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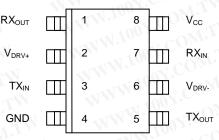
# DS276 Low Power Transceiver Chip

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# FEATURES

- Low-power serial transmitter/receiver for battery-backed systems
- Transmitter steals power from receive signal line to save power
- Single 3V or 5V operation
- Full duplex operation up to 20k bps
- Ultra-low static current
- Compatible with RS-232-E signals

# PIN ASSIGNMENT



DS276 8-Pin DIP (300-mil) DS 276S 8-Pin SOIC (150-mil)

# **PIN DESCRIPTION**

RX <sub>OUT</sub>	RS-232 Receiver Output
V <sub>DRV+</sub>	Transmit Driver Positive Supply
TX <sub>IN</sub>	RS-232 Driver Input
GND	System Ground (0V)
TX <sub>OUT</sub>	RS-232 Driver Output
V <sub>DRV-</sub>	Transmit Driver Negative Supply
RX <sub>IN</sub>	RS-232 Receiver Input
V <sub>CC</sub>	System Logic Supply (3-5V)

# **ORDERING INFORMATION**

DS276	8-Pin DIP
DS276S	8-Pin SOIC

# DESCRIPTION

The DS276 Line-Powered RS-232 Transceiver Chip is a CMOS device that provides a low-cost, very low-power interface to RS-232 serial ports. The receiver input translates RS-232 signal levels to common CMOS/TTL levels. The transmitter can be used with independently supplied positive and negative supplies, but in most cases will be used with the positive supply, sharing the logic supply and the negative supply stolen from the receive RS-232 signal when that signal is in a negative state (marking). By using an external reservoir capacitor and Schottky diode (see Figure 4) this negative supply can be maintained even during full-duplex operation. Since most serial communication ports remain in a negative state statically, using the receive signal for negative power greatly reduces the DS276's static power consumption. This feature is especially important for battery-powered systems such as laptop computers, remote sensors, and portable medical instruments. During an actual communication session, the DS276's transmitter will use system power (3-12 volts) for positive transitions while still employing the receive signal for negative transitions.

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# OPERATION

DS276

Designed for the unique requirements of battery-backed systems, the DS276 provides a low-power full-or half-duplex interface to an RS-232 serial port. Typically, a designer must use an RS-232 device which uses system power during both negative and positive transitions of the transmit signal to the RS-232 port. If the connector to the RS-232 port is left connected for an appreciable time after the communication session has ended, power will statically flow into that port, draining the battery capacity. The DS276 eliminates this static current drain by stealing current from the receive line (RX<sub>IN</sub>) of the RS-232 port when that line is at a negative level (marking). Since most asynchronous communication over an RS-232 connection typically remains in a marking state when data is not being sent, the DS276 will not consume system power in this condition. Sys-tem power would only be used when positive-going transitions are needed on the transmit RS-232 output (TX<sub>OUT</sub>) when data is sent. However, since synchronous communication sessions typically exhibit a very low duty-cycle, overall system power consumption remains low.

# **RECEIVER SECTION**

The RX<sub>IN</sub> pin is the receive input for an RS-232 signal whose levels can range from ±3 to ±15 volts. A negative data signal is called a mark while a positive data signal is called a space. These signals are inverted and then level-shifted to normal +3 or +5 volt CMOS/TTL logic levels. The logic output associated with RX<sub>IN</sub> is RX<sub>OUT</sub> which swings from V<sub>CC</sub> to ground. Therefore, a mark on RX<sub>IN</sub> produces a logic 1 at RX<sub>OUT</sub>; a space produces a logic 0.

The input threshold of  $RX_{IN}$  is typically around 1.8 volts with 500 millivolts of hysteresis to improve noise rejection. Therefore, an input positive-going signal must exceed 1.8 volts to cause  $RX_{OUT}$  to switch states. A negative-going signal must now be lower than 1.3 volts (typically) to cause  $RX_{OUT}$  to switch again. An open on  $RX_{IN}$  is interpreted as a mark, producing a logic 1 at  $RX_{OUT}$ .

