September 1998

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SIMPLE

SWITCHER

Power

Converter

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Efficiency

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Step-Down

Voltage Regulator

National Semiconductor

LM2675 SIMPLE SWITCHER[®] Power Converter High Efficiency 1A Step-Down Voltage Regulator

General Description

The LM2675 series of regulators are monolithic integrated circuits built with a LMDMOS process. These regulators provide all the active functions for a step-down (buck) switching regulator, capable of driving a 1A load current with excellent line and load regulation. These devices are available in fixed output voltages of 3.3V, 5.0V, 12V, and an adjustable output version.

Requiring a minimum number of external components, these regulators are simple to use and include patented internal frequency compensation (Patent Nos. 5,382,918 and 5,514,947) and a fixed frequency oscillator.

The LM2675 series operates at a switching frequency of 260 kHz, thus allowing smaller sized filter components than what would be needed with lower frequency switching regulators. Because of its very high efficiency (>90%), the copper traces on the printed circuit board are the only heat sinking needed.

A family of standard inductors for use with the LM2675 are available from several different manufacturers. This feature greatly simplifies the design of switch-mode power supplies using these advanced ICs. Also included in the datasheet are selector guides for diodes and capacitors designed to work in switch-mode power supplies.

Other features include a guaranteed $\pm 1.5\%$ tolerance on output voltage within specified input voltages and output load conditions, and $\pm 10\%$ on the oscillator frequency. External shutdown is included, featuring typically 50 µA stand-by current. The output switch includes current limiting, as well as thermal shutdown for full protection under fault conditions.

To simplify the LM2675 buck regulator design procedure, there exists computer design software, *LM267X Made Simple* version 1.0.

Features

- Efficiency up to 96%
- Available in SO-8 and 8-pin DIP packages
- Computer Design Software LM267X Made Simple
- Simple and easy to design with
- Requires only 5 external components
- Uses readily available standard inductors
- 3.3V, 5.0V, 12V, and adjustable output versions
- Adjustable version output voltage range: 1.21V to 37V
 ±1.5% max output voltage tolerance over line and load
- conditions
- Guaranteed 1A output load current
- 0.25Ω DMOS Output Switch
- Wide input voltage range: 8V to 40V
- 260 kHz fixed frequency internal oscillator
- TTL shutdown capability, low power standby mode
 Thermal shutdown and current limit protection
- IN.IOC CON.

Typical Applications

- Simple High Efficiency (>90%) Step-Down (Buck) Regulator
- Efficient Pre-Regulator for Linear Regulators
- Positive-to-Negative Converter



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WWW.100Y.COM.TW WWW.100Y.COM.TW Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

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Supply Voltage	45V
ON/OFF Pin Voltage	$-0.1V \le V_{SH} \le 6V$
Switch Voltage to Ground	-1V
Boost Pin Voltage	V _{sw} + 8V
Feedback Pin Voltage	$-0.3V \le V_{FB} \le 14V$
ESD Susceptibility	
Human Body Model (Note 2)	2 kV
Power Dissipation	Internally Limited

Storage Temperature Range	-65°C to +150°C
Lead Temperature	
M Package	
Vapor Phase (60s)	+215°C
Infrared (15s)	+220°C
N Package (Soldering, 10s)	+260°C
Maximum Junction Temperature	+150°C

Operating Ratings

Supply Voltage	6.5V to 40V
Junction Temperature Range	$-40^{\circ}C \le T_{J} \le +125^{\circ}C$

Electrical Characteristics Specifications with standard type face are for T_J = 25°C, and those with **bold** type face apply over full Operating Temperature Range.

LM2675-3.3

LM2	675-3.3					
Symbol	Parameter	Conditions	Typical (Note 4)	Min (Note 5)	Max (Note 5)	Units
SYSTEM	A PARAMETERS T	est Circuit Figure 2 (Note 3)	TW	A.	-11001.	
VOUT	Output Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	3.3	3.251/ 3.201	3.350/3.399	V
Vout	Output Voltage	$V_{IN} = 6.5V$ to 40V, $I_{LOAD} = 20$ mA to 500 mA	3.3	3.251/ 3.201	3.350/3.399	V
η	Efficiency	V _{IN} = 12V, I _{LOAD} = 1A	86	N	100	%

LM2675-5.0

Symbol	Parameter	Conditions	Typical (Note 4)	Min (Note 5)	Max (Note 5)	Units
SYSTEM	I PARAMETERS T	est Circuit Figure 2 (Note 3)	COM-	- 1	WW	.10
Vout	Output Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	5.0	4.925/4.850	5.075/ 5.150	V
VOUT	Output Voltage	V_{IN} = 6.5V to 40V, I_{LOAD} = 20 mA to 500 mA	5.0	4.925/ 4.850	5.075/ 5.150	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	90	1.1		%
LM2	675-12	CONTA WWW.10	NY.CO	MITW	N.	WW.

