

LC² MOS 8-Bit μP Compatible 12-Bit DAC

AD7548

FEATURES

8-Bit Bus Compatible 12-Bit DAC
All Grades 12-Bit Monotonic Over Full
Temperature Ranges
Operation Specified at +5V, +12V
or +15V Power Supply
Low Gain Drift of 5ppm/°C Maximum
Full 4 Quadrant Multiplication
Skinny DIP and Surface Mount Packages

APPLICATIONS

8-Bit Microprocessor Based Control Systems Programmable Amplifiers Function Generation Servo Control

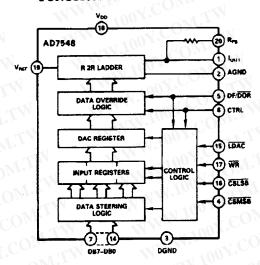
GENERAL DESCRIPTION

The AD7548 is a 12-bit monolithic CMOS D/A converter for use with 8-bit bus microprocessors. Data is loaded in two bytes to input holding registers as shown in the block diagram opposite. The AD7548 can be configured to accept either left- or right-justified data, least significant byte or most significant byte first, using standard TTL compatible control inputs.

A separate load DAC control input allows the user the choice of updating the analog output coincident with loading new data to the DAC input register or at any time after the data loading event. This feature is especially important in multi-DAC systems where simultaneous update of all DACs is required.

The new Linear Compatible CMOS (LC²MOS) process used in the manufacture of the AD7548 allows precision thin-film linear circuitry and high-speed low-power CMOS logic to be integrated on the same small chip. The high-speed logic allows direct interfacing to most of the popular 8-bit microprocessors.

FUNCTIONAL BLOCK DIAGRAM



PRODUCT HIGHLIGHTS

- Microprocessor Compatibility
 High speed input control (TTL/5V CMOS compatible)
 allow direct interfacing to most of the popular 8-bit
 microprocessors.
- Guaranteed Monotonicity
 The AD7548 is guaranteed monotonic to 12-bits over the full temperature range for all grades and at all specified supply voltages.
- Selectable Data Input Format
 Left- or right-justified data, least significant or most significant
 byte first. This allows the AD7548 to be interfaced with
 microprocessors using either Motorola or Intel-type data
 formatting.
- Monolithic Construction
 For increased reliability and reduced package size 0.3"
 20-pin DIP and 20-terminal surface mount packages.
- 5. Single Supply Operation See Figure 8.
- 6. Low Gain Error and Gain Error T.C.

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AD7548 — SPECIFICATIONS $(V_{ea} = +5V, V_{ref} = +10V; V_{PH1} = V_{PH2} = 0V$. All specifications T_{min} to T_{max} unless otherwise specified)

Parameter	J, A Versions	K, B Versions	S Version	T Version	Unite	Test Conditions/Comments
ACCURACY	431	11/1/	V.CO	W	4	AM TO THE
Resolution	12	12	12	12	Bits	COM.
Relative Accuracy	±1	± 1/2	±1/ C	±1/2	LSB max	MIN W. CO. TIN
Differential Monlinearity	±1	±1/2	(±1	± 1/2	LSB max	All grades guaranteed monotonic to 12-bits over temperature.
Full Scale Error	±6	±3	±6	COM.T	LSB max	Measured using internal R _{FB} and includes effects of leakage current and gain TC. Full Scale Error can be trimmed to zero.
Gain Temperature Coefficient²; ∆Gain/∆Temperature Output Leakage Current	N±5	±5	±5	±5	ppm/*C max	Typical value is 2ppm/°C
I _{OUT} (Pin 1)		1/1/1	-110	1	7.7.	M. 100 s. OW. I.
+25℃	±5	±5	±5	±5 ()	nA mex	All digital inputs = 0V
T _{min} to T _{max}	± 25	±25	± 150	±150	oA max	W 1100 1 111
REFERÊNCE INPUT Input Resistance, Pin 19	7 20	7 20	7 20	7 20	kΩ min kΩ max	Typical Input Resistance = 11kΩ
DIGITAL INPUTS				100 >	-0M.1	COM.
VIII (Input High Voltage)	2.4	2.4	2.4	2.4	V min	WWW.
VIL (Input Low Voltage) Inv (Input Current)	0.8	0.8	0.8	0.8	V max	CM WWW.100 COM
+ 25°C	±1.43	±1	±1	±1100	μAmax	V _{IN} = 0V or V _{DD}
T _{min} to T _{max}	±10	±10	± 10	± 10	μA max	THE WAY TO STATE OF THE STATE O
C _{DN} (Input Capacitance) ²	7	7	7	7 10	pF max	11. A WILL TO
POWER SUPPLY VDD Range IDD	4.75/5.25 2	4.75/5.25	4.75/5.25	4.75/5.25 2	V min/V max mA max	Specifications guaranteed over this range All digital inputs V _{II} , or V _{IH}
- 100	300	300	300	300	µA max	All digital inputs OV or VDD

