

DATA SHEET

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For a complete data sheet, please also download:

- The IC04 LOCMOS HE4000B Logic Family Specifications HEF, HEC
- The IC04 LOCMOS HE4000B Logic Package Outlines/Information HEF, HEC

HEF4541B **MSI** **Programmable timer**

Product specification
File under Integrated Circuits, IC04

January 1995

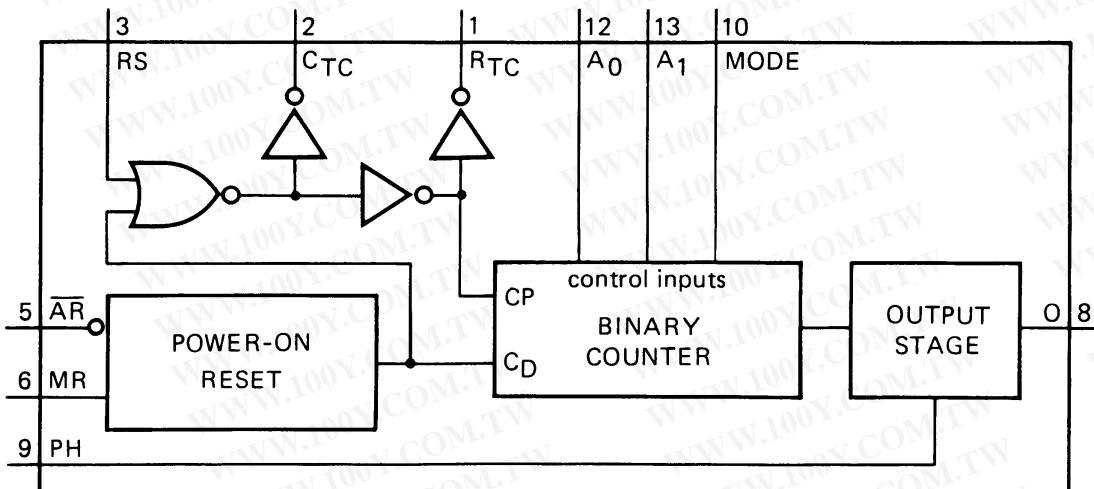
Programmable timer

DESCRIPTION

The HEF4541B is a programmable timer which consists of a 16-stage binary counter, an integrated oscillator to be used with external timing components, an automatic power-on reset and output control logic. The frequency of the oscillator is determined by the external components R_t and C_t within the frequency range 1 Hz to 100 kHz. This oscillator may be replaced by an external clock signal at input RS, the timer advances on the positive-going transition of RS. A LOW on the auto reset input (AR) and a LOW on the master reset input (MR) enables the internal power-on reset. A HIGH level at input MR resets the counter independent on all other inputs. Resetting

disables the oscillator to provide no active power dissipation.

A HIGH at input \overline{AR} turns off the power-on reset to provide a low quiescent power dissipation of the timer. The 16-stage counter divides the oscillator frequency by 2^8 , 2^{10} , 2^{13} or 2^{16} depending on the state of the address inputs (A_0 , A_1). The divided oscillator frequency is available at output O. The phase input (PH) features a complementary output signal. If the mode select input (MODE) is LOW or HIGH the timer can be used respectively as a single transition timer or 2^n frequency divider.



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Fig.1 Functional diagram.

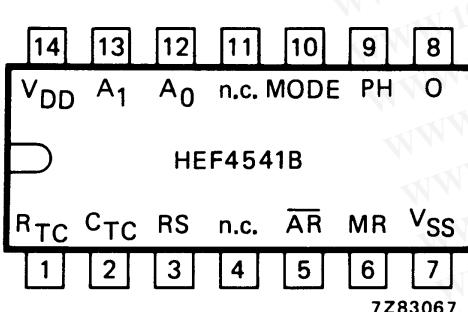


Fig.2 Pinning diagram.