LM2675-12

Symbol	Parameter	Conditions	Typical	Min	Max	Units
	WW 100	Y. WILLING	(Note 4)	(Note 5)	(Note 5)	
SYSTEM	I PARAMETERS Te	est Circuit Figure 2 (Note 3)	. Non	VT.		NV.
V _{OUT}	Output Voltage	V_{IN} = 15V to 40V, I_{LOAD} = 20 mA to 1A	12	11.82/11.64	12.18/ 12.36	V
η	Efficiency	$V_{IN} = 24V, I_{LOAD} = 1A$	94	LA	14	%

Symbol	Parameter	Conditions	Typ (Note 4)	Min (Note 5)	Max (Note 5)	Units
SYSTE	M PARAMETERS T	est Circuit Figure 3 (Note 3)	WW.	NCOr		1
V _{FB}	Feedback Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A V_{OUT} Programmed for 5V (see Circuit of <i>Figure 3</i>)	1.210	1.192/ 1.174	1.228/ 1.246	v
V _{FB}	Feedback Voltage	$V_{IN} = 6.5V \text{ to } 40V, I_{LOAD} = 20 \text{ mA to } 500 \text{ mA}$ V_{OUT} Programmed for 5V (see Circuit of <i>Figure 3</i>)	1.210	1.192/ 1.174	1.228/ 1.246	v
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	90			%

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Electrical Characteristics

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Specifications with standard type face are for $T_J = 25$ °C, and those with **bold type face** apply over **full Operating Temperature Range**. Unless otherwise specified, $V_{IN} = 12V$ for the 3.3V, 5V, and Adjustable versions and $V_{IN} = 24V$ for the 12V version, and $I_{LOAD} = 100$ mA.

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Symbol	Parameters	Conditions	Тур	Min	Max	Units
DEVICE I	PARAMETERS	WW 100Y. TW		W	1001.	A
Iq	Quiescent Current	V _{FEEDBACK} = 8V For 3.3V, 5.0V, and ADJ Versions	2.5	MMM	3.6	mA
	COMTW	V _{FEEDBACK} = 15V For 12V Versions	2.5	WW.	W.100Y	mA
I _{STBY}	Standby Quiescent Current	ON/OFF Pin = 0V	50		100/ 150	μA
ICL	Current Limit	V 1001. ONLI	1.55	1.25/ 1.2	2.1/2.2	A
l <u>i</u>	Output Leakage Current	$V_{IN} = 40V, ON/\overline{OFF}$ Pin = 0V $V_{SWITCH} = 0V$	1	1	25	μA
	NT. YOL TW	$V_{SWITCH} = -1V, ON/\overline{OFF}$ Pin = 0V	6	1	15	mA
R _{DS(ON)}	Switch On-Resistance	I _{SWITCH} = 1A	0.25		0.30/ 0.50	Ω
fo	Oscillator Frequency	Measured at Switch Pin	260	225	275	kHz
D	Maximum Duty Cycle	WWWWWWWW	95	N	AN.	%
	Minimum Duty Cycle	N.W.W.	0			%
I _{BIAS}	Feedback Bias Current	V _{FEEDBACK} = 1.3V ADJ Version Only	85		V	nA
V _{S/D}	ON/OFF Pin Voltage Thesholds	TW WWW.100 Y	1.4	0.8	2.0	٧.
I _{S/D}	ON/OFF Pin Current	ON/OFF Pin = 0V	20	7	37	μA
θ_{JA}	Thermal Resistance	N Package, Junction to Ambient (Note 6)	95 105	M.TW	V	°C/W

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but device parameter specifications may not be guaranteed under these conditions. For guaranteed specifications and test conditions, see the Electrical Characteristics.

Note 2: The human body model is a 100 pF capacitor discharged through a 1.5 k Ω resistor into each pin.

Note 3: External components such as the catch diode, inductor, input and output capacitors, and voltage programming resistors can affect switching regulator performance. When the LM2675 is used as shown in Figures 2, 3 test circuits, system performance will be as specified by the system parameters section of the Electrical Characteristics.

Note 4: Typical numbers are at 25°C and represent the most likely norm.

Note 5: All limits guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% production tested. All limits at temperature extremes are guaranteed via correlation using standard Statistical Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

Note 6: Junction to ambient thermal resistance with approximately 1 square inch of printed circuit board copper surrounding the leads. Additional copper area will lower thermal resistance further. See Application Information section in the application note accompanying this datasheet and the thermal model in LM267X Made WWW.100Y.COM. Simple software.





* No Connections

For Surface Mount Package Order Number LM2675M-3.3, LM2675M-5.0, LM2675M-12 or LM2675M-ADJ See NSC Package Number M08A For DIP Package Order Number LM2675N-3.3, LM2675N-5.0, LM2675N-12 or LM2675N-ADJ See NSC Package Number N08E

Typical Performance Characteristics



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PROCEDURE (Fixed Output Voltage Version)

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C. Select an appropriate inductor from the four manufacturer's part numbers listed in Figure 8. Each manufacturer makes a different style of inductor to allow flexibility in meeting various design requirements. Listed below are some of the

differentiating characteristics of each manufacturer's inductors: Schott: ferrite EP core inductors; these have very low leakage magnetic fields to reduce electro-magnetic interference (EMI) and are the lowest power loss inductors

Renco: ferrite stick core inductors; benefits are typically lowest cost inductors and can withstand E•T and transient peak currents above rated value. Be aware that these inductors have an external magnetic field which may generate more EMI than other types of inductors.