$\begin{array}{l} \text{SPECIFICATIONS}^1 \text{ ($v_{\text{no}} = + 12$V to $+ 15$V, $V_{\text{NST}} = + 10$V; $V_{\text{PNI}} = V_{\text{PNI}2} = 0$V. All specifications T_{min} to T_{max} unless otherwise specified) } \end{array}$

Parameter	J, A Versions	K, B Versions	S Version	T Version	Units 100	Test Conditions/Comments
ACCURACY	1001	Co	TW	W	10	N.C. TIN WY
Resolution	12	12	12	12	Bits	COMP
Relative Accuracy	±1	± 1/2	±1	± 1/2	LSB max	M. T.
Differential Nonlinearity	±1	±1/2	±1	±1/2	LSB max	All grades guaranteed monotonic to 12-bits over temperature.
Full Scale Error	±6	±3	±6	±3	LSB mex	Measured using internal RFB and includes effects of leakage current and gain TC. Full Scale Error can be trimmed to zero.
Gain Temperature Coefficient ² ; \[\Delta \text{Gain} \Delta \text{Temperature} \] Output Leakage Current	±5	± 5	±5/1.T	±5	ppm/*C max	Typical value is 2ppm/°C
Lour (Pin 1)		The	COM			All district forms - ON
+ 25°C	±5	±5	±5	±5 ±150	nA max	All digital inputs = 0V
T _{min} to T _{max}	+ 25	± 25	±150	₹130	n/s max	
REPERENCE INPUT Input Resistance, Pin 19	7 20	7 1 1 0 0 20	7 20	7 20	kΩ min kΩ max	Typical Input Resistance = 11kΩ
DIGITALINPUTS		-111 JU	of (Mrs		COM
V _{IH} (Input High Voltage)	2.4 V	2.4	2.4	2.4	Vmin	MM. 100X
V _{IL} (Input Low Voltage) l _D (Input Current)	0.8	0.8	0.8,	0.8	V max	MMM.To.
+ 25℃	±1	±1	±1	±1	μA max	V _{IN} = 0V or V _{DD}
T _{min} to T _{men}	± 10	± 10	± 10	± 10	μA max	
C _{IN} (Input Capacitance) ²	7	7	7	7	pF max	
POWER SUPPLY		MAL	- 4003		I	
V _{DD} Range	11.4/15.75	11.4/15.75	11.4/15.75	11.4/15.75	V min/V max	Specifications guaranteed over this range
Ipp	13	3	3	3	mA max	All digital inputs VIL or VIH
-00	Li	la i	l i	Li	mA max	All digital inputs OV or VDD

Temperature range as follows: J, K Versions: - 40°C to + 85°C A, R Versions: - 40°C to + 85°C S, T Versions: - 55°C to + 125°C Guaranteed by design but not production tested.

Specifications subject to change without notice.

TIMING CHARACTERISTICS $(v_{10} = +5V, V_{10F} = +10V, V_{POIT} = V_{POI2} = 0V unless otherwise stated)$

Parameter	Limit at T _A = 25°C	Limit ² at T _A = -40°C to +85°C	Limit ² at T _A = -55°C to +125°C	Units	Test Conditions/Comments
t _{DS}	240	240	290	ns min	Data Valid Setup Time
tDH	50	50	70	ns min	Data Valid Hold Time
tows C	30	40	50	ns min	CSMSB or CSLSB to WR Setup Time
tcwn	15	20	25	ns min	CSMSB or CSLSB to WR Hold Time
Lws	30	40	50	ns min	LDAC to WR Setup Time
rws Lwa	15	20	25	ns min	LDAC to WR Hold Time
LWH	250	280	320	ns min	Write Pulse Width

TIMING CHARACTERISTICS $(v_{sp} = +12V to +15V, V_{rest} = +10V, V_{rest} = V_{resc} = 0V unless otherwise stated)$

Parameter	Limit at T _A = 25°C	Limit 2 at T _A = -40°C to +85°C	Limit ² at T _A = -55°C to +125°C	Units	Test Conditions/Comments
DS	160	190	230	ns min	Data Valid Setup Time
DH	30	30	50	ns min	Data Valid Hold Time
CWS	30	40	50	ns min	CSMSB or CSLSB to WR Setup Time
CWH	15 C	20	25	ns min	CSMSB or CSLSB to WR Hold Time
LWS	30	40	50	ns min	LDAC to WR Setup Time
LWH	15	20	25	ns min	LDAC to WR Hold Time
LWR	170	200	240	ns min	Write Pulse Width

AC PERFORMANCE CHARACTERISTICS These characteristics are included for Design Guidance only and are not subject to test. $(V_{REF} = +\ 10V; V_{PIN1} = V_{PN2} = 0V, Output Amplifier is AD544 except where stated)$

