IRS2153(1)D(S)PbF

SELF-OSCILLATING HALF-BRIDGE DRIVER IC

Features

- Integrated 600 V half-bridge gate driver
- C_T, R_T programmable oscillator
- 15.4 V Zener clamp on V_{CC}
- Micropower startup
- Non-latched shutdown on C_T pin (1/6th V_{CC})
- Internal bootstrap FET
- Excellent latch immunity on all inputs and outputs
- +/- 50 V/ns dV/dt immunity
- ESD protection on all pins
- 8-lead SOIC or PDIP package
- Internal deadtime

Product Summary

V_{OFFSET}	600 V Max
Duty cycle	50%
Driver source/sink current	180 mA/260 mA typ.
V_{clamp}	15.4 V typ.
Deadtime	1.1 µs typ. (IRS2153D) 0.6 µs typ. (IRS21531D)

Description

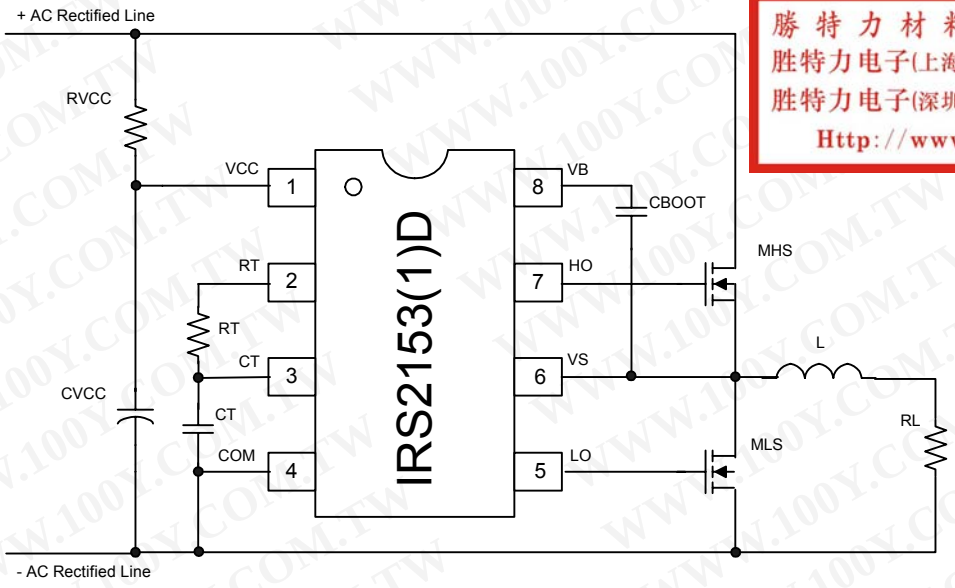
The IRS2153(1)D is based on the popular IR2153 self-oscillating half-bridge gate driver IC using a more advanced silicon platform, and incorporates a high voltage half-bridge gate driver with a front end oscillator similar to the industry standard CMOS 555 timer. HVIC and latch immune CMOS technologies enable rugged monolithic construction. The output driver features a high pulse current buffer stage designed for minimum driver cross-conduction. Noise immunity is achieved with low di/dt peak of the gate drivers.

Package

PDIP8
IRS2153(1)DPbF

SO8
IRS2153(1)DSPbF

Typical Connection Diagram



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Absolute Maximum Ratings

Absolute maximum ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM, all currents are defined positive into any lead. The thermal resistance and power dissipation ratings are measured under board mounted and still air conditions.

Parameter				
Symbol	**Definition**	**Min.**	**Max.**	**Units**
V_B	High side floating supply voltage	-0.3	625	V
V_S	High side floating supply offset voltage	V_B - 25	V_B + 0.3	
V_{HO}	High side floating output voltage	V_S – 0.3	V_B + 0.3	
V_{LO}	Low side output voltage	-0.3	V_{CC} + 0.3	
I_{RT}	R_T pin current	-5	5	mA
V_{RT}	R_T pin voltage	-0.3	V_{CC} + 0.3	V
V_{CT}	C_T pin voltage	-0.3	V_{CC} + 0.3	
I_{CC}	Supply current (Note 1)	---	20	mA
I_{OMAX}	Maximum allowable current at LO and HO due to external power transistor Miller effect.	-500	500	
dV_S/dt	Allowable offset voltage slew rate	-50	50	V/ns
P_D	Maximum power dissipation @ $T_A \leq +25$ ºC, 8-Pin DIP	---	1.0	W
P_D	Maximum power dissipation @ $T_A \leq +25$ ºC, 8-Pin SOIC	---	0.625	
R_{thJA}	Thermal resistance, junction to ambient, 8-Pin DIP	---	85	ºC/W
R_{thJA}	Thermal resistance, junction to ambient, 8-Pin SOIC	---	128	
T_J	Junction temperature	-55	150	ºC
T_S	Storage temperature	-55	150	
T_L	Lead temperature (soldering, 10 seconds)	---	300	

Note 1: This IC contains a zener clamp structure between the chip V_{CC} and COM which has a nominal breakdown voltage of 15.4 V. Please note that this supply pin should not be driven by a DC, low impedance power source greater than the V_{CLAMP} specified in the Electrical Characteristics section.

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Recommended Operating Conditions

For proper operation the device should be used within the recommended conditions.