Pulse: powered iron toroid core inductors; these can also be low cost and can withstand larger than normal E•T and transient peak currents. Toroid inductors have low EMI.

Coilcraft: ferrite drum core inductors; these are the smallest physical size inductors, available only as SMT components. Be aware that these inductors also generate EMI - but less than stick inductors

Complete specifications for these inductors are available from the respective manufacturers. A table listing the manufacturers' phone numbers is located in Figure 9.

2. Output Capacitor Selection (COUT)

A. Select an output capacitor from the output capacitor table in Figure 10. Using the output voltage and the inductance value found in the inductor selection guide, step 1, locate the appropriate capacitor value and voltage rating.

The capacitor list contains through-hole electrolytic capacitors from four different capacitor manufacturers and surface mount tantalum capacitors from two different capacitor manufacturers. It is recommended that both the manufacturers and the manufacturer's series that are listed in the table be used. A table listing the manufacturers' phone numbers is located in Figure 11.

3. Catch Diode Selection (D1)

A. In normal operation, the average current of the catch diode is the load current times the catch diode duty cycle, 1-D (D is the switch duty cycle, which is approximately the output voltage divided by the input voltage). The largest value of the catch diode average current occurs at the maximum load current and maximum input voltage (minimum D). For normal operation, the catch diode current rating must be at least 1.3 times greater than its maximum average current. However, if the power supply design must withstand a continuous output short, the diode should have a current rating equal to the maximum current limit of the LM2675. The most stressful condition for this diode is a shorted output condition.

B. The reverse voltage rating of the diode should be at least 1.25 times the maximum input voltage.

C. Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best performance and efficiency. This Schottky diode must be located close to the LM2675 using short leads and short printed circuit traces.

EXAMPLE (Fixed Output Voltage Version)

C. The inductance value required is 33 µH. From the table in Figure 8, go to the L23 line and choose an inductor part number from any of the four manufacturers shown. (In most instances, both through hole and surface mount inductors are available.)

2. Output Capacitor Selection (COUT)

A. Use the 5.0V section in the output capacitor table in Figure 10. Choose a capacitor value and voltage rating from the line that contains the inductance value of 33 µH. The capacitance and voltage rating values corresponding to the 33 µH inductor are the:

Surface Mount:

68 µF/10V Sprague 594D Series. 100 µF/10V AVX TPS Series. Through Hole: 68 µF/10V Sanyo OS-CON SA Series. 220 µF/35V Sanyo MV-GX Series. 220 µF/35V Nichicon PL Series. 220 µF/35V Panasonic HFQ Series.

3. Catch Diode Selection (D1)

A. Refer to the table shown in Figure 12. In this example, a 1A, 20V Schottky diode will provide the best performance. If the circuit must withstand a continuous shorted output, a higher current Schottky diode is recommended.

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PROCEDURE (Fixed Output Voltage Version) 4. Input Capacitor (CIN)

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EXAMPLE (Fixed Output Voltage Version) 4. Input Capacitor (CIN)

A low ESR aluminum or tantalum bypass capacitor is needed between the input pin and ground to prevent large voltage transients from appearing at the input. This capacitor should be located close to the IC using short leads. In addition, the RMS current rating of the input capacitor should be selected to be at least 1/2 the DC load current. The capacitor manufacturer data sheet must be checked to assure that this current rating is not exceeded. The curves shown in Figure 14 show typical RMS current ratings for several different aluminum electrolytic capacitor values. A parallel connection of two or more capacitors may be required to increase the total minimum RMS current rating to suit the application requirements. For an aluminum electrolytic capacitor, the voltage rating should be at least 1.25 times the maximum input voltage. Caution must be exercised if solid tantalum capacitors are used. The tantalum capacitor voltage rating should be twice the maximum input voltage. The tables in Figure 15 show the recommended application voltage for AVX TPS and Sprague 594D tantalum capacitors. It is also recommended that they be surge current tested by the manufacturer. The TPS series available from AVX, and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line. Use caution when using ceramic capacitors for input bypassing, because it may cause severe ringing at the V_{IN} pin. 5. Boost Capacitor (C_B)

This capacitor develops the necessary voltage to turn the switch gate on fully. All applications should use a 0.01 μ F, 50V ceramic capacitor. WWW.100Y.COM.TV WWW.100Y.COM.T WWW.100Y

The important parameters for the input capacitor are the input voltage rating and the RMS current rating. With a maximum input voltage of 12V, an aluminum electrolytic capacitor with a voltage rating greater than 15V (1.25 x V_{IN}) would be needed. The next higher capacitor voltage rating is 16V.