Parameter Version		= +5V T _A = T _{MIN} , T _{MAX}		12V to + 15V T _A = T _{MIN} , T _{MAX}	Units	Test Conditions/Comments
Output Current Settling Time	1.5 COM	T.T.A.	WW	M.100X.C	nz tàb	To 0.01% of full scale range. l_{OUT} load = 100 Ω , C_{EXT} – 13pF. DAC register alternately loaded with all 1s and all 0s
Digital to Analog Glitch Impulse	400	M.TW M.TW	330	M.M.100A	nV-sec typ	Measured with $V_{RBF} = 0V$, $I_{OUT} load = 100\Omega$, $C_{EXT} = 13 pF$. DAC register alternately loaded with all 1s and all 0s
Multiplying Feedthrough Error ³	3" C	SMA	3	NWW.10	mV p-p typ	V _{REF} = ± 5V, 10kHz sine wave DAC register loaded with all 0s.
Total Harmonic Distortion	-85	GOM.TV	- 85	MMM	dB typ	V _{REF} = 6V rms @ 1kHz. DAC register loaded with all 1s.
Power Supply Rejection Δ GAIN/Δ V _{DD}	± 0.015	± 0.03	±0.01	±0.02	% per % max	ΔV _{DD} = ±5%
Output Capacitance L _{OUT} (Pin I)	200 100	200 100	200 100	200 100	pF max pF max	DAC register loaded with all 1s. DAC register loaded with all 0s.
Output Noise Voltage Density (10Hz-100kHz)	15	TOOX.COL	15	- WW	nV/√ H z typ	Measured between R _{FB} and l _{OUT}

NOTES

Guaranteed by design but not production tested.

*Temperature range as follows: J, K Versions: -40°C to +85°C A, B Versions: -40°C to +85°C C, T.Versions: -55°C m + 155°C

Specifications ambject to change without notice.

Feedthrough can be further reduced by connecting the metal lid on the ceramic package (D-20) to DGND.

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ABSOLUTE MAXIMUM RATINGS* (T _A = +25°C unless otherwise noted)	
V _{DD} (pin 18) to DGND	+ 17V
	±25V
	±25V
Digital Input Voltage (pins 4-17) to DGND	-0.3V, V _{DD} +0.3V
V _{PIN 1} to DGND	$-0.3V, V_{DD} + 0.3V$
AGND to DGND	$-0.3V$, $V_{DD} + 0.3V$
Power Dissipation (Any Package) To +75°C Derates above +75°C	450mW

Operating Temperature Range				
Commercial (J, K versions)		•		40 to +85℃
Industrial (A, B versions)		Ō	M.	-40°C to +85°C
Extended (S, T versions)				-55°C to +125°C
Storage Temperature	o7 .		0	-65°C to +150°C
Lead Temperature (Soldering, 10sec	2)	٠,		+300°C

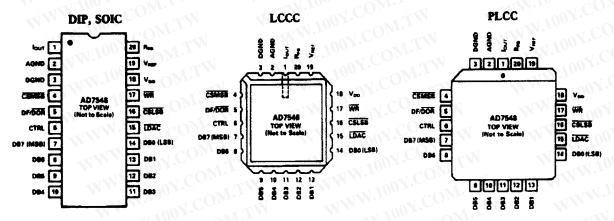
*Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

CAUTION

ESD (electrostatic discharge) sensitive device. The digital control inputs are diode protected; however, permanent damage may occur on unconnected devices subject to high energy electrostatic fields. Unused devices must be stored in conductive foam or shunts. The protective foam should be discharged to the destination socket before devices are removed.



PIN CONFIGURATIONS



ORDERING GUIDE¹

Model ²	Temperature Range	Relative Accuracy	Full-Scale Error	Package Option ³
AD7548JN	-40°C to +85°C	± 1LSB	±6LSB	N-20
AD7548KN	-40°C to +85°C	± 1/2LSB	±3LSB	N-20
AD7548JP	-40°C to +85°C	± 1LSB	±6LSB	P-20A
AD7548KP	-40°C to +85°C	± 1/2LSB	±3LSB	P-20A
AD7548JR	-40°C to +85°C	± 1LSB	±6LSB	R-20
AD7548KR	-40°C to +85°C	± 1/21.SB	± 31.SB	R-20
AD7548AQ	-40°C to +85°C	±1LSB	±6LSB	Q-20
AD7548BQ	-40°C to +85°C	± 1/2LSB	±3LSB	Q-20
AD7548SQ	-55°C to +125°C	±1LSB	±6LSB	Q-20
AD7548TO	-55°C to +125°C	± 1/2LSB	± 3LSB	Q-20
AD7548SE	-55°C to +125°C	± 1LSB	±6LSB	E-20A
AD7548TE	-55°C to +125°C	± 1/21.SB	± 3LSB	E-20A
		1	1	

NOTE

¹Analog Devices reserves the right to ship ceramic (package outline D-20) packages in lieu of cerdip (package outline Q-20) packages.

To order MIL-STD-883, Class B processed parts, add/883B to part number. Contact your local sales office for military data sheet.

³E - Leadless Ceramic Chip Carrier; N - Plastic DIP; P - Plastic Leaded Chip Carrier; Q - Cerdip; R - SOIC.