# TRANSMITTER SECTION

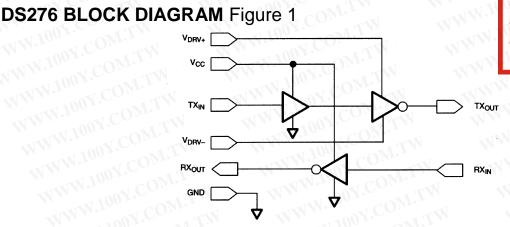
 $TX_{IN}$  is the CMOS/TTL-compatible input for data from the user system. A logic 1 at  $TX_{IN}$  produces a mark (negative data signal) at  $TX_{OUT}$  while a logic 0 produces a space (positive data signal). As mentioned earlier, the transmitter section employs a unique driver design that can use the  $RX_{IN}$  line for swinging to negative levels.  $RX_{IN}$  can be connected via external circuitry to  $V_{DRV}$  to allow stored charge to supply this voltage during marking (or idle) states. When  $TX_{OUT}$  needs to transition to a positive level, it uses the  $V_{DRV+}$  power pin for this level.  $V_{DRV+}$  can be a voltage supply between 3 to 12 volts, and in many situations it can be tied directly to the  $V_{CC}$  supply. It is important to note that  $V_{DRV+}$  must be greater than or equal to  $V_{CC}$  at all times.

The voltage range on  $V_{DRV+}$  permits the use of a 9V battery in order to provide a higher voltage level when  $TX_{OUT}$  is in a space state. When  $V_{CC}$  is shut off to the DS276 and  $V_{DRV+}$  is still powered (as might happen in a battery-backed condition), only a small leakage current (about 50-100 nA) will be drawn. If  $TX_{OUT}$  is loaded during such a condition,  $V_{DRV+}$  will draw current only if  $RX_{IN}$  is not in a negative state. During normal operation ( $V_{CC} = 3$  or 5 volts),  $V_{DRV+}$  will draw less than 2 uA when  $TX_{OUT}$  is marking. Of course, when  $TX_{OUT}$  is spacing,  $V_{DRV+}$  will draw substantially more current—about 3 mA, depending upon its voltage and the impedance that  $TX_{OUT}$  sees. The  $TX_{OUT}$  output is slew rate-limited to less than 30 volts/us in accordance with RS-232 specifications. In the event  $TX_{OUT}$  should be inadvertently shorted to ground, internal current-limiting circuitry prevents damage, even if continuously shorted.

# **RS-232 COMPATIBILITY**

The intent of the DS276 is not so much to meet all the requirements of the RS-232 specification as to offer a low-power solution that will work with most RS-232 ports with a connector length of less than 10

feet. As a prime example, the DS276 will not meet the RS-232 requirement that the signal levels be at least  $\pm 5$  volts minimum when terminated by a 3 k $\Omega$  load and V<sub>DRV+</sub> = +3-5 volts. Typically 2.5 to 4 volts will be present at TX<sub>OUT</sub> when spacing under this condition, depending on the supply voltage. However, since most RS-232 receivers will correctly interpret any voltage over 2 volts as a space, there will be no problem transmitting data.



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# **APPLICATIONS INFORMATION**

The DS276 is designed as a low-cost, RS-232-E interface expressly tailored for the unique requirements of battery-operated handheld products. As shown in the electrical specifications, the DS276 draws exceptionally low operating and static current. During normal operation when data from the handheld system is sent from the  $TX_{OUT}$  output, the DS276 only draws significant  $V_{DRV+}$  current when  $TX_{OUT}$  transitions positively (spacing). This current flows primarily into the RS-232 receiver's 3-7 k $\Omega$  load at the other end of the attaching cable. When  $TX_{OUT}$  is marking (a negative data signal), the  $V_{DRV+}$  current falls dramatically since the negative voltage is provided by the transmit signal from the other end of the cable. This represents a large reduction in overall operating current, since typical RS-232 interface chips use charge-pump circuits to establish both positive and negative levels at the transmit driver output. To obtain the lowest power consumption from the DS276, observe the following guidelines: First, to minimize  $V_{DRV+}$  current when connected to an RS-232 port, always maintain  $TX_{IN}$  at a logic 1 when data is not being transmitted (idle state). This will force  $TX_{OUT}$  into the marking state, minimizing  $V_{DRV+}$  current will drop significantly when  $V_{CC}$  is grounded. Therefore, if  $V_{DRV+}$  is derived independently from  $V_{CC}$  (for example connected to a 9V battery), the logic supply voltage can be turned off to achieve the lowest possible power state.

