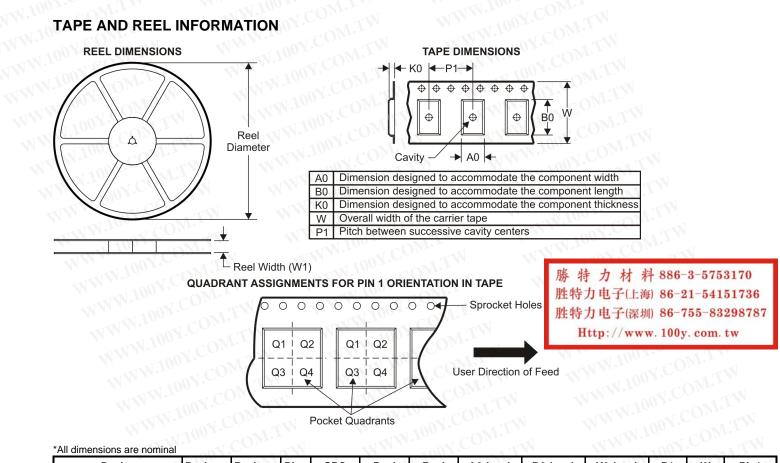
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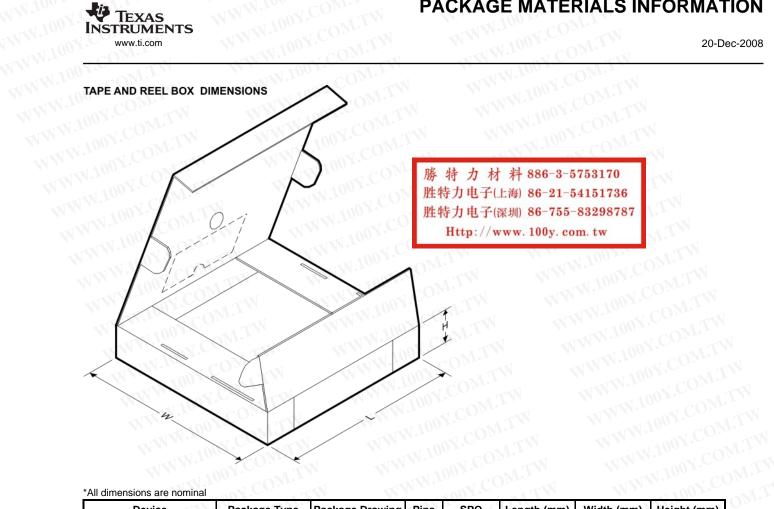
20-Dec-2008



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadran
DPA2350EA/250	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
PA2350EA/2K5	MSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
PA2350UA/2K5	SOIC	D.C	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA350EA/250	MSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA350EA/2K5	MSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA350UA/2K5	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
DPA4350EA/250	SSOP/ QSOP	DBQ	16	250	180.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
DPA4350EA/2K5	SSOP/ QSOP	DBQ	16	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
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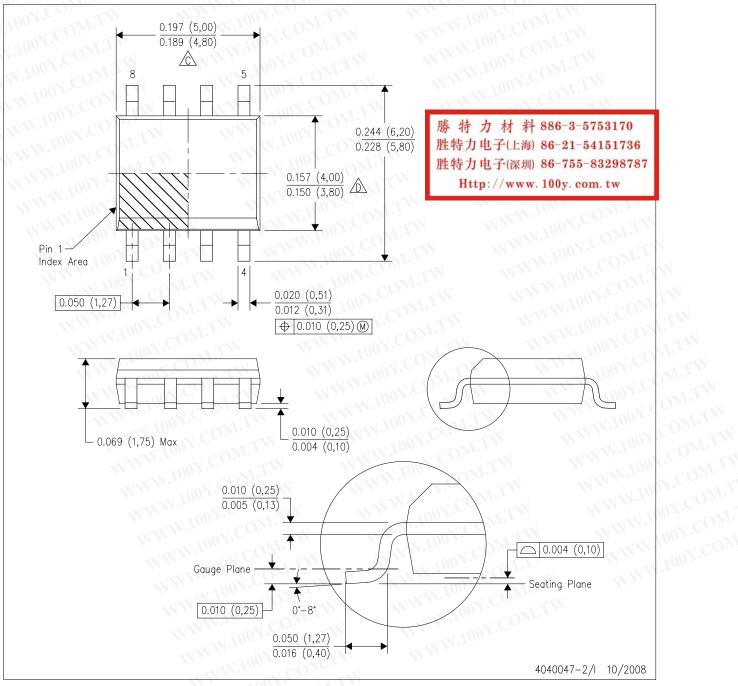