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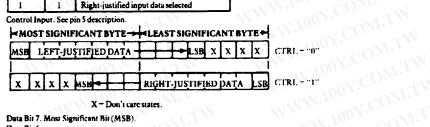
AD7548

PIN FUNCTION DESCRIPTION

PIN	MNEMONIC	DESCRIP	TION								
i	logr	DAC current OUT bus. Normally terminated at virtual ground of output amplifier. Analog Ground.									
2	AGND										
. 3	DGND	Digital Ground.									
4	CSMSB	Chip Select	Most Sig	nificant (MS) Byte. Active Low Input. Used in	combination with WK to load						
		external da	a into the	input register or in combination with WR and	LDAC to load external data into						
		both input	and DAC	registers.	1001.						
- 5	DF/DOR	Data Format/Data Override. When this input is LOW, data in the DAC register is forced to one of									
				lected by CTRL. When the override signal is re							
		returns to re	effect the	value in the DAC register. With DF/DOR HIG	H. CTRL selects either a left						
		or right just	ified inpu	it data format. For normal operation. DF/DOR	is held HIGH.						
		DF/DOR	CTRL	FUNCTION	W.100 COM. 1						
		0	0	DAC register contents overridden by all 0's	W. TIOOY.						
		0	. 1	DAC register contents overriden by all 1's	TAN M. TO COM.						
			0 0	Left-justified input data selected	N V COOK						
		11	$\sqrt{1}$	Right-justified input data selected	TAN TAN COMP.						

DF/DOR	CTRL	FUNCTION
0	0	DAC register contents overridden by all 0's
0	1	DAC register contents overriden by all 1's
1	0	Left-justified input data selected
1	1	Right-justified input data selected

WWW.100Y.COM.TW WWW.100Y.COM.TW WWW.100Y.COM Control Input. See pin 5 description.



X = Don't care states.

		MSB LEFT-JUSTIFIED DATA + + + LSB X X X X X X CTRL - """
		X X X MSB RIGHT-JUSTIFIED DATA LSB CTRL - "1"
		X = Don't care states.
7.	DB7	Data Bir 7. Moss Significant Bir (MSB).
8	D86	Data Bit 6.
9	DB5	Data Bit 5.
10	DB4	Data Bit 4.
11	DB3	Data Bit 3.
12	DB2	Data Bit 2.
13	DB1	Data Bit 1.
14	DB0	Data Bit 0. Least Significant Bit (LSB).
15	LDAC	Load DAC Input, active LOW. This signal, in combination with others, is used to load the DAC
		register from either the input register or the external data bus.
16	CSUSB	Chip Select Least Significant (LS) Byte. Active LOW input. Used in combination with WR to load
		external data into the input register or in combination with WR and LDAC to load external data
		into both input and DAC registers.

anto the ADJ 548 input register and in transferring data from the input register to the DAC regist								
WR	CSMSB	CSLSB	LDAC	FUNCTION				
0	AW	0	1	Load LS Byte to Input Register.				

WRITE Input. This active low signal, in combination with others is used in loading external data

-				
0		0	1	Load LS Byte to Input Register.
0	I .	0	0	Load LS Byte to Input Register and DAC Register.
0	0	1	1	Load MS Byte to Input Register.
0	0	1	0	Load MS Byte to Input Register and DAC Register.
0	1	<1	0	Load Input Register to DAC Register.
1	X	X	X	No Data Transfer_

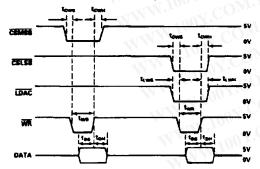
19

+ 5V to + 15V Supply Input. Reference Voltage Input.

Feedback Resistor. Used for normal D/A conversion.

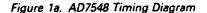
CONTROL INPUT INFORMATION

Figure 1a shows the data load timing diagram for the AD7548. Figure 1b shows the simplified input control structure of the AD7548.



NOTES 1. ALL INPUT SIGNAL RISE AND FALL TIMES MEASURED FROM 10% TO 90% OF +SV t. = t,= 20ms.

- 2. TIMING MEASUREMENT REFERENCE LEVEL IS $\frac{V_{\rm H}+V_{\rm L}}{2}$
- 3. CSMSS (MM 4) AND CSLSS (MN 16) MAY BE INTERCHANGED.
- 4. FOR LEFT JUSTIFIED DATA CTRL = +0V WITH DF/DOR = +5V, FOR RIGHT-JUSTIFIED DATA CTRL = +5V WITH DF/DOR = +3V.



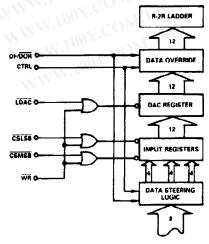


Figure 1b. Simplified AD7548 Input Control Structure

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GENERAL CIRCUIT INFORMATION

The simplified D/A circuit is shown in Figure 2. An inverted R-2R ladder structure is used, which steers binarily weighted currents between I_{OUT} and AGND, thus maintaining a constant current in each ladder leg independent of the switch state.

The input resistance at V_{REF} is constant and equal to the value "R" in Figure 2. Since the input resistance is constant, the reference terminal can be driven by a reference voltage or a reference current, ac or dc, of positive or negative polarity. (If a current source is used, a low temperature coefficient external R_{FR} is recommended to define scale factor).