HEF4541BP(N): 14-lead DIL; plastic
 (SOT27-1)

HEF4541BD(F): 14-lead DIL; ceramic (cerdip)
 (SOT73)

HEF4541BT(D): 14-lead SO; plastic
 (SOT108-1)

(): Package Designator North America

FAMILY DATA, I_{DD} LIMITS category MSI

See Family Specifications

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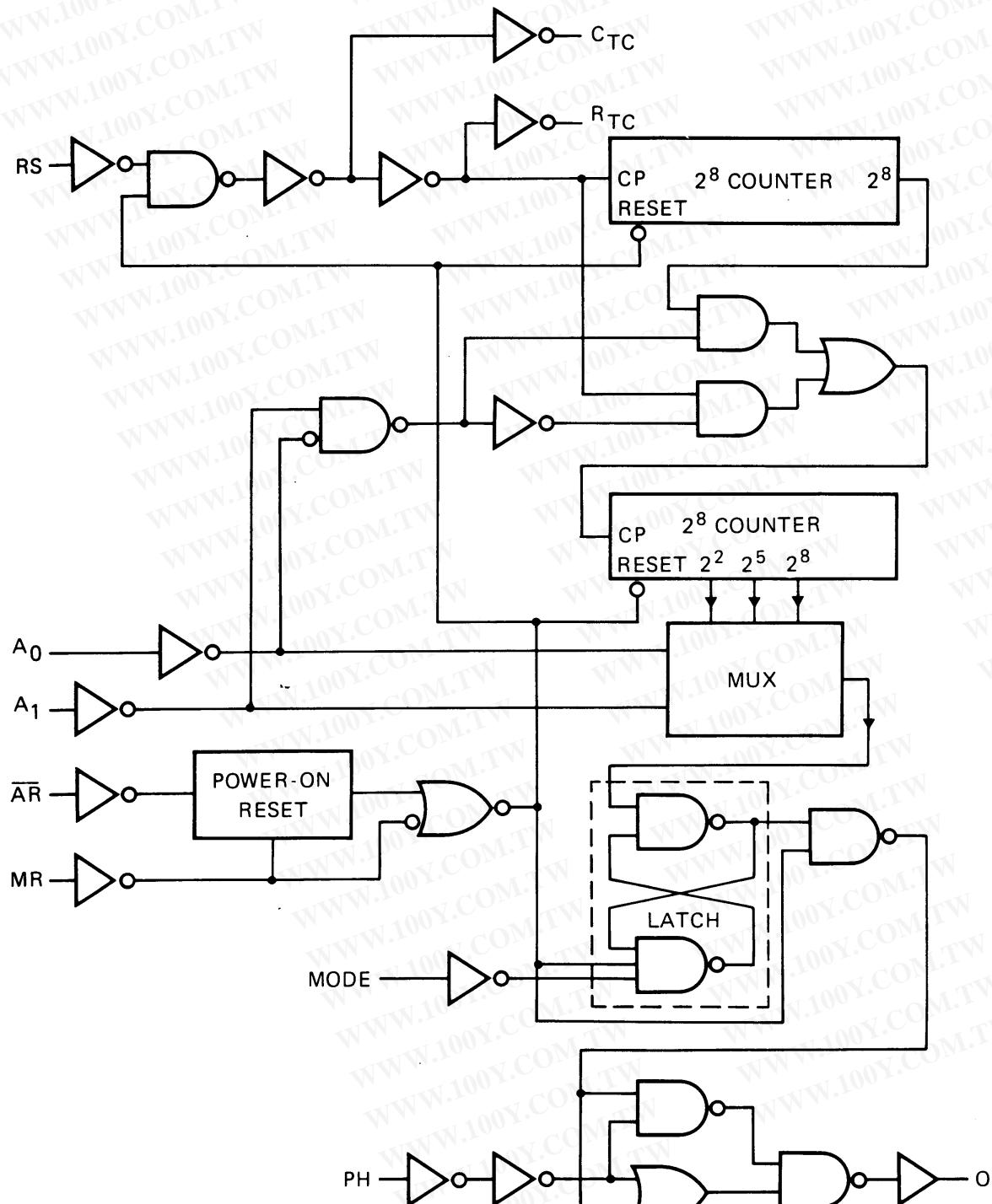


Fig.3 Logic diagram.