# FULL-DUPLEX OPERATION

The DS276 is intended for full-duplex operation using the full-duplex circuit shown in Figure 4 to generate a negative rail from  $RX_{IN}$ . The 22  $\mu$ F capacitor forms a negative-charge reservoir; consequently, when the TXD line  $RX_{IN}$  is spacing (positive),  $TX_{OUT}$  still has a negative source available for a time period determined by the capacitor and the load resistance at the other end (3-7 k $\Omega$ ).

# SUPPLY VOLTAGE OPTIONS

The DS276 is intended primarily for use in single supply 3- or 5- volts systems. However, several supply configurations are possible.

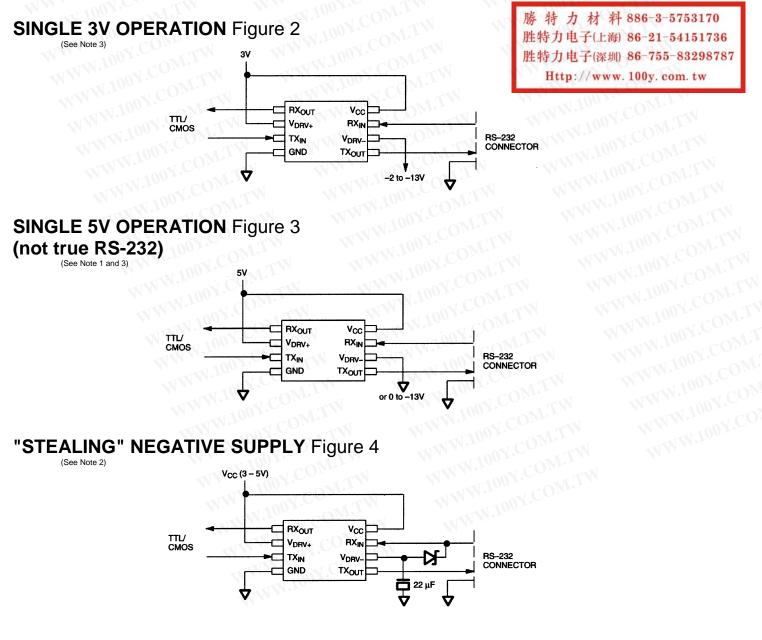
# **3V OPERATION**

The simplest configuration is to use a single 3V supply for  $V_{CC}$  and  $V_{drv+}$ , and connect  $V_{drv-}$  to ground. This will result in the lowest power consumption and will give adequate serial communication between two similar devices over short distances, and into larger loads than the 3 k $\Omega$  RS-232 standard (Figure 2). If V<sub>drv+</sub> is increased to 5V, and V<sub>drv-</sub> decreased (to less than -2V) communication with standard RS-232 devices is possible, although of course the output voltage swing of the DS276 remains below the RS-232 specification. The V<sub>drv-</sub> supply can be derived using the "stealing" technique shown in Figure 4.

# **5V OPERATION**

The use of a single 5V supply for  $V_{CC}$  and  $V_{drv+}$ , and  $V_{drv-}$  derived using the circuit in Figure 4 can produce reliable communication with standard RS-232 devices, although the DS276 output voltage swings are below the RS-232 minimum (Figure 3).

Increasing the magnitude of the voltage to  $V_{drv+}$  to 10 volts or more will result in "true" RS-232 output voltage levels.



# NOTES:

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 This circuit as shown does not meet the RS-232 requirement for signal levels (high-level output voltage). However, as most RS-232 receivers will interpret any voltage over 2V as a space this will normally be of no consequence. Alternatively, V<sub>DRV+</sub> can be supplied independently from a higher voltage supply.