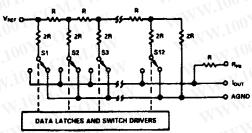


Figure 2. AD7548 Simplified Functional Diagram

EQUIVALENT CIRCUIT ANALYSIS

Figure 3 shows an equivalent circuit for the analog section of the AD7548 D/A converter. The current source $I_{\rm LEAKAGE}$ is composed of surface and junction leakages. The resistor $R_{\rm O}$, denotes the equivalent output resistance of the DAC which varies with input code (excluding all 0's code) from 0.8R to 2R, where R is typically $11k\Omega$. $C_{\rm OUT}$ is the capacitance due to the current steering switches and varies from about 50pF to 120pF (typical values) depending upon the digital input. $g(V_{\rm REF}, N)$ is the Thevenin equivalent voltage generator due to the reference input voltage, $V_{\rm REF}$, and the transfer function of R-2R ladder, N.

For further information on CMOS multiplying D/A converters refer to "Application Guide to CMOS Multiplying D/A Converters" available from Analog Devices, Publication Number G479-15-8/78.

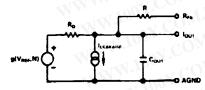


Figure 3. AD7548 Equivalent Analog Output Circuit

DATA LOADING

The AD7548 accepts incoming data in either left-justified format or right-justified format depending on the control inputs DF/DOR and CTRL.

(See pin description of \overline{DDR} and \overline{CTRL} on preceding page).

Two operating modes are possible for controlling the transfer of data from the input register to the DAC register, the automatic transfer mode and the strobed transfer mode.

AUTOMATIC TRANSFER MODE

This is the simplest and fastest method of transferring data to the DAC register. It is facilitated by connecting LDAC to either CSMSB, as shown in Figure 10, or CSLSB.

Figure 4 shows the timing diagram for automatic transfer of 8

+4-bit data to the DAC register. The first write cycle loads the first byte of data to the input register. The second write cycle loads the second byte of data to the input register and automatically transfers both bytes to the DAC register.

Updating a single byte (High or Low) in the DAC register can be achieved in one write cycle using the automatic transfer mode.

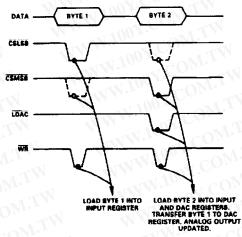


Figure 4. Automatic Transfer Mode

STROBED TRANSFER MODE

Figure 5 shows the timing diagram for the strobed transfer of 8 + 4-bit data to the DAC register. Three write cycles are required for this transfer mode. The first two write cycles sequentially load bytes 1 and 2 into the input register. The third write cycle transfers data from the input register to the DAC register.

The strobed transfer mode allows the DAC registers of several AD7548's to be updated simultaneously, as shown in Figure 13, by means of a master strobe signal connected to the LDAC of each device.

A single byte of data (High or Low) can be transferred to the DAC register in two write cycles using the strobed transfer

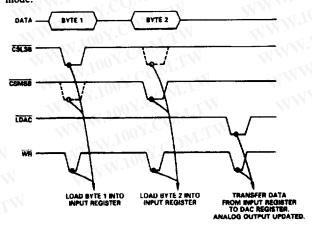


Figure 5. Strobed Transfer Mode

DATA OVERRIDE

The contents of the DAC register can be overridden by pulling DF/DOR (pin 5) LOW. The CTRL (pin 6) input then determines whether the DAC register data is overidden by all 0s (CTRL LOW) or all 1s (CTRL HIGH). This feature allows the user to calibrate the AD7548 in circuits such as Figure 6 without calling on the microprocessor to load calibration data.

UNIPOLAR BINARY OPERATION (2-QUADRANT MULTIPLICATION)

Figure 6 shows the analog circuit connections required for unipolar binary operation. With a dc input voltage or current (positive or negative polarity) applied at pin 19, the circuit is a unipolar D/A converter. With an ac input voltage the circuit provides 2-quadrant multiplication (digitally controlled attenuation).

Table I shows the code relationship for the circuit of Figure 6. For full scale trimming the DAC register is loaded with 1111 1111 1111. This is most easily accomplished by using the data override function. R1 is then adjusted for $V_{\rm OUT} = -V_{\rm IN}$ (4095/4096). Alternatively full scale can be adjusted by omitting R1 and R2 and trimming the reference voltage magnitude.

Capacitor C1 provides phase compensation and helps prevent overshoot and ringing when using high speed op amps.