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PINNING

A ₀ , A ₁	address inputs
MODE	mode select input
AR	auto reset input
MR	master reset input
PH	phase input
R _{TC}	external resistor connection (R _t)
C _{TC}	external capacitor connection (C _t)
RS	external resistor connection (R _S) or external clock input

FREQUENCY SELECTION TABLE

A ₀	A ₁	NUMBER OF COUNTER STAGES n	$\frac{f_{osc}}{f_{out}} = 2^n$
L	L	13	8 192
L	H	10	1 024
H	L	8	256
H	H	16	65 536

FUNCTION TABLE

INPUTS				MODE
AR	MR	PH	MODE	
H	L	X	X	auto reset disabled
L	L	X	X	auto reset enabled (1)
X	H	X	X	master reset active
X	L	X	H	normal operation selected
X	L	X	L	division to output
X	L	L	X	single-cycle mode (2)
X	L	H	X	output initially LOW, after reset
				output initially HIGH, after reset

Notes

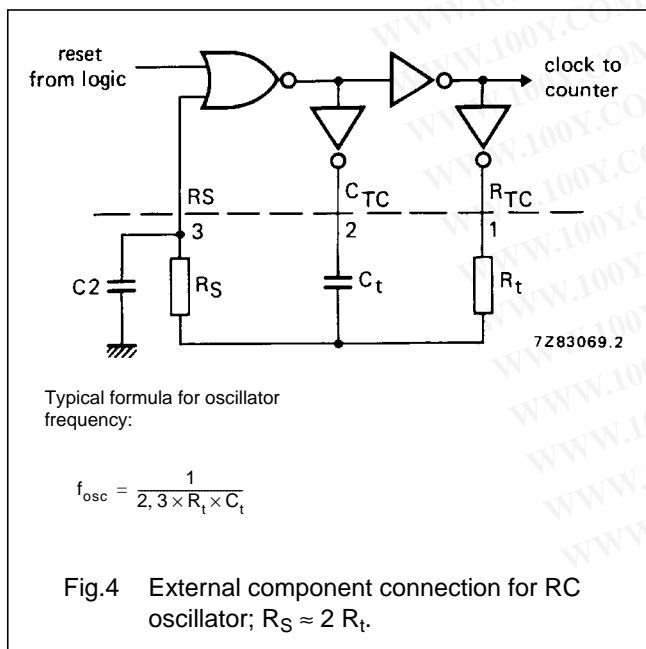
- For correct power-on reset, the supply voltage should be above 8.5 V. For V_{DD} < 8.5 V, disable the autoreset and connect AR to V_{DD}.
- The timer is initialized on a reset pulse and the output changes state after 2ⁿ⁻¹ counts and remains in that state (latched). Reset of this latch is obtained by master reset or by a LOW to HIGH transition on the MODE input.

H = HIGH state (the more positive voltage)

L = LOW state (the less positive voltage)

X = state is immaterial

RC oscillator



Timing component limitations

The oscillator frequency is mainly determined by R_tC_t, provided R_t << R_S and R_SC₂ << R_tC_t. The function of R_S is to minimize the influence of the forward voltage across the input protection diodes on the frequency. The stray capacitance C₂ should be kept as small as possible. In consideration of accuracy, C_t must be larger than the inherent stray capacitance. R_t must be larger than the LOC莫斯 'ON' resistance in series with it, which typically is 500 Ω at V_{DD} = 5 V, 300 Ω at V_{DD} = 10 V and 200 Ω at V_{DD} = 15 V.

The recommended values for these components to maintain agreement with the typical oscillation formula are:

C_t ≥ 100 pF, up to any typical value,
 10 kΩ ≤ R_t ≤ 1 MΩ.

