March 2000

M2679 SIMPLE

**SWITCHER 5A** 

Step-Down Voltage

Regulator

Adjustable

**Current Limit** 

National Semiconductor

## LM2679 SIMPLE SWITCHER<sup>®</sup> 5A Step-Down Voltage Regulator with Adjustable Current Limit

## **General Description**

The LM2679 series of regulators are monolithic integrated circuits which provide all of the active functions for a step-down (buck) switching regulator capable of driving up to 5A loads with excellent line and load regulation characteristics. High efficiency (>90%) is obtained through the use of a low ON-resistance DMOS power switch. The series consists of fixed output voltages of 3.3V, 5V and 12V and an adjustable output version.

The SIMPLE SWITCHER concept provides for a complete design using a minimum number of external components. A high fixed frequency oscillator (260KHz) allows the use of physically smaller sized components. A family of standard inductors for use with the LM2679 are available from several manufacturers to greatly simplify the design process.

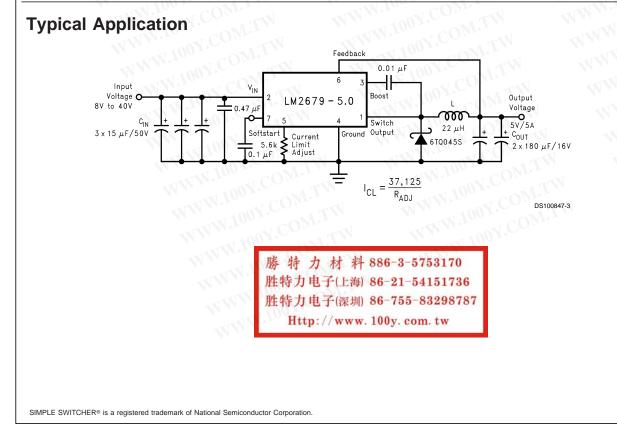
Other features include the ability to reduce the input surge current at power-ON by adding a softstart timing capacitor to gradually turn on the regulator. The LM2679 series also has built in thermal shutdown and resistor programmable current limit of the power MOSFET switch to protect the device and load circuitry under fault conditions. The output voltage is guaranteed to a  $\pm 2\%$  tolerance. The clock frequency is controlled to within a  $\pm 11\%$  tolerance.