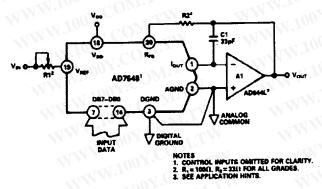


Figure 6. Unipolar Binary Operation

Table I. Unipolar Binary Code Table for Circuit of Figure 6

	ary Number in OAC Register	Analog Output, Vout
MSB		LSB
1111	1111 111	$1 - V_{IN} \left(\frac{4095}{4096} \right)$
1000	0000 000	$0 - V_{IN} \left(\frac{2048}{4096} \right) = -1/2 V_{IN}$
0000	0000 000	$-V_{IN}\left(\frac{1}{4096}\right)$
0000	0000 000	0 OV

BIPOLAR OPERATION (4.QUADRANT MULTIPLICATION)

Figure 7 and Table II illustrate the recommended circuit and code relationship for bipolar operation. The circuit uses offset binary input coding. However, 2's complement coding can be accommodated if the MSB is inverted (done in software) before data is loaded into the DAC.

With the DAC register loaded to 1000 0000, adjust R1 for $V_{OUT} = 0V$ (alternatively one can omit R1 and R2 and adjust the ratio of R3 and R4 for $V_{OUT} = 0V$). Full scale trimming can be accomplished by adjusting the amplitude of V_{IN} or by varying the value of R5.

R3, R4 and R5 must be selected to match within 0.01% and they should be the same type of resistor (preferably metal film) so that their temperature coefficients match. Mismatch of R3 to R4 causes both offset and full scale error. Mismatch of R5 to R4 and R3 causes full scale error.

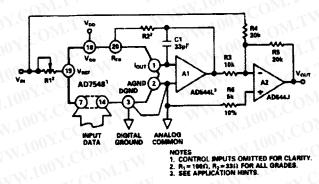


Figure 7. Bipolar Operation (Offset Binary Coding)

Table II. Bipolar Code Table for Offset Binary Circuit of Figure 7

MSE		Binary Nu DAC Ro		Analog Output, V _{OUT}
		1111	- 11	$+V_{1N}\left(\frac{2047}{2048}\right)$
100	0	0000	0001	$+V_{IN}\left(\frac{1}{2048}\right)$
100	0	0000	0000	ov
011		1111	1111	$-V_{IN}\left(\frac{1}{2048}\right)$
000	0	0000	0000	$-V_{IN}\left(\frac{2048}{2048}\right)$

AD7548

SINGLE SUPPLY OPERATION

Figure 8 shows the AD7548 connected in a voltage switching mode. The input voltage is connected to I_{OUT} . The D/A converter output voltage is taken from the V_{REF} pin and has a constant impedance equal to R. R_{FR} is not used in this circuit. The input voltage V_{IN} must always be positive with respect to AGND in order to prevent an internal diode from turning on. To maintain linearity the input voltage should remain within 2.5V of AGND with V_{DD} from +12V to +15V.

The output voltage VOUT of Figure 8 is expressed as

$$V_{OUT} = (V_{IN}) (D) \left(\frac{R_1 + R_2}{R_1} \right)$$

Where D is a fractional representation of the digital input word $(0 \le D \le 4095/4096)$.

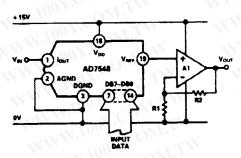


Figure 8. Single Supply Operation Using Voltage Switching Mode

APPLICATION HINTS

Output Offset: CMOS D/A converters in circuits such as Figures 6 and 7 exhibit a code dependent output resistance which in turn cause a code dependent amplifier noise gain. The effect is a code dependent differential nonlinearity term at the amplifier output which, depends on V_{OS} where V_{OS} is the amplifier input offset voltage. To maintain monotonic operation it is recommended that V_{OS} be no greater than $(25 \times 10^{-6})(V_{REF})$ over the temperature range of operation. Suitable op amps are AD517L and AD544L. The AD517L is best suited for fixed reference applications with low bandwidth requirements: it has extremely low offset $(50\mu V)$ and in most applications will not require, an offset trim. The AD544L has a much wider bandwidth and higher slew rate and is recommended for multiplying and other applications requiring fast settling. An offset trim on the AD544L may be necessary in some circuits.

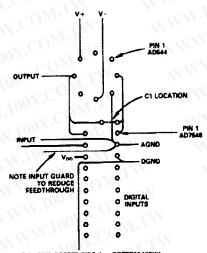
General Ground Management: AC or transient voltages between AGND and DGND can cause noise injection into the analog output. The simplest method of ensuring that voltages at AGND and DGND are equal is to tie AGND and DGND together at

the AD7548. In more complex systems where the AGND and DGND intertic is on the backplane, it is recommended that two diodes be connected in inverse parallel between the AD7548 AGND and DGND pins (1N914 or equivalent).

Temperature Coefficients: The gain temperature coefficient of the AD7548 has a maximum value of Sppm/°C and typical value of 2ppm/°C. This corresponds to worst case gain shifts of 2LSBs and 0.8LSBs respectively over a 100°C temperature range. When trim resistors R1 and R2 are used to adjust full scale range, the temperature coefficient of R1 and R2 should also be taken into account. The reader is referred to Analog Devices Application Note "Gain Error and Gain Temperature Coefficient of CMOS Multiplying DACs", Publication Number E630-10-6/81.

High Frequency Considerations: AD7548 output capacitance works in conjunction with the amplifier feedback resistance to add a pole to the open loop response. This can cause ringing or oscillation. Stability can be restored by adding a phase compensation capacitor in parallel with the feedback resistor.