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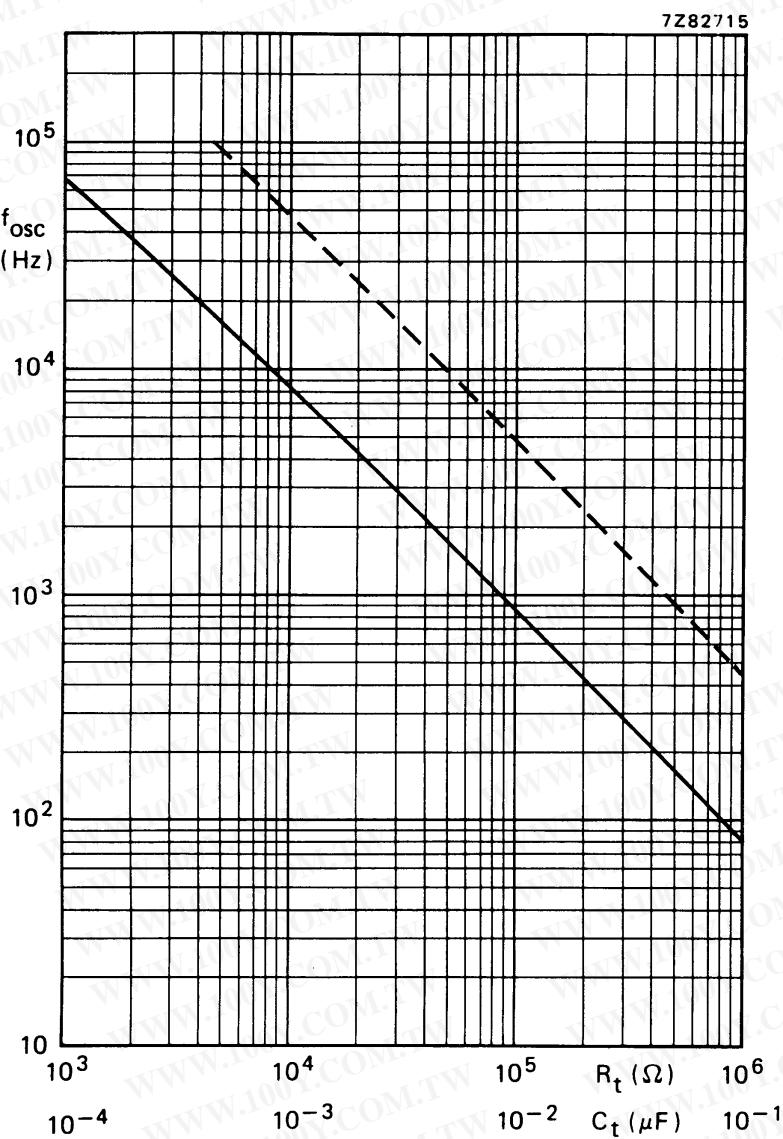


Fig.5 RC oscillator frequency as a function of R_t and C_t at $V_{DD} = 5$ to 15 V; $T_{amb} = 25^\circ\text{C}$.