Parameter		Min.	Max.	Units
Symbol	Definition			
V_{BS}	High side floating supply voltage	V_{CC} - 0.7	V_{CLAMP}	V
V_S	Steady state side floating supply offset voltage	-3.0 (Note 2)	600	
V_{CC}	Supply voltage	V_{CCUV+} +0.1 V	$V_{CC\ CLAMP}$	
I_{CC}	Supply current	(Note 3)	5	mA
T_J	Junction temperature	-40	125	ºC

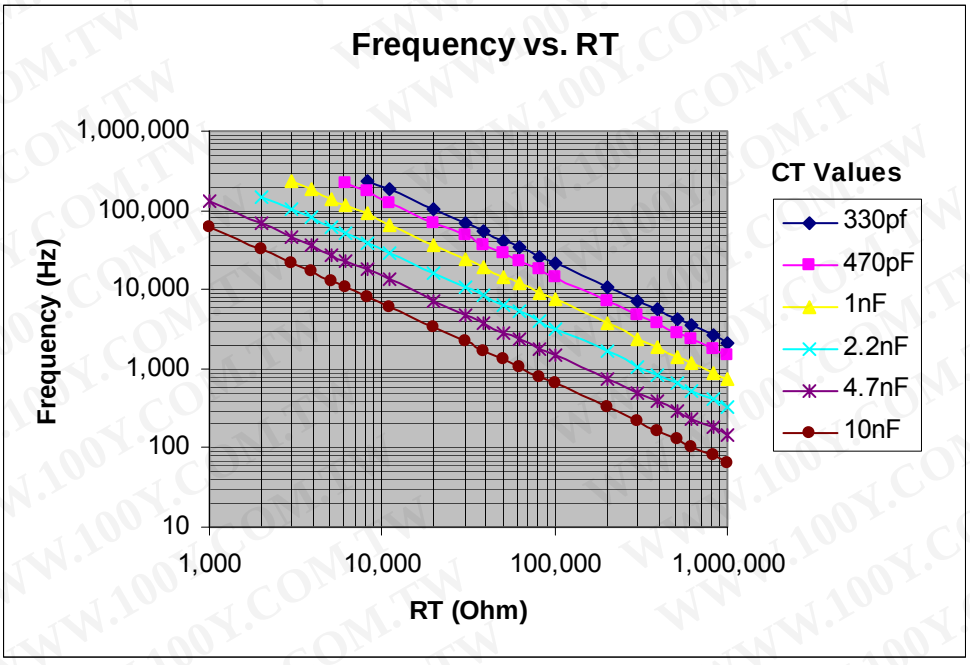
Note 2: It is recommended to avoid output switching conditions where the negative-going spikes at the V_S node would decrease V_S below ground by more than -5 V.

Note 3: Enough current should be supplied to the V_{CC} pin of the IC to keep the internal 15.6 V zener diode clamping the voltage at this pin.

Recommended Component Values

Parameter		Min.	Max.	Units
Symbol	Component			
R_T	Timing resistor value	1	---	kΩ
C_T	C_T pin capacitor value	330	---	pF

V_{BIAS} (V_{CC}, V_{BS}) = 14 V, V_S=0 V and T_A = 25 °C, CLO = CHO = 1 nF.



For further information, see Fig. 12.

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Electrical Characteristics

V_{BIAS} (V_{CC}, V_{BS}) = 14 V, C_T = 1 nF, V_S=0 V and T_A = 25 °C unless otherwise specified. The output voltage and current (V_O and I_O) parameters are referenced to COM and are applicable to the respective output leads: HO or LO. CLO = CHO = 1 nF.

Symbol	Definition	Min	Typ	Max	Units	Test Conditions
Low Voltage Supply Characteristics						
V_{CCUV+}	Rising V_{CC} undervoltage lockout threshold	10.0	11.0	12.0	V	
V_{CCUV-}	Falling V_{CC} undervoltage lockout threshold	8.0	9.0	10.0		
$V_{CCUVHYS}$	V_{CC} undervoltage lockout hysteresis	1.6	2.0	2.4		
I_{QCCUV}	Micropower startup V_{CC} supply current	---	130	170	µA	$V_{CC} \le V_{CCUV-}$
I_{QCC}	Quiescent V_{CC} supply current	---	800	1000		
I_{CC}	V_{CC} supply current	---	1.8	---	mA	R_T = 36.9 kΩ
$V_{CC\ CLAMP}$	V_{CC} zener clamp voltage	14.4	15.4	16.8	V	I_{CC} = 5 mA
Floating Supply Characteristics						
I_{QBS}	Quiescent V_{BS} supply current	---	60	80	µA	
V_{BSUV+}	V_{BS} supply undervoltage positive going threshold	8.0	9.0	9.5	V	
V_{BSUV-}	V_{BS} supply undervoltage negative going threshold	7.0	8.0	9.0		
I_{LK}	Offset supply leakage current	---	---	50	µA	V_B = V_S = 600 V
Oscillator I/O Characteristics						
f_{OSC}	Oscillator frequency	18.4	19.0	19.6	kHz	R_T = 36.5 kΩ
		88	93	100		R_T = 7.15 kΩ
d	R_T pin duty cycle	---	50	---	%	f_o < 100 kHz
I_{CT}	C_T pin current	---	0.02	1.0	µA	
I_{CTUV}	UV-mode C_T pin pulldown current	0.20	0.30	0.6	mA	V_{CC} = 7 V
V_{CT+}	Upper C_T ramp voltage threshold	---	9.32	---	V	
V_{CT-}	Lower C_T ramp voltage threshold	---	4.66	---		
V_{CTSD}	C_T voltage shutdown threshold	2.2	2.3	2.4		
V_{RT+}	High-level R_T output voltage, V_{CC} - V_{RT}	---	10	50	mV	I_{RT} = -100 µA
		---	100	300		I_{RT} = -1 mA
V_{RT-}	Low-level R_T output voltage	---	10	50		I_{RT} = 100 µA
		---	100	300		I_{RT} = 1 mA
V_{RTUV}	UV-mode R_T output voltage	---	0	100		$V_{CC} \le V_{CCUV-}$
V_{RTSD}	SD-mode R_T output voltage, V_{CC} - V_{RT}	---	10	50		I_{RT} = -100 µA, V_{CT} = 0 V
		---	100	300		I_{RT} = -1 mA, V_{CT} = 0 V

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Electrical Characteristics

V_{BIAS} (V_{CC}, V_{BS}) = 14 V, C_T = 1 nF, V_S=0 V and T_A = 25 °C unless otherwise specified. The output voltage and current (V_O and I_O) parameters are referenced to COM and are applicable to the respective output leads: HO or LO. CLO = CHO = 1 nF.