Feedthrough: The dynamic performance of the AD7548 will depend upon the gain and phase stability of the output amplifier, together with the optimum choice of PC board layout and decoupling components. A suggested printed circuit layout for Figure 6 is shown in Figure 9 which minimizes feedthrough from $V_{\rm REP}$ to the output in multiplying applications.



LAYOUT SHOWS COPPER SIDE (i.e., BOTTOM VIEW)
GAIN TRIM RESISTORS R1 AND R2 OF FIGURE 6 ARE NOT INCLUDED.

Figure 9. Suggested Layout for AD7548 and Op Amp

For additional information on multiplying DACs refer to "Application Guide to CMOS Multiplying D/A Converters", Publication Number G479-15-8/78, available from Analog Devices.

MICROPROCESSOR INTERFACING AD7548 – MC6800 INTERFACE

A typical 6800 configuration using the automatic transfer mode of the AD7548 is shown in Figure 10. Table III gives a sample loading routine written in re-entrant form. Data load and store instructions use extended addressing. The 12-bit data to be passed to the subroutine is stored in locations XXYY and XXYY + 1. The data is considered right-justified with the four most significant bits occupying the lower half of XXYY + 1. The AD7548 is assigned a base address of PPQQ. This address selects the low byte register of the AD7548. Address PPQQ + 1 selects both the high byte register and the LDAC control input.

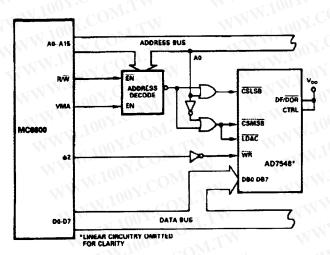


Figure 10. AD7548 - MC6800 Interface (Automatic Transfer Mode)

Table III. Sample Routine for AD7548 - MC6800 Interface

	ISR	WWZZ	Jump to AD7548 subroutine
WWZZ	PSH A		Push A onto stack
	TPA		
	PSH A		Push CCR onto stack
	LDA A	\$XXYY	
	STA A	\$PPQQ	Load low byte to AD7548
	LDAA	XXYY + 1	
	STA A	\$PPQQ+1	Load high byte to AD7548 and update analog output
	PULA		
	TAP		Pull CCR from stack
	PUL A		Pull A from stack
	RTS		Return to main program

AD7548 - 8085A INTERFACE

Figure 11 shows a typical AD7548 to 8085A microprocessor interface configured for automatic transfer of 8 + 4-bit right-justified data. Table IV gives a sample loading routine written in re-entrant form. The 12-bit data to be passed to the subroutine is stored in locations XXYY and XXYY + 1. The four most significant data bits occupy the lower half of XXYY + 1. As before, addresses PPQQ and PPQQ + 1 select the CSLSB and CSMSB/LDAC control inputs respectively. Since only two instructions (LHLD, SHLD) are required to both fetch and load the 12-bit data word to the AD7548, it may be more efficient to insert these instructions as required in the main program rather than use a subroutine such as illustrated here.

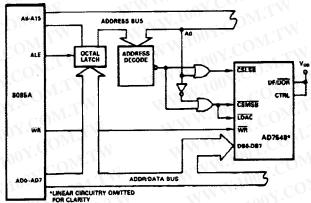


Figure 11. AD7548 - 8085A Interface (Automatic Transfer Mode)

Table IV. Sample Routine for AD7548-8085A Interface

	CALL	7548	
7548	PUSH	PSW	Push register contents onto stack
	PUSH LHLD SHLD POP	H XXYY PPQQ H	Fetch 12-bit data Load 12-bit data Pop register contents from stack
	POP	PSW	
	RET		Return to main program

AD7548

AD7548 - MC6809 INTERFACE

The AD7548 can be interfaced to the MC6809 microprocessor as shown in Figure 12 for automatic transfer of 8 + 4-bit data. Similar to the 8085A instructions LHLD and SHLD, the 6809 has two instructions to fetch and store 12-bit (16-bit) data to the AD7548, LDD and STD. However, in the 6809, the high byte of data is moved first, then the low byte (this is the opposite of the 8085A). This means that if the 12-bit data is assumed to reside at addresses XXYY and XXYY+1 then XXYY must contain the high byte. It also means that the address decoding logic of Figure 11 must be slightly changed so that the even-order

AD7548 address, PPQQ from before, selects the CSMSB input to load the high byte first. In this automatic transfer configuration LDAC is tied to the CSLSB input. The AD7548 analog output can thus be updated using only two instructions as follows:

LDD \$XXYY
STD \$PPQQ

The strobed transfer configuration is shown in Figure 13 with a dedicated decoder output assigned to each chip select input. The common LDAC signal allows simultaneous update of both AD7548 DAC registers.