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- 2. The capacitor is charged negatively whenever  $RX_{IN}$  is in a marking (or idle) state. When the DS276 is transmitting marking data and  $RX_{IN}$  is spacing the capacitor will discharge towards ground with a time constant determined by the capacitor value and the value of the load resistance. The value shown should store sufficient charge for reliable operation up to 20 kbps.
- 3.  $RX_{IN}$  must never be allowed to reach a negative voltage with respect to  $V_{DRV}$  or excessive currents will be drawn. Therefore, if negative voltage swings are present on  $RX_{IN}$ ,  $V_{DRV}$  should not be connected to ground and the circuit shown in Figure 4 should be used.

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# ABSOLUTE MAXIMUM RATINGS\*

 $V_{DR+}$ 

V<sub>DR-</sub>

TX<sub>IN</sub>

TX<sub>OUT</sub> RX<sub>OUT</sub>

Operating Temperature Storage Temperature Soldering Temperature

RX<sub>IN</sub>

-0.3V to +7.0V
-0.3V to +13V
-13V to +0.3V
-15V to +15V
-0.3V to VCC+0.3V
-15V to +15V
-0.3V to VCC+0.3V
0°C to 70°C
-55°C to +125°C
260°C for 10 seconds

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\* This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operation sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods of time may affect reliability.

PARAMETER	SYMBOL	CONDITION	MIN	TYP	MAX	UNITS	NOTES
Logic Supply Voltage	V <sub>CC</sub>	WWW.100N.	2.7	3-5	5.5	V	CON1
Transmit Driver Supply	V <sub>DR+</sub> V <sub>DR+</sub>	V <sub>CC</sub> =5V±10% V <sub>CC</sub> 2.7-3.6V	V <sub>CC</sub> V <sub>CC</sub>	5-12 3-5	13 5.5	V V	LCO1 N.COM
Transmit Driver Supply	V <sub>DR-</sub>	WW.100	-15	Wn	0	V	OD.Yo
High-level Input Voltage	V <sub>IH</sub>	1.WWW.IO	20	M.I.	V <sub>CC</sub> +0.3	V	100X.CON
Low-level Input Voltage	V <sub>IL</sub>	WW	-0.3	OW.1	0.8	V	.100 L
RX <sub>IN</sub> Input Voltage	V <sub>RS</sub>		-15	COM	+15	V	N.101

PARAMETER	SYMBOL		MIN	TYP	MAX	UNITS	NOTES
Logic Supply Voltage	V <sub>CC</sub>	W WILL	2.7	107.CC	3.6	V V	10
Dynamic Supply Current	V.100Y.CO	WIN	N.N.N.	100Y.C	TIM		WW I
	I <sub>DRV1</sub>	$TX_{IN} = V_{CC}$	WWW.	400	800	🔨 uA	2
	I <sub>CC1</sub>	$TX_{IN} = V_{CC}$	WWW	40	100	uA	2
	I <sub>DRV1</sub>	$TX_{IN} = GND$	W	3.8	5	mA	2 2
	I <sub>CC1</sub>	$TX_{IN} = GND$		40	100	uA	2
Static Supply Current	WW.100	COMITY	N I	NN.10		$W_{III}$	
	I <sub>DRV2</sub>	$TX_{IN} = V_{CC}$	N	1.5	10	uA	3
	I <sub>CC2</sub>	$TX_{IN} = V_{CC}$		10	15	uA	3
	I <sub>DRV2</sub>	$TX_{IN} = GND$		3.8	5	mA	3
	I <sub>CC2</sub>	$TX_{IN} = GND$		10	20	uA	3
Driver Leakage Current	I <sub>DRV3</sub>	$V_{\rm CC} = 0$		0.05	1	uA	4
		$V_{DRV+}=V_{CC}=$ 2.7V					
TX <sub>OUT</sub> Level High	V <sub>OTXH</sub>	$V_{DRV}=0$	2	2.4		V	5
		$V_{DRV+}=4.5V,$					
		$V_{DRV}$ =-12V	3.8	4		V	6