Symbol	Definition	Min	Typ	Max	Units	Test Conditions
Gate Driver Output Characteristics						
V_{OH}	High-level output voltage	---	V_{CC}	---	V	I_O = 0 A
V_{OL}	Low-level output voltage	---	COM	---		
V_{OL_UV}	UV-mode output voltage	---	COM	---		I_O = 0 A, $V_{CC} \le V_{CCUV-}$
t_r	Output rise time	---	120	220	ns	
t_f	Output fall time	---	50	80		
t_{sd}	Shutdown propagation delay	---	350	---		
t_d	Output deadtime (HO or LO) (IRS2153D)	0.65	1.1	1.75	µs	
t_d	Output deadtime (HO or LO) (IRS21531D)	0.35	0.6	0.85	µs	
I_{O+}	Output source current	---	180	---	mA	
I_{O-}	Output sink current	---	260	---		
Bootstrap FET Characteristics						
V_{B_ON}	V_B when the bootstrap FET is on	---	13.7	---	V	
I_{B_CAP}	V_B source current when FET is on	40	55	---	mA	C_{BS}=0.1 uF
I_{B_10V}	V_B source current when FET is on	10	12	---		V_B=10 V

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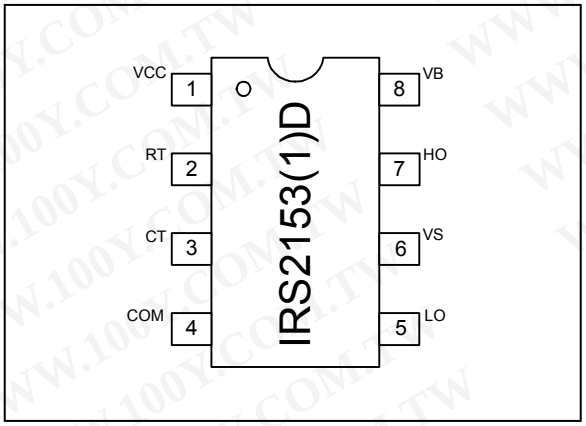
Lead Definitions



Lead	
Symbol	**Description**
V_{CC}	Logic and internal gate drive supply voltage
R_T	Oscillator timing resistor input
C_T	Oscillator timing capacitor input
COM	IC power and signal ground
LO	Low-side gate driver output
V_S	High voltage floating supply return
HO	High-side gate driver output
V_B	High side gate driver floating supply

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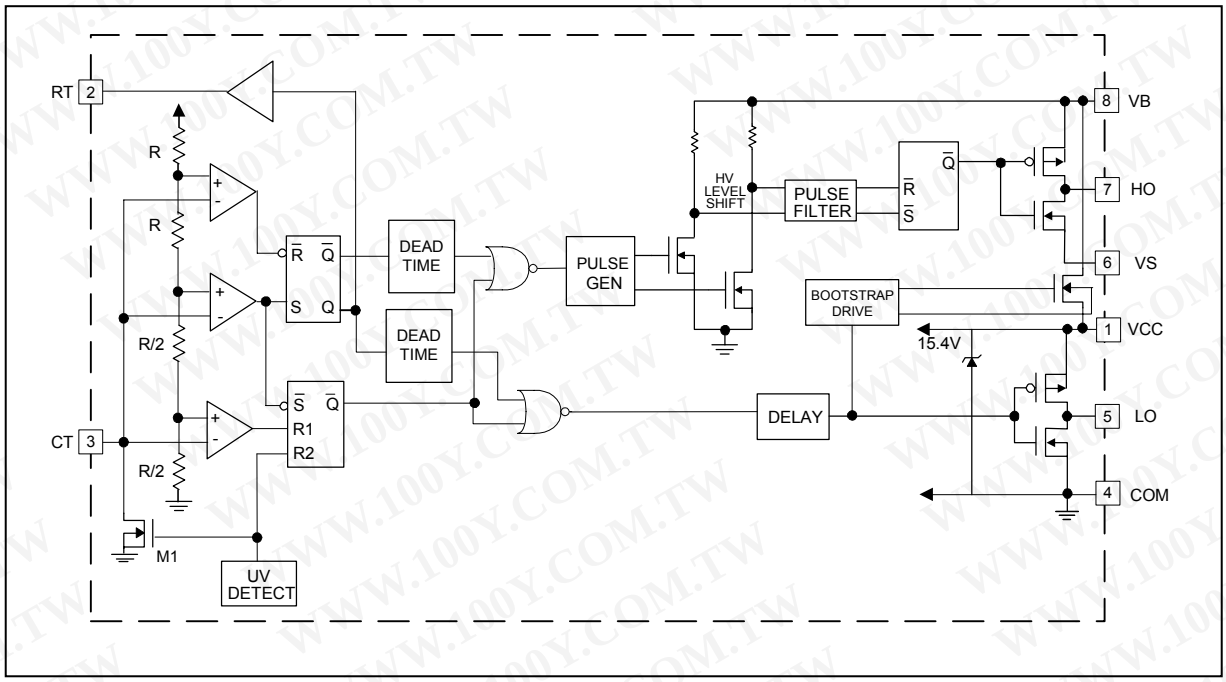
Functional Block Diagram