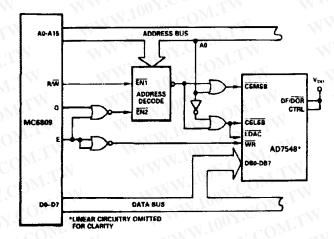


Figure 12. AD7548 - MC6809 Interface (Automatic Transfer Mode)

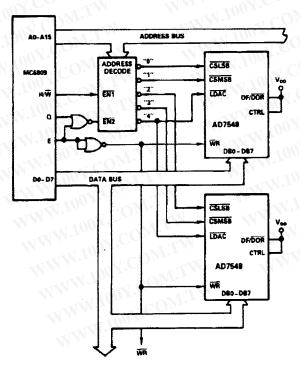


Figure 13. AD7548 - MC6809 Interface (Strobed Transfer Mode)

AD7548 - 6502 INTERFACE

Figure 14 shows a typical AD7548 to 6502 microprocessor interface configured for automatic transfer of right-justified data. As a programming example, Figure 15 shows a flow chart for producing a 12-bit (4095-step-max) voltage ramp under 6502 control. Index registers X and Y of the 6502 form a 12-bit counter with the X-register holding the low byte of data and the Y-register the high byte. Table V shows the program listing. The X-register is compared with FF_H and the Y-register with 10_H to determine when the ramp voltage has reached its maximum value (FFF_H). By changing the comparison data in the program the maximum ramp output voltage can be varied from levels corresponding to FFF_H down to 000_H. In the program listing of Table V the AD7548 has been assigned contiguous addresses 0400 (low byte) and 0401 (high byte and DAC register).

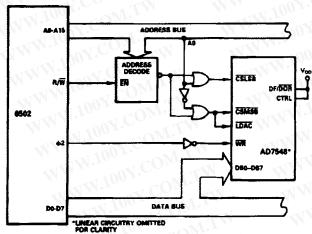


Figure 14. AD7548 - 6502 Interface (Automatic Transfer Mode)

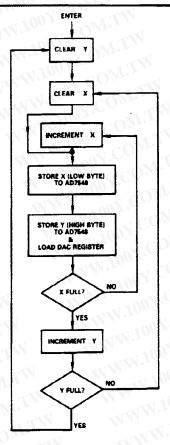


Figure 15. Flow Chart for Voltage Ramp Generation

Table V. Program Listing for Figure 15

ADDRESS	OP-CODE	MNEMONIC	OPERAND
0000	A0	LDY	# 00
01	00	WW	OY.CO
02	A2	LDX	# 00
03	00		00^{1} .
04	4C	JMP	0008
05	08	111	700 r.
06	00	MM,	. OOY.C.
07	E8	INX	1.700
08	8E	STX	0400
09	00	-3130	W.T.
OA (O)	04	// /	100 r.
OB	8C	STY	0401
0C	01		M.100
0D	04	W W	
0E	E0	CPX	# FF
0F	FF	I.M.	,
10	D0	BNE	0007
11	F5		
12	C8	INY	
13	C0	CPY	# 10
14	10	1	
15	D0	BNE	0002
16	EB		
17	FO	BEQ	0000
0018	E7		

AD7548 - Z80 INTERFACE

Figure 16 shows a typical AD7548 to Z80 microprocessor interface configured for automatic transfer of right-justified data. Similar to the 8085A and 6809 cases, 16-bit load instructions are available in the Z80 which can fetch and load 12-bit data to the AD7548. Since the low byte of data is moved first and assuming the 12bit data resides at addresses XXYY and XXYY + 1, address XXYY must contain the low byte. As before, addresses PPQQ and PPQQ + 1 select the AD7548 CSLSB and CSMSB/LDAC control inputs respectively. Choosing the Z80 register pair BC to hold the 12-bit data, the two instructions required to update the AD7548 analog output are as follows:

> LD BC, (XXYY) LD (PPQQ), BC

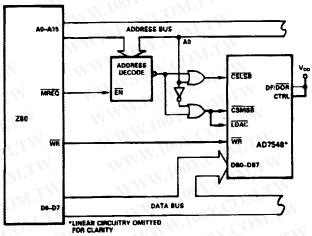
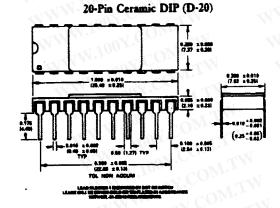


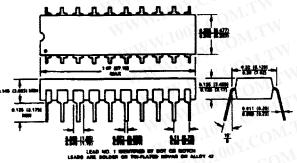
Figure 16. AD7548 - Z80 Interface (Automatic Transfer Mode)

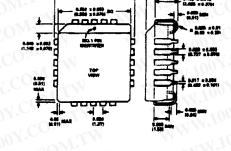
MECHANICAL INFORMATION **OUTLINE DIMENSIONS**

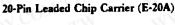
20-Pin Plastic Leaded Chip Carrier (P-20A)

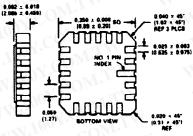


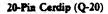


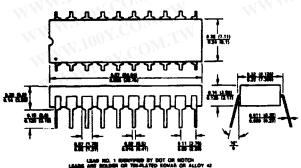












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