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ELECTRICAL CHARA PARAMETER	SYMBOL	WT	MIN	TYP	MAX	$A = 0^{\circ}C$ UNITS	NOTES
TX <sub>OUT</sub> Level Low	V <sub>OTXL</sub>	$V_{DRV}$ += $V_{CC}$ =2.7V $V_{DRV}$ =0	WW WW	0.2	0.3	v	5
	WW.10	or. COM.TW	N N	-11	-10	O V	6
TX <sub>OUT</sub> Short Circuit Current	I <sub>SC</sub>	$V_{DRV+} = 5.5V,$ $V_{DRV-} = -12V$	-1	WWW	85	mA	7
TX <sub>OUT</sub> Output Slew Rate	t <sub>SR</sub>	100X. COM.T		N.	30	V/us	
Propagation Delay	t <sub>PD</sub>	N.100Y. COM.		5	NN.100	us	8
RX <sub>IN</sub> Input Threshold Low	V <sub>TL</sub>	W.100Y.COM	0.8	1.0	1.6	V	
RX <sub>IN</sub> Input Threshold High	V <sub>TH</sub>	WW.100X.CON	1.2	2.0	2.4	v	M.I.W
RX <sub>IN</sub> Threshold Hysteresis	V <sub>HYS</sub>	1W 100Y.CC	0.4	1.0	WW	V	9
RX <sub>OUT</sub> Output Current High	I <sub>OH</sub>	$V_{CC} = 2.7V$ $V_{OH} = 2V$	OM.T	4	-0.5	mA	COM.TY
RX <sub>OUT</sub> Output Current Low	I <sub>OL</sub>	$V_{CC} = 2.7V$ $V_{OL} = 0.4V$	0.5	TW	M	mA	N.COM

# NOTES:

- 1.  $V_{DRV+}$  must be greater than or equal to  $V_{CC}$ ,  $RX_{IN}$  must be greater than  $V_{DRV-}$ .
- See test circuit in Figure 5. 2.
- 3. See test circuit in Figure 6.
- 4. See test circuit in Figure 7.
- 5.  $R_L = 3k\Omega$  to ground. Max data rate = 20k bps.
- 6.  $R_L = 3k\Omega$  to ground. Max data rate = 50k bps.
- 7.  $TX_{IN} = V_{IL}$ .
- WWW.100Y.COM.TW 8. See test circuit in Figure 8. WWW.100Y.COM.TW
- 9.  $V_{HYS} = V_{TH} V_{TL}$ .