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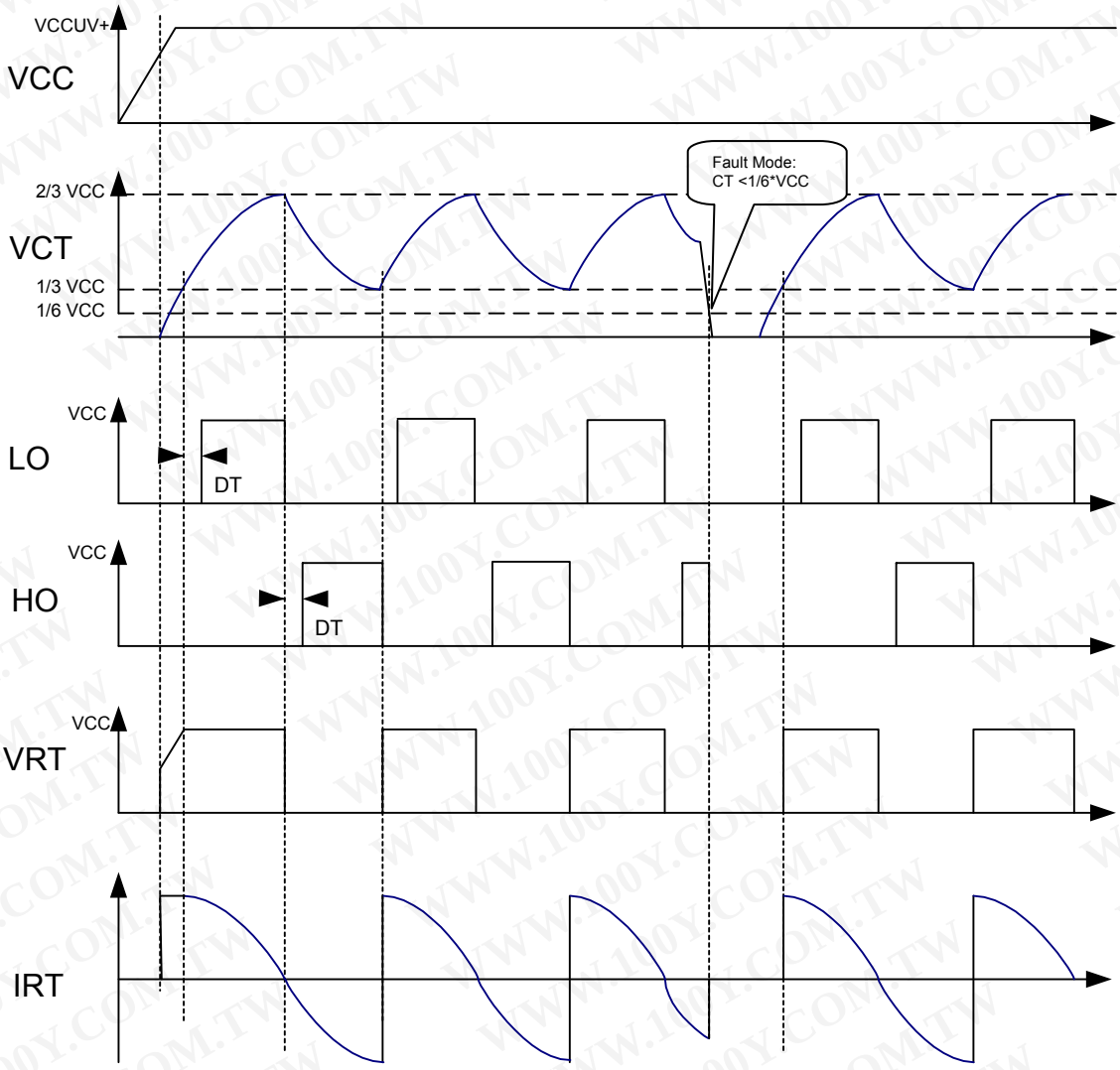
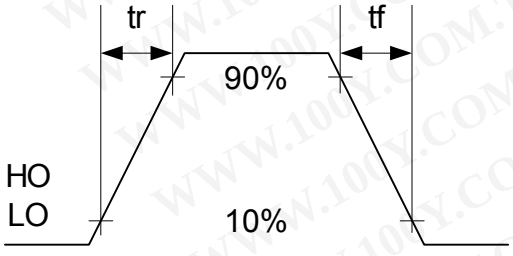
Timing Diagram

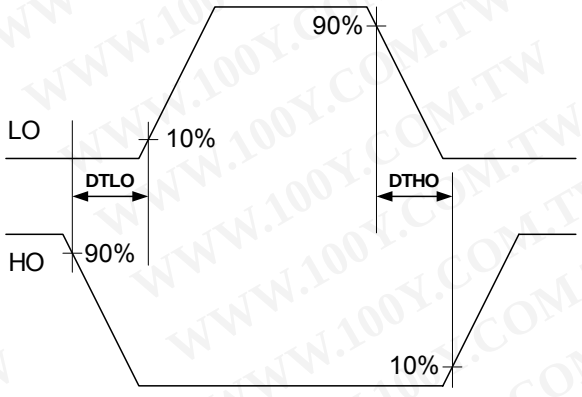
Operating Mode



Switching Time Waveform



Deadtime Waveform



Functional Description

Under-voltage Lock-Out Mode (UVLO)

The under-voltage lockout mode (UVLO) is defined as the state the IC is in when V_{CC} is below the turn-on threshold of the IC. The IRS2153(1)D under voltage lock-out is designed to maintain an ultra low supply current of less than 170 μA, and to guarantee the IC is fully functional before the high and low side output drivers are activated. During under voltage lock-out mode, the high and low-side driver outputs HO and LO are both low.

Supply voltage

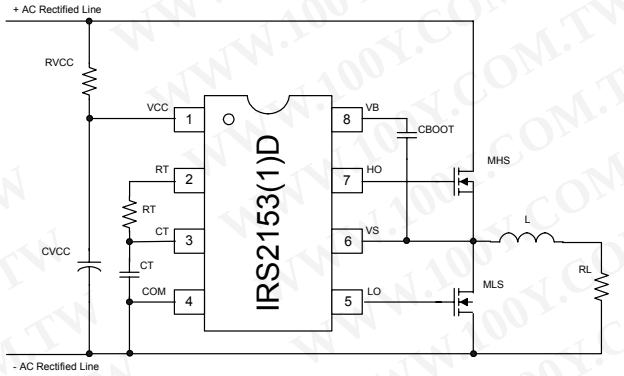


Fig. 1 Typical Connection Diagram

Fig. 1 shows an example of supply voltage. The start-up capacitor (C_{VCC}) is charged by current through supply resistor (R_{VCC}) minus the start-up current drawn by the IC. This resistor is chosen to provide sufficient current to supply the IRS2153(1)D from the DC bus. C_{VCC} should be large enough to hold the voltage at Vcc above the UVLO threshold for one half cycle of the line voltage as it will only be charged at the peak, typically 0.1 uF. It will be necessary for R_{VCC} to dissipate around 1 W.

The use of a two diode charge pump made of DC1, DC2 and CVS (Fig. 2) from the half bridge (V_S) is also possible however the above approach is simplest and the dissipation in R_{VCC} should not be unacceptably high.