The RMS current rating requirement for the input capacitor in a buck regulator is approximately 1/2 the DC load current. In this example, with a 1A load, a capacitor with a RMS current rating of at least 500 mA is needed. The curves shown in Figure 14 can be used to select an appropriate input capacitor. From the curves, locate the 16V line and note which capacitor values have RMS current ratings greater than 500 mA. For a through hole design, a 330 µF/16V electrolytic capacitor (Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or equivalent) would be adequate. Other types or other manufacturers' capacitors can be used provided the RMS ripple current ratings are adequate. Additionally, for a complete surface mount design, electrolytic capacitors such as the Sanyo CV-C or CV-BS and the Nichicon WF or UR and the NIC Components NACZ series could be considered. For surface mount designs, solid tantalum capacitors can be used, but caution must be exercised with regard to the capacitor surge current rating and voltage rating. In this example, checking Figure 15, and the Sprague 594D series datasheet, a Sprague 594D 15 µF, 25V capacitor is adequate.

5. Boost Capacitor (C_B)

WWW.10 Y.COM.TW For this application, and all applications, use a 0.01 µF, 50V ceramic capacitor.



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	675 Se	eries B	uck Re	gulator	Design Pro	ocedure	(Fixed (Dutput) (Continued)
Ind	Induc-		Sc	hott	Ren	<u>co</u>	Pulse F	ngineering	Coilcraf
Ref. Desg.	tance (µH)	Current (A)	Through Hole	Surface Mount	Through Hole	Surface Mount	Through Hole	Surface Mount	Surface Mount
L4	68	0.32	67143940	67144310	RL-1284-68-43	RL1500-68	PE-53804	PE-53804-S	DO1608-68
L5	47	0.37	67148310	67148420	RL-1284-47-43	RL1500-47	PE-53805	PE-53805-S	DO1608-473
L6	33	0.44	67148320	67148430	RL-1284-33-43	RL1500-33	PE-53806	PE-53806-S	DO1608-333
L7	22	0.52	67148330	67148440	RL-1284-22-43	RL1500-22	PE-53807	PE-53807-S	DO1608-223
L9	220	0.32	67143960	67144330	RL-5470-3	RL1500-220	PE-53809	PE-53809-S	DO3308-224
L10	150	0.39	67143970	67144340	RL-5470-4	RL1500-150	PE-53810	PE-53810-S	DO3308-154
L11	100	0.48	67143980	67144350	RL-5470-5	RL1500-100	PE-53811	PE-53811-S	DO3308-104
L12	68	0.58	67143990	67144360	RL-5470-6	RL1500-68	PE-53812	PE-53812-S	DO3308-68
L13	47	0.70	67144000	67144380	RL-5470-7	RL1500-47	PE-53813	PE-53813-S	DO3308-473
L14	33	0.83	67148340	67148450	RL-1284-33-43	RL1500-33	PE-53814	PE-53814-S	DO3308-333
L15	22	0.99	67148350	67148460	RL-1284-22-43	RL1500-22	PE-53815	PE-53815-S	DO3308-223
L18	220	0.55	67144040	67144420	RL-5471-2	RL1500-220	PE-53818	PE-53818-S	DO3316-224
L19	150	0.66	67144050	67144430	RL-5471-3	RL1500-150	PE-53819	PE-53819-S	DO3316-154
L20	100	0.82	67144060	67144440	RL-5471-4	RL1500-100	PE-53820	PE-53820-S	DO3316-104
L21	68	0.99	67144070	67144450	RL-5471-5	RL1500-68	PE-53821	PE-53821-S	DO3316-683
L22	47	1.17	67144080	67144460	RL-5471-6		PE-53822	PE-53822-S	DO3316-473
L23	33	1.40	67144090	67144470	RL-5471-7	cott.	PE-53823	PE-53823-S	DO3316-333
L24	22	1.70	67148370	67148480	RL-1283-22-43		PE-53824	PE-53824-S	DO3316-223
L27	220	1.00	67144110	67144490	RL-5471-2	TCOM.	PE-53827	PE-53827-S	DO5022P-22
L28	150	1.20	67144120	67144500	RL-5471-3	Mo-	PE-53828	PE-53828-S	DO5022P-1
L29	100	1.47	67144130	67144510	RL-5471-4	N.COM	PE-53829	PE-53829-S	DO5022P-1
1.30	68	1.78	67144140	67144520	RL-5471-5		PE-53830	PE-53830-S	DO5022P-6

Coilcraft Inc.	Phone	(800) 322-2645
	FAX	(708) 639-1469
Coilcraft Inc., Europe	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
Pulse Engineering Inc.	Phone	(619) 674-8100
	FAX	(619) 674-8262
Pulse Engineering Inc.,	Phone	+353 93 24 107
Europe	FAX	+353 93 24 459
Renco Electronics Inc.	Phone	(800) 645-5828
	FAX	(516) 586-5562
Schott Corp.	Phone	(612) 475-1173
	FAX	(612) 475-1786