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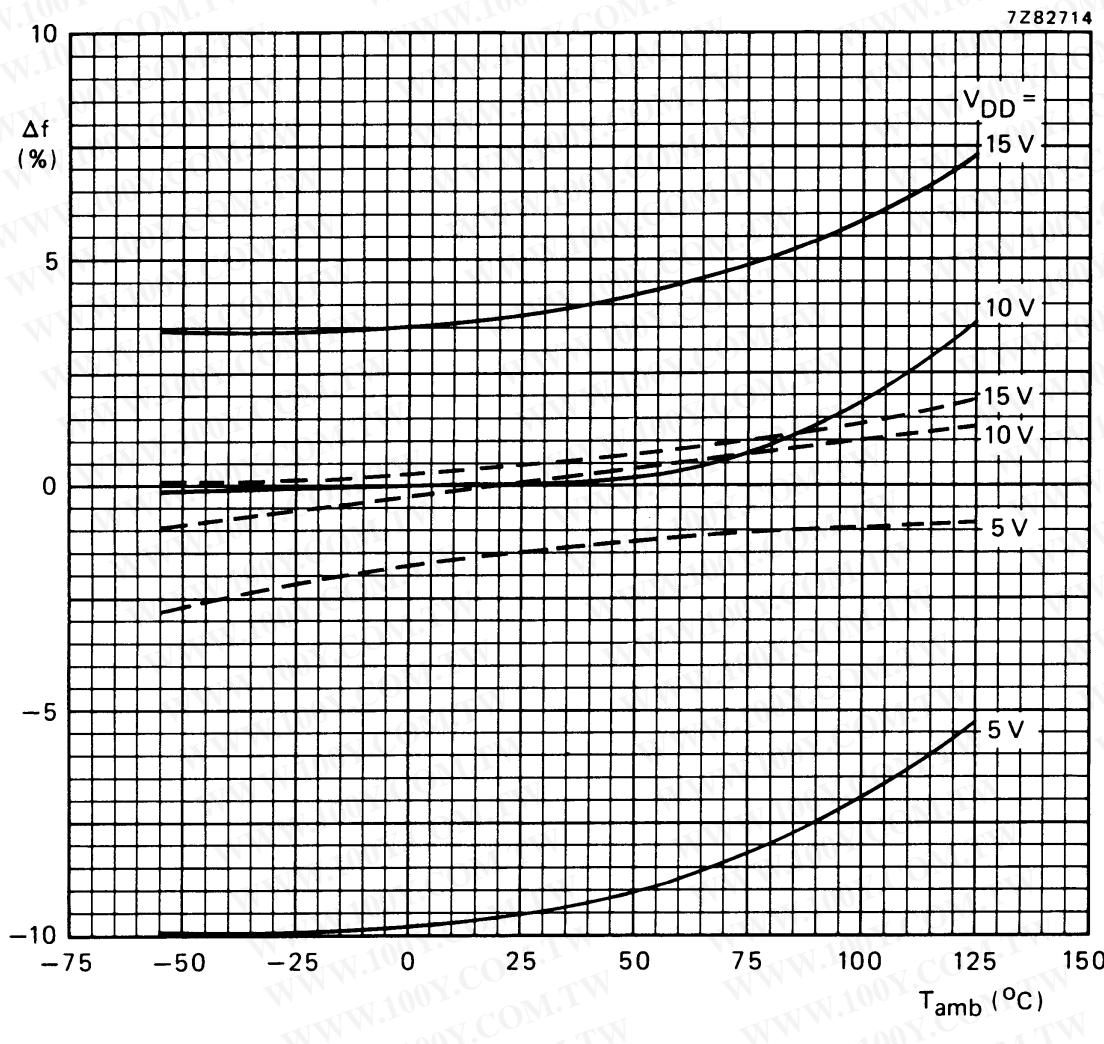


Fig.6 Frequency deviation (Δf) as a function of ambient temperature; referenced at : f_{osc} at $T_{amb} = 25^\circ\text{C}$ and $V_{DD} = 10 \text{ V}$.

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DC CHARACTERISTICS

 $V_{SS} = 0 \text{ V}$

	V_{DD} V	V_{OL} V	V_{OH} V	SYMBOL	T_{amb} ($^{\circ}\text{C}$)						
					-40	+25	+85	MIN.	MAX.	MIN.	MAX.
Supply current power-on reset enabled (note)	5			I_D	-	80	-	20	80	-	230 μA
	10				-	750	-	250	600	-	700 μA
	15				-	1600	-	500	1300	-	1500 μA
Supply voltage for automatic reset initialization (note)				V_{DD}	-	-	8,5	5	-	-	- V
Output current HIGH; C_{TC} , R_{TC}	5		4,6	$-I_{OH}$	0,5	-	0,4	-	-	0,3	- mA
	10		9,5		1,4	-	1,2	-	-	0,95	- mA
	15		13,5		4,8	-	4,0	-	-	3,2	- mA
	5		2,5		1,4	-	1,2	-	-	0,95	- mA
Output current LOW; C_{TC} , R_{TC}	5	0,4		I_{OL}	0,33	-	0,27	-	-	0,20	- mA
	10	0,5			1,00	-	0,85	-	-	0,68	- mA
	15	1,5			3,20	-	2,70	-	-	2,30	- mA

Note

- All inputs at 0 V or V_{DD} ; except input \overline{AR} = input MR = 0 V (power-on reset active).