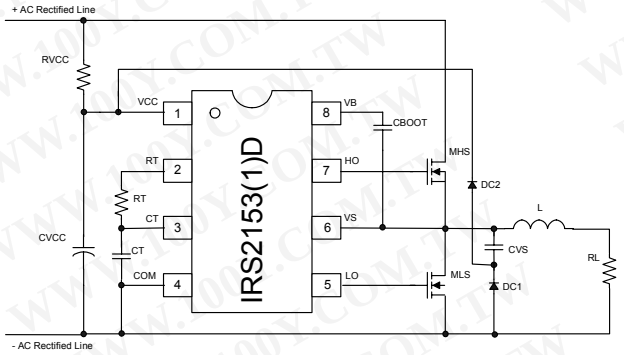


Fig. 2 Charge pump circuit

The supply resistor (R_{VCC}) must be selected such that enough supply current is available over all operating conditions.

Once the capacitor voltage on V_{CC} reaches the start-up threshold V_{CCUV+}, the IC turns on and HO and LO begin to oscillate.

Bootstrap MOSFET

The internal bootstrap FET and supply capacitor (C_{BOOT}) comprise the supply voltage for the high side driver circuitry. The internal boostrap FET only turns on when LO is high. To guarantee that the high-side supply is charged up before the first pulse on pin HO, the first pulse from the output drivers comes from the LO pin.

Normal operating mode

Once the V_{CCUV+} threshold is passed, the MOSFET M1 opens, RT increases to approximately V_{CC} (V_{CC}-V_{RT+}) and the external CT capacitor starts charging. Once the CT voltage reaches V_{CT-} (about 1/3 of V_{CC}), established by an internal resistor ladder, LO turns on with a delay equivalent to the deadtime (t_d). Once the CT voltage reaches V_{CT+} (approximately 2/3 of V_{CC}), LO goes low, RT goes down to approximately ground (V_{RT-}), the CT capacitor discharges and the deadtime circuit is activated. At the end of the deadtime, HO goes high. Once the CT voltage reaches V_{CT-}, HO goes low, RT goes high again, the deadtime is activated. At the end of the deadtime, LO goes high and the cycle starts over again.

The following equation provides the oscillator frequency:

$$f \sim \frac{1}{1.453 \times RT \times CT}$$

This equation can vary slightly from actual measurements due to internal comparator over- and under-shoot delays. For a more accurate determination of the output frequency, the frequency characteristic curves should be used (RT vs. Frequency, page 3).

Shut-down

If CT is pulled down below V_{CTSD} (approximately 1/6 of V_{CC}) by an external circuit, CT doesn't charge up and oscillation stops. LO is held low and the bootstrap FET is off. Oscillation will resume once CT is able to charge up again to V_{CT-}.

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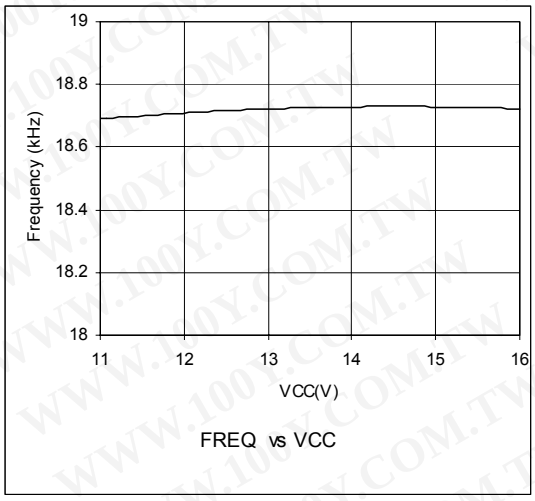

Fig. 3

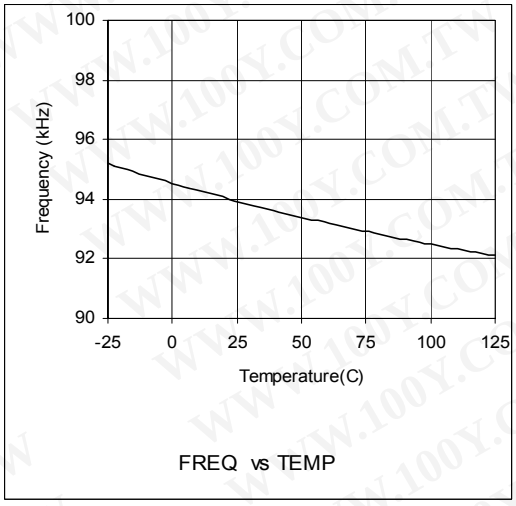

Fig. 4

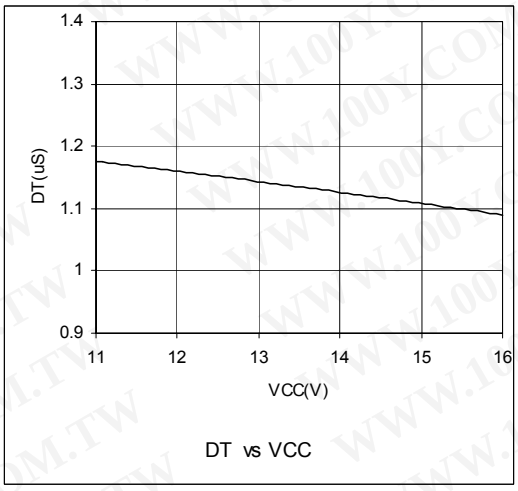

Fig. 5 (IRS2153D)

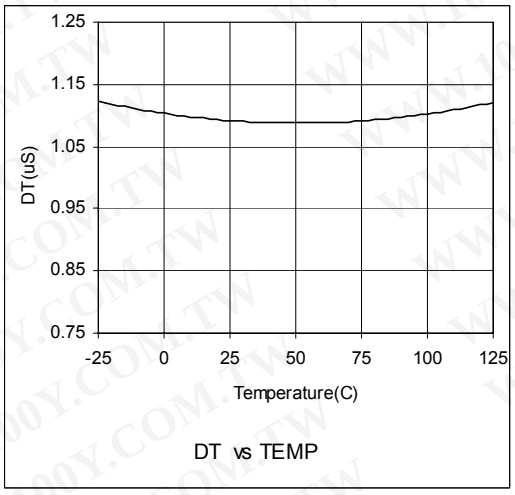

Fig. 6 (IRS2153D)