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$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	ELECTRICAL CHARA	CTERIST	ICS-5V OPE	<b>RATION</b> $(t_A = 0^{\circ}C \text{ to } 70^{\circ}C)$					
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	PARAMETER	SYMBOL	WTN	MIN	TYP	MAX	UNITS	NOTES	
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	Logic Supply Voltage	V <sub>CC</sub>	WT	4.5	5	5.5	V		
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	Dynamic Supply Current	VWW.100Y	COMTW	WY	11.	oy.cu	WILL	-	
$\begin{array}{ c c c c c c c c c c c c c c c c c c c$	N. LUU T. COM. TW	I <sub>DRV1</sub>	$TX_{IN} = V_{CC}$	N			uA	1	
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$		I <sub>CC1</sub>		4				1	
$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$				l				1	
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	The second secon	I <sub>CC1</sub>	$TX_{IN} = GND$	<u> </u>	40	100	uA	1	
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	Static Supply Current	WWW-	100 r. COM.		NN N	11.100	N.COM.	No.	
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$							-1		
$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$				TY					
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $				T.T.		-11		-	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	Driver Leakage Current			M.TW			1001		
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	WIN 1002 COM.T	-DRIV		DW.T.W	1	WW	N.100	ION IN	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $			4.5V	OM.1		1	W.100 -	COM'1	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	TX <sub>OUT</sub> Level High	V <sub>OTXH</sub>	$V_{DRV}=0$	3.3	3.8		V	4	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $		WT.N		LCOM	LM.		WW.100	COM.1	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	WWW.100X.CO.	WTM		10	<u>11</u>	V	V	5	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $		WT.M		PY.CC	MT.M	1	N.V.	IOY.COM	
$\begin{array}{ c c c c c c c c c c c c c c c c c c c$	WWW. 100Y.C	WLIN		DOX.CC	MIN	10	V	1001.00	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	TX <sub>OUT</sub> Level Low	V <sub>OTXL</sub>		100×.C	-11	-10	V V	4.00	
TX_{OUT} Short Circuit CurrentIsc $V_{DRV+} = 12V$ , $V_{DRV-} = -12V$ 85mA6TX_{OUT} Output Slew Rate $t_{SR}$ 30V/usPropagation Delay $t_{PD}$ 5us7RX_{IN} Input Threshold Low $V_{TL}$ 0.81.26VRX_{IN} Input Threshold High $V_{TH}$ 1.622VRX_{IN} Threshold Hysteresis $V_{HYS}$ 0.50.8V8RX_{OUT} Output Current High $I_{OH}$ $V_{CC} = 4.5V$ $V_{OH} = 2.4V$ -1mARX_{OUT} Output Current Low $I_{OL}$ $V_{CC} = 4.5V$ -1mA		COM.TW		1007.0	0.2	0.3	v	N.105	
$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	TX <sub>OUT</sub> Short Circuit Current	I <sub>SC</sub>		N.1001	COM	TN		W.100X-	
Propagation Delay $t_{PD}$ 5us7RX_{IN} Input Threshold Low $V_{TL}$ 0.81.26VRX_{IN} Input Threshold High $V_{TH}$ 1.622VRX_{IN} Threshold Hysteresis $V_{HYS}$ 0.50.8V8RX_{OUT} Output Current HighI_{OH} $V_{CC} = 4.5V$ $V_{OH} = 2.4V$ -1mARX_{OUT} Output Current LowI_{OL} $V_{CC} = 4.5V$ -1mA	WWW	OV.CO.		N.100		85	mA	6	
RX_{IN} Input Threshold Low $V_{TL}$ 0.81.26VRX_{IN} Input Threshold High $V_{TH}$ 1.622VRX_{IN} Threshold Hysteresis $V_{HYS}$ 0.50.8V8RX_{OUT} Output Current HighI_{OH} $V_{CC} = 4.5V$ $V_{OH} = 2.4V$ -1mARX_{OUT} Output Current LowI_{OL} $V_{CC} = 4.5V$ -1mA	TX <sub>OUT</sub> Output Slew Rate	t <sub>SR</sub>	Les MI	W.10	N.C.C	30	V/us	W.100	
RX IN Input Threshold High $V_{TH}$ 1.622VRX IN Threshold Hysteresis $V_{HYS}$ 0.50.8V8RX_{OUT} Output Current HighI_{OH} $V_{CC} = 4.5V$ $V_{OH} = 2.4V$ -1mARX_{OUT} Output Current LowI_{OL} $V_{CC} = 4.5V$ $V_{CC} = 4.5V$ -1mA	Propagation Delay	t <sub>PD</sub>	IN WI.	.Www.1	5	OM.TW	us	7.100	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	RX <sub>IN</sub> Input Threshold Low	V <sub>TL</sub>	VI.TW V	0.8	1.2	6	V	N.W.IC	
RX_{OUT} Output Current High $I_{OH}$ $V_{CC} = 4.5V$ $-1$ mARX_{OUT} Output Current Low $I_{OL}$ $V_{CC} = 4.5V$ $-1$ mA	RX <sub>IN</sub> Input Threshold High	V <sub>TH</sub>	MI.IW	1.6	2	2	V	.WW.	
$V_{OH} = 2.4V \qquad -1 \qquad \text{mA}$ $RX_{OUT} \text{ Output Current Low} \qquad I_{OL} \qquad V_{CC} = 4.5V \qquad \qquad$	RX <sub>IN</sub> Threshold Hysteresis	V <sub>HYS</sub>	ONI.TW	0.5	0.8	COM	V	8	
$RX_{OUT} \text{ Output Current Low } I_{OL} \qquad V_{CC} = 4.5V \qquad \qquad$	RX <sub>OUT</sub> Output Current High	I <sub>OH</sub>	00	NY	W.100	N.CON	mA	NA .	
	RX <sub>OUT</sub> Output Current Low	I <sub>OL</sub>		3	WW.1	DOX.CO	mA		

# NOTES:

- 1. See test circuit in Figure 9.
- 2. See test circuit in Figure 10.
- 3. See test circuit in Figure 11.
- 4.  $R_L = 3 k\Omega$  to ground. Max data rate = 20 kbps.
- 5.  $R_L = 3 k\Omega$  to ground. Max data rate = 100 kbps.