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	J Jenes D	uck Regul	ator Des	ign Procedu	ure (Fixed (Output)	Continued)
COM	<u>TN</u>		.100X.C	Output Ca	apacitor	W.1001.	CONT
Output	WT	Surface	Mount		Through H	lole	· ·
Voltage (V)	Inductance (µH)	Sprague 594D Series (µF/V)	AVX TPS Series (µF/V)	Sanyo OS-CON SA Series (µF/V)	Sanyo MV-GX Series (µF/V)	Nichicon PL Series (µF/V)	Panasonic HFQ Series (µF/V)
-1 C	22	120/6.3	100/10	100/10	330/35	330/35	330/35
	33	120/6.3	100/10	68/10	220/35	220/35	220/35
	47	68/10	100/10	68/10	150/35	150/35	150/35
3.3	68	120/6.3	100/10	100/10	120/35	120/35	120/35
	100	120/6.3	100/10	100/10	120/35	120/35	120/35
	150	120/6.3	100/10	100/10	120/35	120/35	120/35
-x1100	22	100/16	100/10	100/10	330/35	330/35	330/35
	33	68/10	10010	68/10	220/35	220/35	220/35
1 10	47	68/10	100/10	68/10	150/35	150/35	150/35
5.0	68	100/16	100/10	100/10	120/35	120/35	120/35
	100	100/16	100/10	100/10	120/35	120/35	120/35
	150	100/16	100/10	100/10	120/35	120/35	120/35
VN	22	120/20	(2x) 68/20	68/20	330/35	330/35	330/35
	33	68/25	68/20	68/20	220/35	220/35	220/35
	47	47/20	68/20	47/20	150/35	150/35	150/35
12	68	47/20	68/20	47/20	120/35	120/35	120/35
	100	47/20	68/20	47/20	120/35	120/35	120/35
	150	47/20	68/20	47/20	120/35	120/35	120/35
	220	47/20	68/20	47/20	120/35	120/35	120/35

Nichicon Corp.	Phone	(847) 843-7500
	FAX	(847) 843-2798
Panasonic	Phone	(714) 373-7857
	FAX	(714) 373-7102
AVX Corp.	Phone	(803) 448-9411
	FAX	(803) 448-1943
Sprague/Vishay	Phone	(207) 324-4140
	FAX	(207) 324-7223
Sanyo Corp.	Phone	(619) 661-6322
	FAX	(619) 661-1055

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	1A Di	odes	3A Diodes		
V _R	Surface Mount	Through Hole	Surface Mount	Through Hole	
20V	SK12	1N5817	SK32	1N5820	
	B120	SR102	10	SR302	
30V	SK13	1N5818	SK33	1N5821	
	B130	11DQ03	30WQ03F	31DQ03	
	MBRS130	SR103	NN.	. Mar	
40V	SK14	1N5819	SK34	1N5822	
	B140	11DQ04	30BQ040	MBR340	
	MBRS140	SR104	30WQ04F	31DQ04	
	10BQ040		MBRS340	SR304	
	10MQ040		MBRD340	100	
	15MQ040	- 1	-15	W.r.	
50V	SK15	MBR150	SK35	MBR350	
	B150	11DQ05	30WQ05F	31DQ05	
	10BQ050	SR105		SR305	

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FIGURE 12. Schottky Diode Selection Table

V.100Y.COM.	LM.	FIGURE 12. Scho
International Rectifier Corp.	Phone	(310) 322-3331
NN. COL	FAX	(310) 322-3332
Motorola, Inc.	Phone	(800) 521-6274
	FAX	(602) 244-6609
General Instruments Corp.	Phone	(516) 847-3000
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850





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AVX TPS

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Recommended Application Voltage	Voltage Rating
+85°C Ratin	g
3.3	6.3
5	10
10	20
12	25
15	35

Sprague 594D WW.100Y.COM.T

Recommended Application Voltage	Voltage Rating
+85°C Ratin	g cO
2.5	4
3.3	6.3
5	10
8	16
12	20
18	25
24	35
29	50

WWW.100Y.COM. FIGURE 15. Recommended Application Voltage for AVX TPS and Sprague 594D Tantalum Chip Capacitors Derated for 85°C.

LM2675 Series Buck Regulator Design Procedure (Adjustable Output)

PROCEDURE (Adjustable Output Voltage Version)

To simplify the buck regulator design procedure, National Semiconductor is making available computer design software to be used with the SIMPLE SWITCHER line of switching regulators. LM267X Made Simple version 1.0 is available for use on Windows 3.1, NT, or 95 operating systems.

Given:

V_{OUT} = Regulated Output Voltage

V_{IN}(max) = Maximum Input Voltage

ILOAD(max) = Maximum Load Current

F = Switching Frequency (Fixed at a nominal 260 kHz).