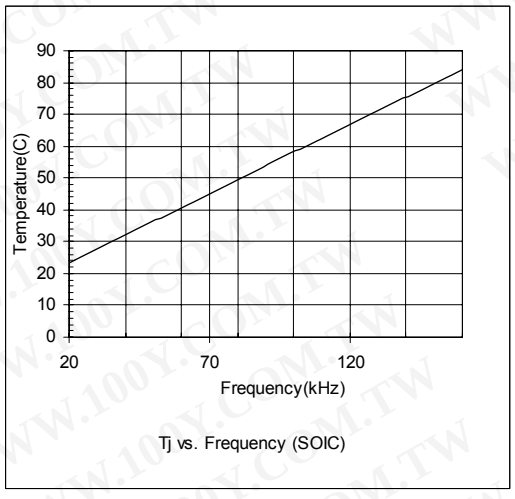

Fig. 7

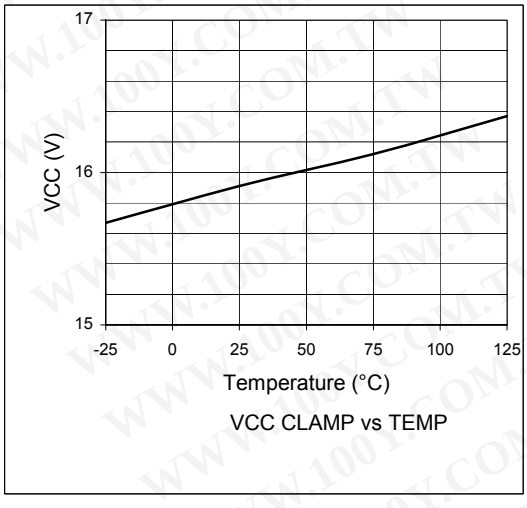

Fig. 8

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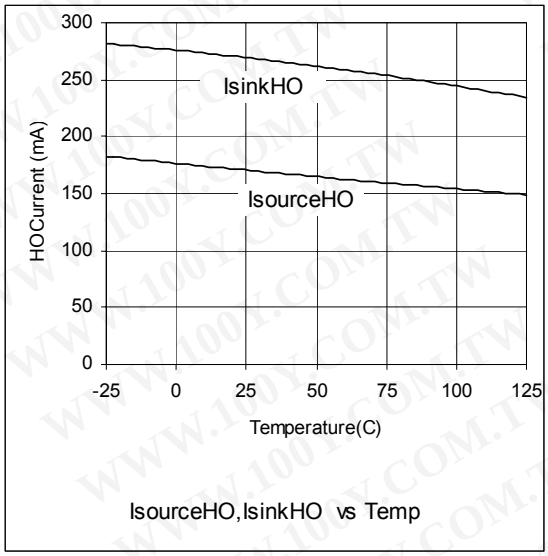