AC CHARACTERISTICS

 $V_{SS} = 0 \text{ V}$; $T_{amb} = 25 \text{ }^{\circ}\text{C}$; input transition times $\leq 20 \text{ ns}$

	V_{DD} V	TYPICAL FORMULA FOR P (μW) ⁽¹⁾		
Dynamic power dissipation per package (P)	5	1 300	$f_i + f_o C_L V_{DD}^2$	
	10	5 300	$f_i + f_o C_L V_{DD}^2$	
	15	12 000	$f_i + f_o C_L V_{DD}^2$	
Total power dissipation when using the on-chip oscillator (P)	5	1 300	$f_{osc} + f_o C_L V_{DD}^2 + 2C_t V_{DD}^2 f_{osc} + 10 V_{DD}$	
	10	5 300	$f_{osc} + f_o C_L V_{DD}^2 + 2C_t V_{DD}^2 f_{osc} + 100 V_{DD}$	
	15	12 000	$f_{osc} + f_o C_L V_{DD}^2 + 2C_t V_{DD}^2 f_{osc} + 400 V_{DD}$	

Notes

- where:

f_i = input frequency (MHz)
 f_o = output frequency (MHz)
 C_L = load capacitance (pF)
 V_{DD} = supply voltage (V)
 C_t = timing capacitance (pF)
 f_{osc} = oscillator frequency (MHz)

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AC CHARACTERISTICS

$V_{SS} = 0 \text{ V}$; $T_{amb} = 25^\circ\text{C}$; $C_L = 50 \text{ pF}$; input transition times $\leq 20 \text{ ns}$

	V_{DD} V	SYMBOL	MIN.	TYP.	MAX.	TYPICAL EXTRAPOLATION FORMULA
Propagation delays $RS \rightarrow O$ 2^8 selected HIGH to LOW LOW to HIGH	5 10 15	$t_{PHL};$ t_{PLH}	375 150 110	750 300 220	ns ns ns	$348 \text{ ns} + (0,55 \text{ ns/pF}) C_L$ $139 \text{ ns} + (0,23 \text{ ns/pF}) C_L$ $102 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
$RS \rightarrow O$ 2^{10} selected HIGH to LOW LOW to HIGH	5 10 15	$t_{PHL};$ t_{PLH}	425 165 120	850 330 240	ns ns ns	$398 \text{ ns} + (0,55 \text{ ns/pF}) C_L$ $154 \text{ ns} + (0,23 \text{ ns/pF}) C_L$ $112 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
$RS \rightarrow O$ 2^{13} selected HIGH to LOW LOW to HIGH	5 10 15	$t_{PHL};$ t_{PLH}	510 190 135	1020 380 270	ns ns ns	$483 \text{ ns} + (0,55 \text{ ns/pF}) C_L$ $179 \text{ ns} + (0,23 \text{ ns/pF}) C_L$ $127 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
$RS \rightarrow O$ 2^{16} selected HIGH to LOW LOW to HIGH	5 10 15	$t_{PHL};$ t_{PLH}	575 210 150	1150 420 300	ns ns ns	$548 \text{ ns} + (0,55 \text{ ns/pF}) C_L$ $199 \text{ ns} + (0,23 \text{ ns/pF}) C_L$ $142 \text{ ns} + (0,16 \text{ ns/pF}) C_L$
Minimum clock pulse width; LOW	5 10 15	t_{WRSL}	60 30 24	30 15 12	ns ns ns	
Minimum reset pulse width; HIGH	5 10 15	t_{WMRH}	60 30 24	30 15 12	ns ns ns	
Maximum clock pulse frequency	5 10 15	f_{max}	8 15 18	16 30 36	MHz MHz MHz	
Oscillator frequency	5 10 15	f_{osc}		90 90 90	kHz kHz kHz	$R_t = 5 \text{ k}\Omega$ $C_t = 1 \text{ nF}$ $R_S = 10 \text{ k}\Omega$
Oscillator frequency	5 10 15	f_{osc}		8 8 8	kHz kHz kHz	$R_t = 56 \text{ k}\Omega$ $C_t = 1 \text{ nF}$ $R_S = 120 \text{ k}\Omega$