1. Programming Output Voltage (Selecting R1 and R2, as shown in Figure 3)

Use the following formula to select the appropriate resistor values.

$$V_{OUT} = V_{REF} \left(1 + \frac{R_2}{R_1} \right)$$

where $V_{RFF} = 1.21V$

Select a value for R₁ between 240 Ω and 1.5 k Ω . The lower resistor values minimize noise pickup in the sensitive feedback pin. (For the lowest temperature coefficient and the best stability with time, use 1% metal film resistors.)

$$R_2 = R_1 \left(\frac{V_{OUT}}{V_{REF}} - 1 \right)$$

EXAMPLE (Adjustable Output Voltage Version)

Given:

 $V_{OUT} = 20V$

 $V_{IN}(max) = 28V$

 $I_{LOAD}(max) = 1A$

F = Switching Frequency (Fixed at a nominal 260 kHz). 1. Programming Output Voltage (Selecting R1 and R2, as shown in Figure 3)

Select R_1 to be 1 k Ω , 1%. Solve for R_2 .

$$R_{2} = R_{1} \left(\frac{V_{OUT}}{V_{REF}} - 1 \right) = 1 \text{ k} \Omega \left(\frac{20V}{1.23V} - 1 \right)$$

 $R_2 = 1k (16.53 - 1) = 15.53 k\Omega$, closest 1% value is 15.4 kΩ. $R_2 = 15.4 \text{ k}\Omega.$

WWW.100Y.COM.T LM2675 Series Buck Regulator Design Procedure (Adjustable Output) (Continued) PROCEDURE (Adjustable Output Voltage Version) 2. Inductor Selection (L1) 2. Inductor Selection (L1) A. Calculate the inductor Volt • microsecond constant E • T (V • µs), from the following formula: $E \cdot T = (V_{IN(MAX)} - V_{OUT} - V_{SAT}) \cdot \frac{V_{UN}}{V_{IN(MAX)} - V_{SAT} + V_D}$ V_{OUT} + V_D $\cdot \frac{1000}{260}$ (V· μ s) where V_{SAT} = internal switch saturation voltage = 0.25V and V_{D} = diode forward voltage drop = 0.5V B. Use the E • T value from the previous formula and match it B. E • T = 21.6 (V • µs) with the E • T number on the vertical axis of the Inductor Value Selection Guide shown in Figure 7. C. On the horizontal axis, select the maximum load current. **C.** $I_{LOAD}(max) = 1A$ D. Identify the inductance region intersected by the E • T value and the Maximum Load Current value. Each region is identified the inductance region intersected by the 21.6 (V • µs) by an inductance value and an inductor code (LXX). inductor code is L30. E. Select an appropriate inductor from the four manufacturer's part numbers listed in Figure 8. For information on the different types of inductors, see the inductor selection in the fixed numbers. output voltage design procedure. 3. Output Capacitor Selection (COUT) 3. Output Capacitor Selection (Cout) A. Select an output capacitor from the capacitor code selection guide in Figure 16. Using the inductance value found in the inductor selection guide, step 1, locate the appropriate capacitor code corresponding to the desired output voltage. C20. B. Select an appropriate capacitor value and voltage rating, using the capacitor code, from the output capacitor selection

table in Figure 17. There are two solid tantalum (surface mount) capacitor manufacturers and four electrolytic (through hole) capacitor manufacturers to choose from. It is recommended that both the manufacturers and the manufacturer's series that are listed in the table be used. A table listing the manufacturers' phone numbers is located in Fiaure 11.

EXAMPLE (Adjustable Output Voltage Version)

A. Calculate the inductor Volt • microsecond constant (E • T),

$$E \cdot T = (28 - 20 - 0.25) \cdot \frac{20 + 0.5}{28 - 0.25 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$$
$$E \cdot T = (7.75) \cdot \frac{20.5}{28.25} \cdot 3.85 (V \cdot \mu s) = 21.6 (V \cdot \mu s)$$

D. From the inductor value selection guide shown in Figure 7, horizontal line and the 1A vertical line is 68 µH, and the

E. From the table in Figure 8, locate line L30, and select an inductor part number from the list of manufacturers part

A. Use the appropriate row of the capacitor code selection guide, in Figure 16. For this example, use the 15-20V row. The capacitor code corresponding to an inductance of 68 µH is

B. From the output capacitor selection table in Figure 17, choose a capacitor value (and voltage rating) that intersects the capacitor code(s) selected in section A, C20. The capacitance and voltage rating values corresponding to the capacitor code C20 are the:

Surface Mount:

33 µF/25V Sprague 594D Series. 33 µF/25V AVX TPS Series. Through Hole:

33 µF/25V Sanyo OS-CON SC Series.

120 µF/35V Sanyo MV-GX Series.

120 µF/35V Nichicon PL Series. 120 µF/35V Panasonic HFQ Series.

Other manufacturers or other types of capacitors may also be used, provided the capacitor specifications (especially the 100 kHz ESR) closely match the characteristics of the capacitors listed in the output capacitor table. Refer to the capacitor manufacturers' data sheet for this information.