Fig. 9

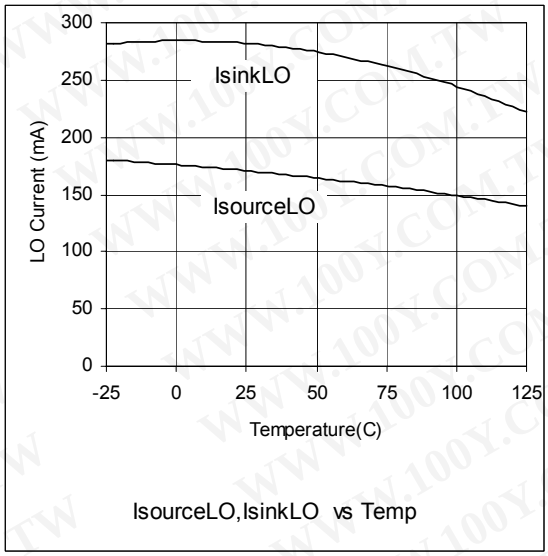

Fig. 10

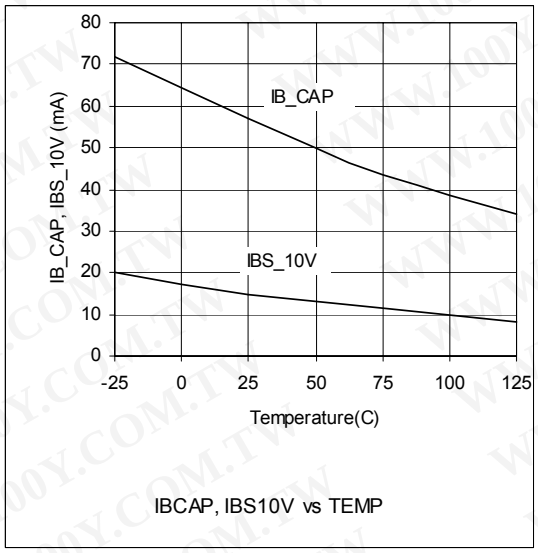

Fig. 11

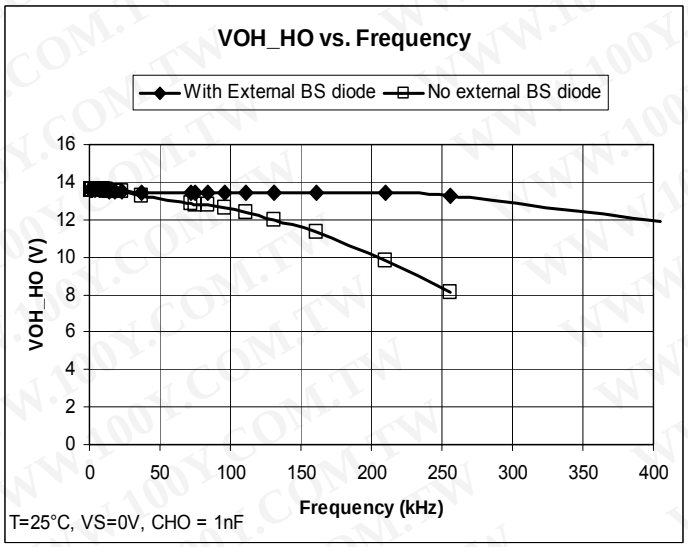

Fig. 12

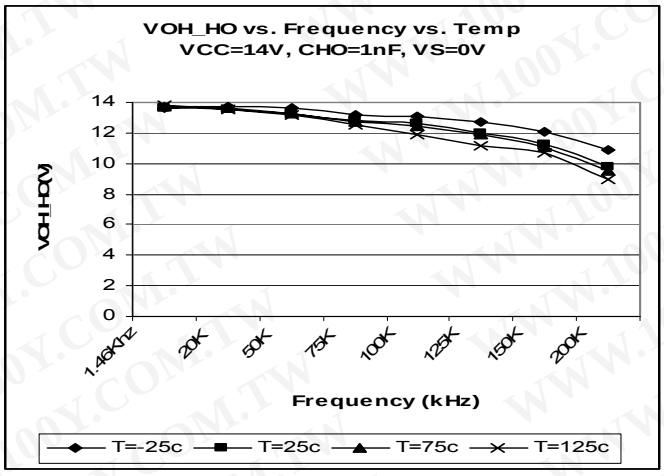

Fig. 13

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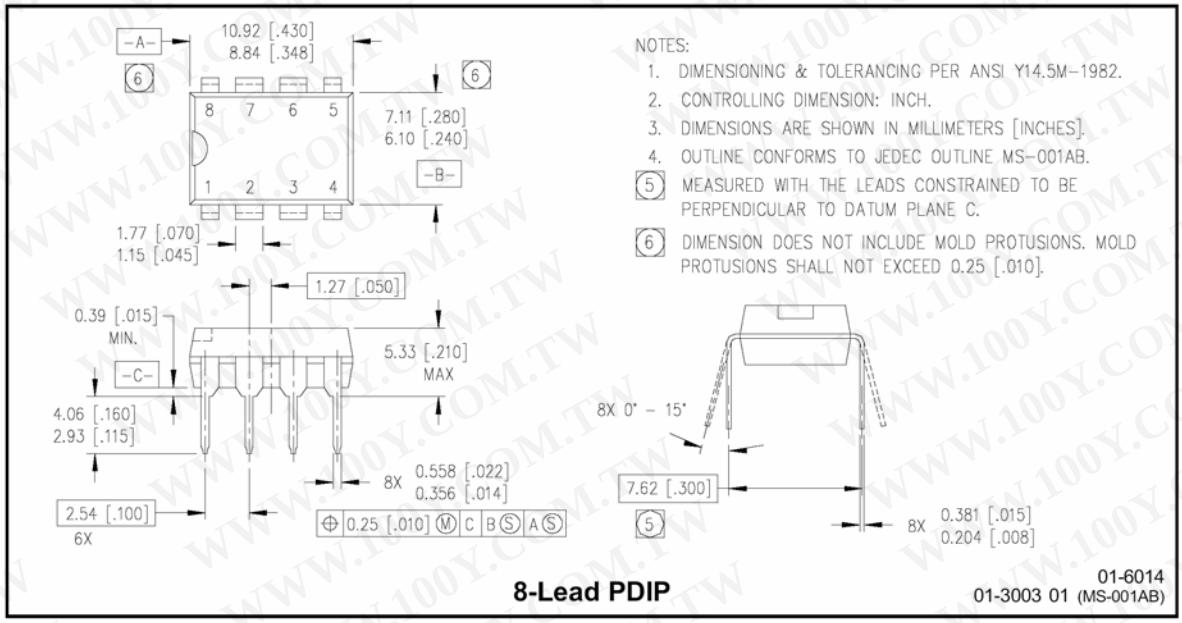
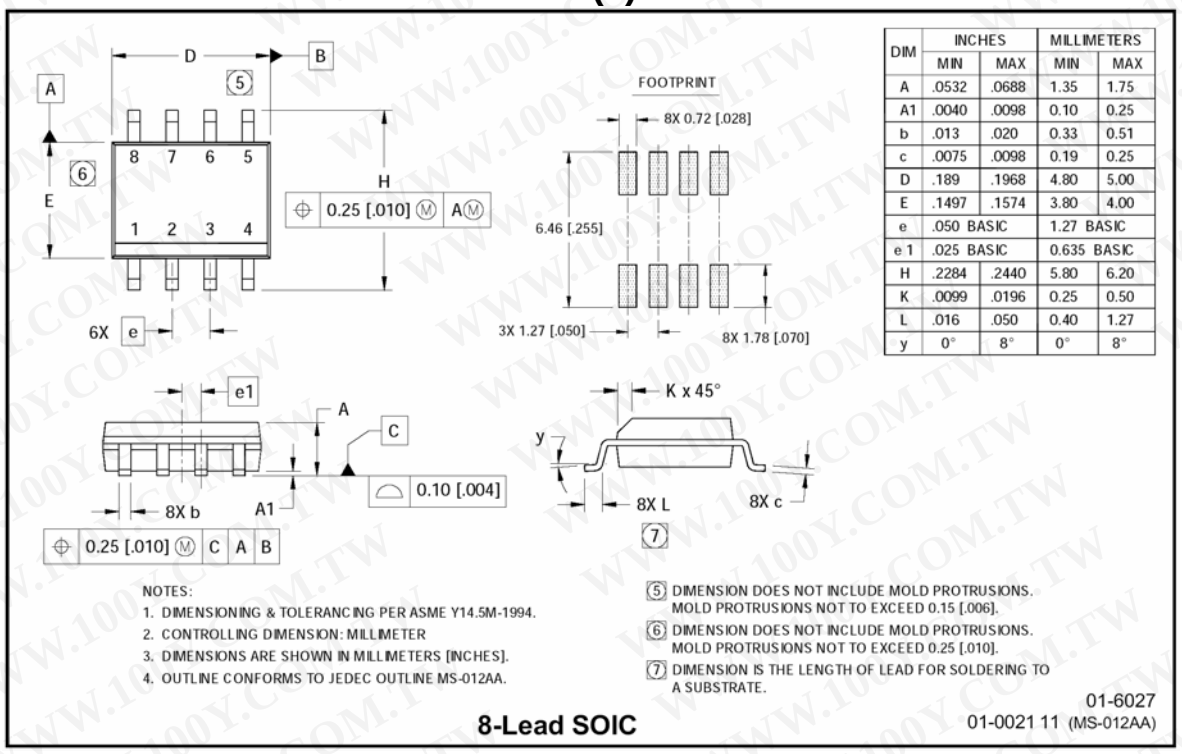