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DPA2350EA/250	MSOP	DGK	8	250	190.5	212.7	31.8
PA2350EA/2K5	MSOP	DGK	8	2500	346.0	346.0	29.0
PA2350UA/2K5	SOIC	D	8	2500	346.0	346.0	29.0
OPA350EA/250	MSOP	DGK	8	250	190.5	212.7	31.8
OPA350EA/2K5	MSOP	DGK	8	2500	346.0	346.0	29.0
OPA350UA/2K5	SOIC	D	8	2500	346.0	346.0	29.0
DPA4350EA/250	SSOP/QSOP	DBQ	16	250	190.5	212.7	31.8
PA4350EA/2K5	SSOP/QSOP	DBQ	16	2500	346.0	346.0	29.0
PA4350UA/2K5	SOIC	D	14	2500	346.0	346.0	33.0

D (R-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE

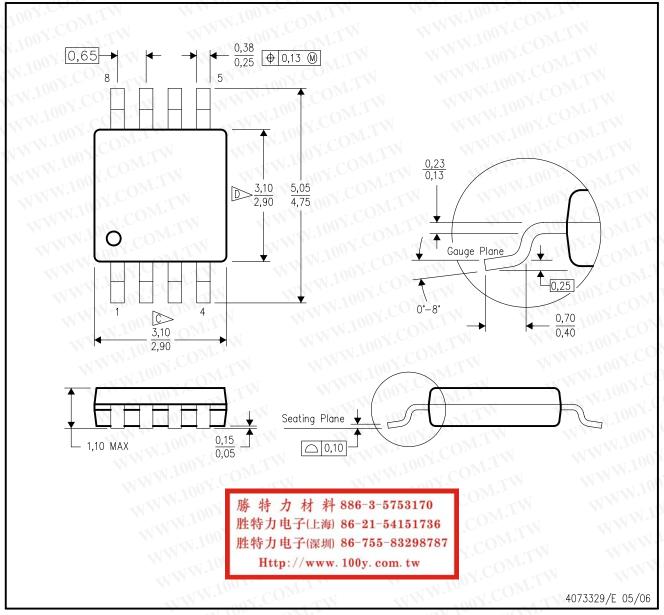


- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
- Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
- E. Reference JEDEC MS-012 variation AA.



DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE

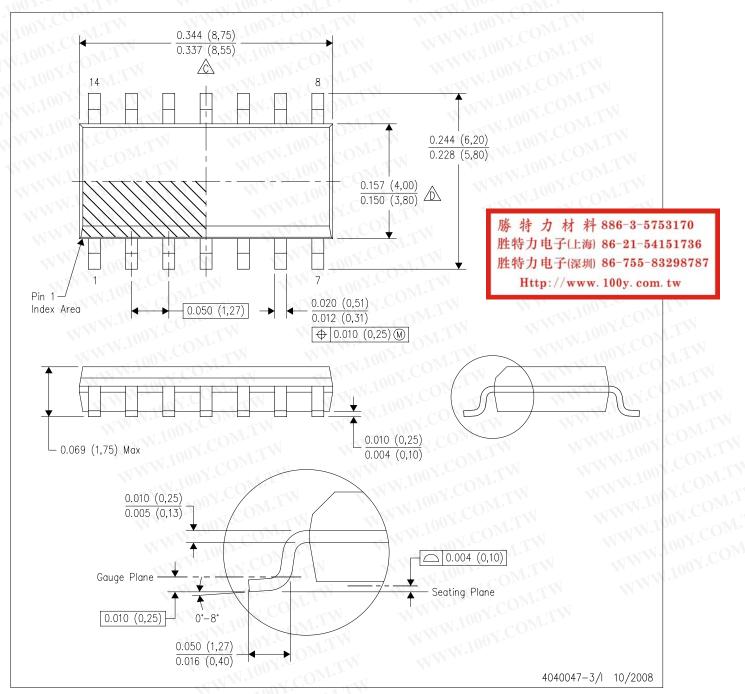


- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
- Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
- E. Falls within JEDEC MO-187 variation AA, except interlead flash.