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PROCEDURE (Adjustable Output Voltage Version)

4. Catch Diode Selection (D1)

A. In normal operation, the average current of the catch diode is the load current times the catch diode duty cycle, 1-D (D is the switch duty cycle, which is approximately $V_{\text{OUT}}/V_{\text{IN}}).$ The largest value of the catch diode average current occurs at the maximum input voltage (minimum D). For normal operation, the catch diode current rating must be at least 1.3 times greater than its maximum average current. However, if the power supply design must withstand a continuous output short, the diode should have a current rating greater than the maximum current limit of the LM2675. The most stressful condition for this diode is a shorted output condition.

B. The reverse voltage rating of the diode should be at least 1.25 times the maximum input voltage.

C. Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best performance and efficiency. The Schottky diode must be located close to the LM2675 using short leads and short printed circuit traces.

5. Input Capacitor (CIN)

A low ESR aluminum or tantalum bypass capacitor is needed between the input pin and ground to prevent large voltage transients from appearing at the input. This capacitor should be located close to the IC using short leads. In addition, the RMS current rating of the input capacitor should be selected to be at least 1/2 the DC load current. The capacitor manufacturer data sheet must be checked to assure that this current rating is not exceeded. The curves shown in Figure 14 show typical RMS current ratings for several different aluminum electrolytic capacitor values. A parallel connection of two or more capacitors may be required to increase the total minimum RMS current rating to suit the application requirements. For an aluminum electrolytic capacitor, the voltage rating should be at least 1.25 times the maximum input voltage. Caution must be exercised if solid tantalum capacitors are used. The tantalum capacitor voltage rating should be twice the maximum input voltage. The tables in Figure 15 show the recommended application voltage for AVX TPS and Sprague 594D tantalum capacitors. It is also recommended that they be surge current tested by the manufacturer. The TPS series available from AVX, and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line. Use caution when using ceramic capacitors for input bypassing, because it may cause severe ringing at the $V_{\mbox{\scriptsize IN}}$ pin.

6. Boost Capacitor (C_B)

This capacitor develops the necessary voltage to turn the switch gate on fully. All applications should use a 0.01 µF, 50V ceramic capacitor. WWW.100Y.COM.T

EXAMPLE (Adjustable Output Voltage Version)

4. Catch Diode Selection (D1)

A. Refer to the table shown in Figure 12. Schottky diodes provide the best performance, and in this example a 1A, 40V Schottky diode would be a good choice. If the circuit must withstand a continuous shorted output, a higher current (at least 2.2A) Schottky diode is recommended.

5. Input Capacitor (CIN)

The important parameters for the input capacitor are the input voltage rating and the RMS current rating. With a maximum input voltage of 28V, an aluminum electrolytic capacitor with a voltage rating of at least 35V (1.25 x V_{IN}) would be needed. The RMS current rating requirement for the input capacitor in a buck regulator is approximately 1/2 the DC load current. In this example, with a 1A load, a capacitor with a RMS current rating of at least 500 mA is needed. The curves shown in Figure 14 can be used to select an appropriate input capacitor. From the curves, locate the 35V line and note which capacitor values have RMS current ratings greater than 500 mA. For a through hole design, a 330 µF/35V electrolytic capacitor (Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or equivalent) would be adequate. Other types or other manufacturers' capacitors can be used provided the RMS ripple current ratings are adequate. Additionally, for a complete surface mount design, electrolytic capacitors such as the Sanyo CV-C or CV-BS, and the Nichicon WF or UR and the NIC Components NACZ series could be considered. For surface mount designs, solid tantalum capacitors can be used, but caution must be exercised with regard to the capacitor surge current rating and voltage rating. In this example, checking Figure 15, and the Sprague 594D series datasheet, a Sprague 594D 15 µF, 50V capacitor is adequate.

6. Boost Capacitor (CB)

For this application, and all applications, use a 0.01 $\mu\text{F},\,50\text{V}$ ceramic capacitor.

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Case	Output	11001.	Inductance (µH)						
Style (Note 7)	Voltage (V)	22	33	47	68	100	150	220	
SM and TH	1.21-2.50	N.1 -	COT .	<u> </u>	<u> </u>	C1	C2	C3	
SM and TH	2.50-3.75	$-\pi^{0}$		11-	C1	C2	C3	C3	
SM and TH	3.75-5.0	11-2		C4	C5	C6	C6	C6	
SM and TH	5.0-6.25	00	C4	C7	C6	C6	C6	C6	
SM and TH	6.25-7.5	C8	C4	C7	C6 <	C6	C6	C6	
SM and TH	7.5–10.0	C9	C10	C11	C12	C13	C13	C13	
SM and TH	10.0–12.5	C14	C11	C12	C12	C13	C13	C13	
SM and TH	12.5–15.0	C15	C16	C17	C17	C17	C17	C17	
SM and TH	15.0-20.0	C18	C19	C20	C20	C20	C20	C20	
SM and TH	20.0-30.0	C21	C22	C22	C22	C22	C22	C22	
TH	30.0-37.0	C23	C24	C24	C25	C25	C25	C25	

Note 7: SM - Surface Mount. TH - Through Hole W.100Y.CO W.100Y.COM.TW

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FIGURE 16. Capacitor Code Selection Guide WWW.100Y.COM WWW.I