IRS2153(1)DPbF



DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	.0532	.0688	1.35	1.75
A1	.0040	.0098	0.10	0.25
b	.013	.020	0.33	0.51
c	.0075	.0098	0.19	0.25
D	.189	.1968	4.80	5.00
E	.1497	.1574	3.80	4.00
e	.050 BASIC		1.27 BASIC	
e1	.025 BASIC		0.635 BASIC	
H	.2284	.2440	5.80	6.20
K	.0099	.0196	0.25	0.50
L	.016	.050	0.40	1.27
y	0°	8°	0°	8°

IRS2153(1)DSPbF

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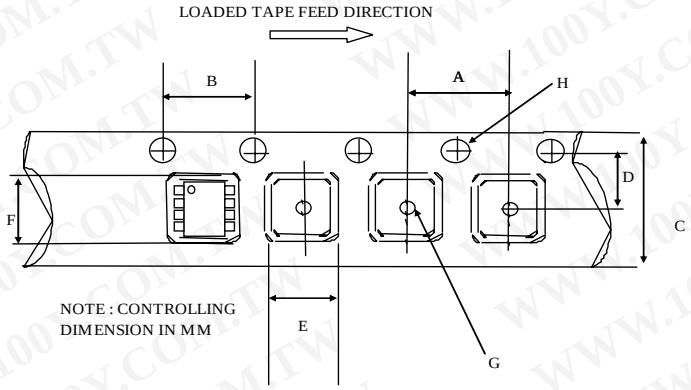
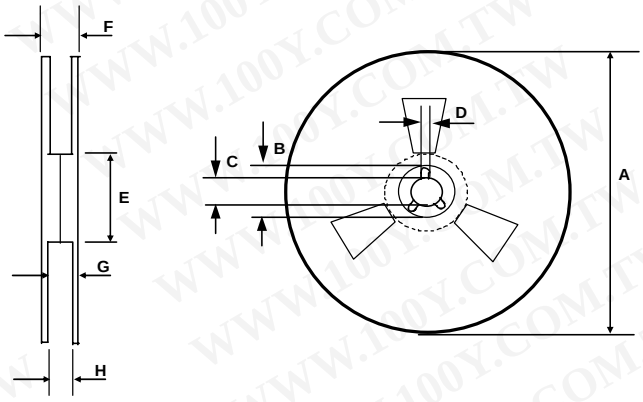


CARRIER TAPE DIMENSION FOR 8SOICN

	Metric		Imperial	
Code	Min	Max	Min	Max
A	7.90	8.10	0.311	0.318
B	3.90	4.10	0.153	0.161
C	11.70	12.30	0.46	0.484
D	5.45	5.55	0.214	0.218
E	6.30	6.50	0.248	0.255
F	5.10	5.30	0.200	0.208
G	1.50	n/a	0.059	n/a
H	1.50	1.60	0.059	0.062



REEL DIMENSIONS FOR 8SOICN

	Metric		Imperial	
Code	Min	Max	Min	Max
A	329.60	330.25	12.976	13.001
B	20.95	21.45	0.824	0.844
C	12.80	13.20	0.503	0.519
D	1.95	2.45	0.767	0.096
E	98.00	102.00	3.858	4.015
F	n/a	18.40	n/a	0.724
G	14.50	17.10	0.570	0.673
H	12.40	14.40	0.488	0.566

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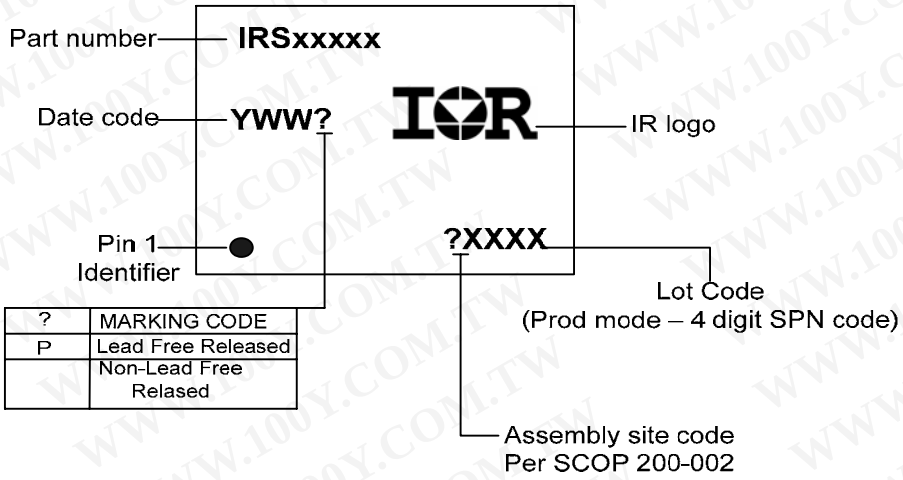
PART MARKING INFORMATION



?	MARKING CODE
P	Lead Free Released
	Non-Lead Free Relased

ORDER INFORMATION

8-Lead PDIP IRS2153DPbF

8-Lead PDIP IRS21531DPbF

8-Lead SOIC IRS2153DSPbF

8-Lead SOIC IRS21531DSPbF

8-Lead SOIC Tape & Reel IRS2153DSTRPbF

8-Lead SOIC Tape & Reel IRS21531DSTRPbF

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The SOIC-8 is MSL2 qualified.
This product has been designed and qualified for the industrial level.
Qualification standards can be found at www.irf.com <http://www.irf.com>
IR WORLD HEADQUARTERS: 233 Kansas St., El Segundo, California 90245, Tel: (310) 252-7105
Data and specifications subject to change without notice. 6/27/2006