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WW.100Y.COM.TW 6.  $TX_{IN} = V_{IL}$ .

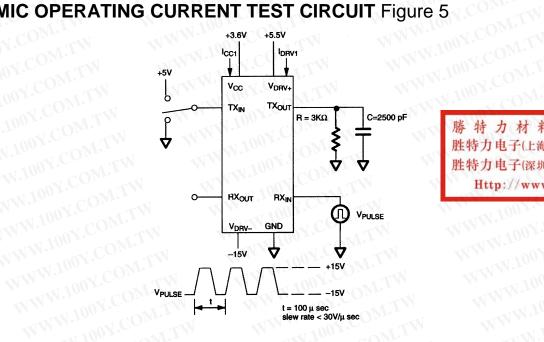
COMTW

- 7. See test circuit in Figure 12.
- 8.  $V_{HYS} = V_{TH} V_{TL}$ .

# 00X.COM.TW **TEST CIRCUIT** Figure 5 DYNAMIC OPERATING CURRENT

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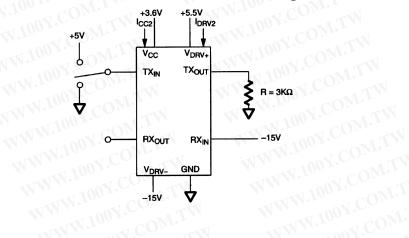
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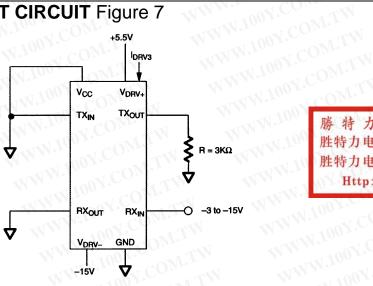
勝特力材料 886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787 Http://www. 100y. com. tw

# STATIC OPERATING CURRENT TEST CIRCUIT Figure 6



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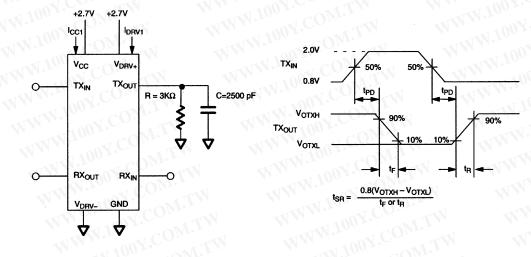
# DRIVER LEAKAGE TEST CIRCUIT Figure 7



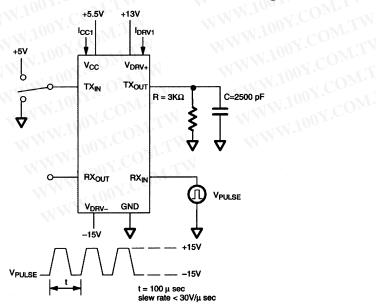
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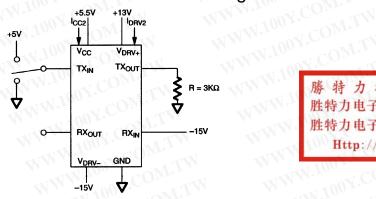
# **PROPAGATION DELAY TEST CIRCUIT** Figure 8



# **DYNAMIC OPERATING CURRENT TEST CIRCUIT** Figure 9



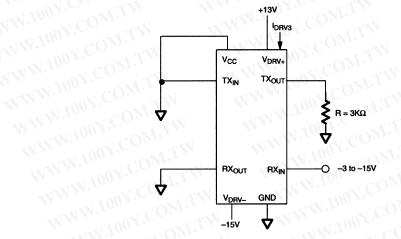
# WW.100Y.COM.TW STATIC OPERATING CURRENT TEST CIRCUIT Figure 10 COM. WWW.II



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# DRIVER LEAKAGE TEST CIRCUIT Figure 11



# **PROPAGATION DELAY TEST CIRCUIT** Figure 12