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Output Capacitor							
Cap.	Surface	Mount	Through Hole				
Ref. Desg. #	Sprague 594D Series (µF/V)	AVX TPS Series (μF/V)	Sanyo OS-CON SA Series (μF/V)	Sanyo MV-GX Series (µF/V)	Nichicon PL Series (µF/V)	Panasonic HFQ Series (µF/V)	
C1	120/6.3	100/10	100/10	220/35	220/35	220/35	
C2	120/6.3	100/10	100/10	150/35	150/35	150/35	
C3	120/6.3	100/10	100/35	120/35	120/35	120/35	
C4	68/10	100/10	68/10	220/35	220/35	220/35	
C5	100/16	100/10	100/10	150/35	150/35	150/35	
C6	100/16	100/10	100/10	120/35	120/35	120/35	
C7	68/10	100/10	68/10	150/35	150/35	150/35	
C8	100/16	100/10	100/10	330/35	330/35	330/35	
C9	100/16	100/16	100/16	330/35	330/35	330/35	
C10	100/16	100/16	68/16	220/35	220/35	220/35	
C11	100/16	100/16	68/16	150/35	150/35	150/35	
C12	100/16	100/16	68/16	120/35	120/35	120/35	
C13	100/16	100/16	100/16	120/35	120/35	120/35	
C14	100/16	100/16	100/16	220/35	220/35	220/35	
C15	47/20	68/20	47/20	220/35	220/35	220/35	
C16	47/20	68/20	47/20	150/35	150/35	150/35	
C17	47/20	68/20	47/20	120/35	120/35	120/35	
C18	68/25	(2x) 33/25	47/25 (Note 8)	220/35	220/35	220/35	
C19	33/25	33/25	33/25 (Note 8)	150/35	150/35	150/35	
C20	33/25	33/25	33/25 (Note 8)	120/35	120/35	120/35	
C21	33/35	(2x) 22/25	(Note 9)	150/35	150/35	150/35	
C22	33/35	22/35	(Note 9)	120/35	120/35	120/35	
C23	(Note 9)	(Note 9)	(Note 9)	220/50	100/50	120/50	
C24	(Note 9)	(Note 9)	(Note 9)	150/50	100/50	120/50	
C25	(Note 9)	(Note 9)	(Note 9)	150/50	82/50	82/50	

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Note 8: The SC series of Os-Con capacitors (others are SA series)

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Note 9: The voltage ratings of the surface mount tantalum chip and Os-Con capacitors are too low to work at these voltages. WWW.100Y.COM.TW

FIGURE 17. Output Capacitor Selection Table WWW.100 WWW.100Y.COM

Application Information

TYPICAL SURFACE MOUNT PC BOARD LAYOUT, FIXED OUTPUT (4X SIZE)





 $\begin{array}{l} C_{IN} - 15 \ \mu\text{F}, \ 50\text{V}, \ Solid \ Tantalum \ Sprague, \ "594D \ series" \\ C_{OUT} - 68 \ \mu\text{F}, \ 16\text{V}, \ Solid \ Tantalum \ Sprague, \ "594D \ series" \\ D1 - 1A, \ 40\text{V} \ Schottky \ Rectifier, \ Surface \ Mount \\ L1 - 33 \ \mu\text{H}, \ L23, \ Colicraft \ DO316 \\ \end{array}$

C_B - 0.01 µF, 50V, Ceramic

TYPICAL SURFACE MOUNT PC BOARD LAYOUT, ADJUSTABLE OUTPUT (4X SIZE)



 $\begin{array}{l} C_{IN} \text{ - 15 } \mu\text{F}, 50\text{V}, \text{Solid Tantalum Sprague, "594D series"} \\ C_{OUT} \text{ - 33 } \mu\text{F}, 25\text{V}, \text{Solid Tantalum Sprague, "594D series"} \\ \text{D1 - 1A, 40V Schottky Rectifier, Surface Mount} \\ \text{L1 - 68 } \mu\text{H}, \text{L30, Coilcraft DO3316} \end{array}$

 C_B - 0.01 μF , 50V, Ceramic R1 - 1k, 1%

R2 - Use formula in Design Procedure

FIGURE 18. PC Board Layout

Layout is very important in switching regulator designs. Rapidly switching currents associated with wiring inductance can generate voltage transients which can cause problems. For minimal inductance and ground loops, the wires indicated by heavy lines (in *Figure 2* and *Figure 3*) should be wide printed circuit traces and should be kept as short as possible. For best results, external components should be located as close to the switcher IC as possible using ground plane construction or single point grounding.

If **open core inductors are used**, special care must be taken as to the location and positioning of this type of inductor. Allowing the inductor flux to intersect sensitive feedback, IC ground path, and C_{OUT} wiring can cause problems.

When using the adjustable version, special care must be taken as to the location of the feedback resistors and the associated wiring. Physically locate both resistors near the IC, and route the wiring away from the inductor, especially an open core type of inductor.

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