D (R-PDSO-G14)

PLASTIC SMALL-OUTLINE PACKAGE



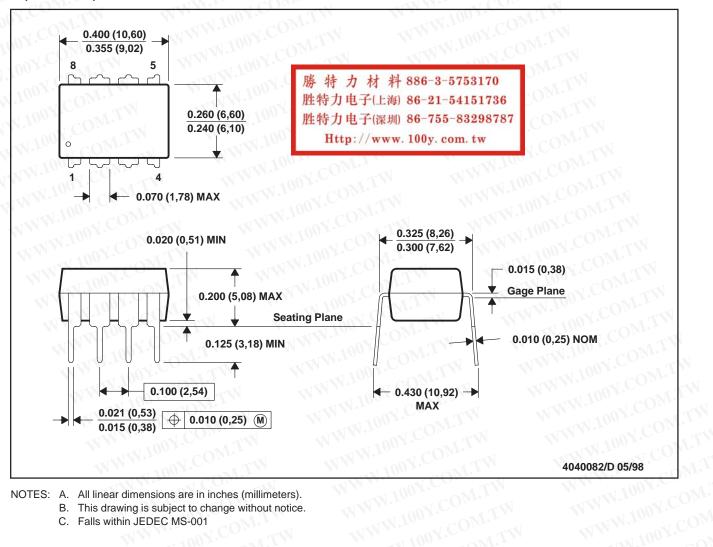
- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 (0,15) per end.
- Body width does not include interlead flash. Interlead flash shall not exceed .017 (0,43) per side.
- E. Reference JEDEC MS-012 variation AB.



P (R-PDIP-T8)

VWW.100Y.COM.TW

PLASTIC DUAL-IN-LINE



NOTES: A. All linear dimensions are in inches (millimeters).

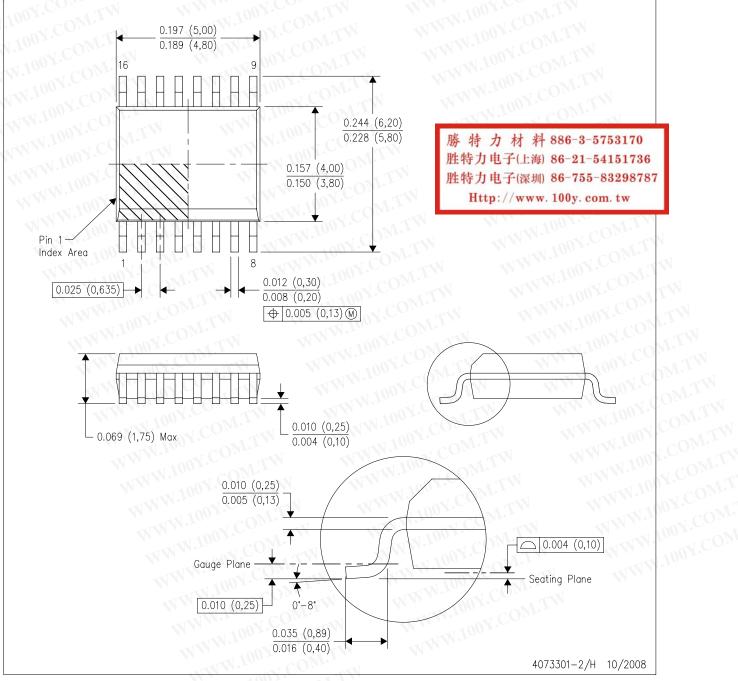
- B. This drawing is subject to change without notice.
- C. Falls within JEDEC MS-001 WWW.100Y.C

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DBQ (R-PDSO-G16)

PLASTIC SMALL-OUTLINE PACKAGE



- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- C. Body dimensions do not include mold flash or protrusion not to exceed 0.006 (0,15) per side.
- D. Falls within JEDEC MO-137 variation AB.



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