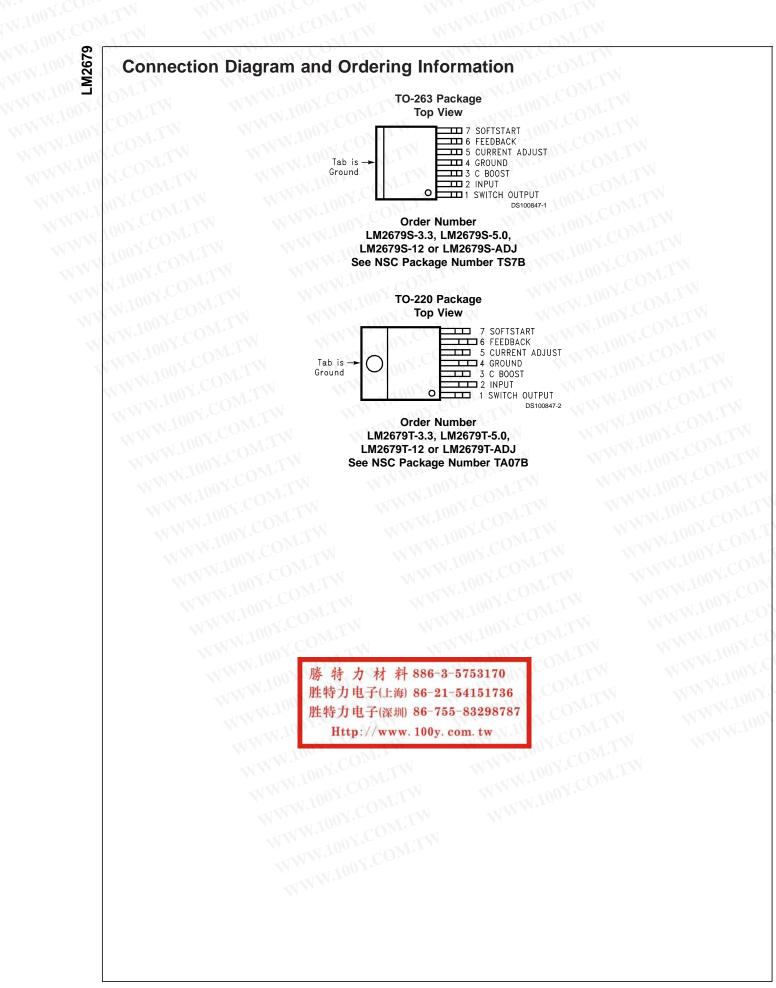
### Features

- Efficiency up to 92%
- Simple and easy to design with (using off-the-shelf external components)
- Resistor programmable peak current limit over a range of 3A to 7A.
- 120 mΩ DMOS output switch
- 3.3V, 5V and 12V fixed output and adjustable (1.2V to 37V) versions
- ±2%maximum output tolerance over full line and load conditions
- Wide input voltage range: 8V to 40V
- 260 KHz fixed frequency internal oscillator
- Softstart capability
- -40 to +125°C operating junction temperature range

## Applications

- Simple to design, high efficiency (>90%) step-down switching regulators
- Efficient system pre-regulator for linear voltage regulators
- Battery chargers





Absolute Maximum F	Ratings (Note 1)	Storage Temperature Range	–65°C to 150°
If Military/Aerospace specified	devices are required.	Soldering Temperature	
please contact the National Semi		Wave	4 sec, 260°
Distributors for availability and	specifications.	Infrared	10 sec, 240°
Input Supply Voltage	45V	Vapor Phase	75 sec, 219°
Softstart Pin Voltage	-0.1V to 6V	One section of Destination	
Switch Voltage to Ground	–1V to V <sub>IN</sub>	Operating Ratings	
Boost Pin Voltage	V <sub>SW</sub> + 8V	Supply Voltage	8V to 40
Feedback Pin Voltage	-0.3V to 14V	Junction Temperature Range (T <sub>J</sub> )	-40°C to 125°
Power Dissipation	Internally Limited	ON WWW.	
ESD (Note 2)	2 kV		

Symbol	Parameter	Conditions	Typical (Note 3)	Min (Note 4)	Max (Note 4)	Units
Vout	Output Voltage	$V_{IN} = 8V$ to 40V, 100mA $\leq I_{OUT} \leq 5A$	3.3	3.234/ <b>3.201</b>	3.366/3.399	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	82	NN	.100Y.C	%

Symbol	Parameter	Conditions	Typical (Note 3)	Min (Note 4)	Max (Note 4)	Units
V <sub>OUT</sub>	Output Voltage	$V_{IN} = 8V$ to 40V, 100mA $\leq I_{OUT} \leq 5A$	5.0	4.900/ <b>4.850</b>	5.100/ <b>5.150</b>	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	84	Wn	NW	%

Symbol	Parameter	Conditions	Typical (Note 3)	Min (Note 4)	Max (Note 4)	Units
V <sub>OUT</sub>	Output Voltage	$V_{IN} = 15V$ to 40V, 100mA $\leq I_{OUT} \leq 5A$	12	11.76/11.64	12.24/12.36	V
η	Efficiency	$V_{IN} = 24V, I_{LOAD} = 5A$	92	COMM	W	%

## LM2679-ADJ

LM2	679-ADJ					
Symbol	Parameter	Conditions	Typ (Note 3)	Min (Note 4)	Max (Note 4)	Units
$V_{FB}$	Feedback Voltage	$V_{IN}$ = 8V to 40V, 100mA $\leq I_{OUT} \leq$ 5A $V_{OUT}$ Programmed for 5V	1.21	1.186/ <b>1.174</b>	1.234/ <b>1.246</b>	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 5A$	84	100 - 00	1.1	%

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## All Output Voltage Versions Electrical Characteristics

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Limits appearing in **bold type face** apply over the entire junction temperature range of operation,  $-40^{\circ}$ C to 125°C. Specifications appearing in normal type apply for T<sub>A</sub> = T<sub>J</sub> = 25°C. Unless otherwise specified V<sub>IN</sub>=12V for the 3.3V, 5V and Adjustable versions and V<sub>IN</sub>=24V for the 12V version.

Symbol	Parameter	Conditions	Тур	Min	Max	Units
DEVICE	PARAMETERS	WWW. MOON WWWWWWWWWWWWWWWWWWWWW	14	N.CO. TV	N	•
	Quiescent Current	$V_{\text{FEEDBACK}} = 8V$ For 3.3V, 5.0V, and ADJ Versions $V_{\text{FEEDBACK}} = 15V$ For 12V Versions	4.2	0X.COM.T 100X.COM.T 100X.COM	6	mA
V <sub>ADJ</sub>	Current Limit Adjust Voltage	WWW.100Y.COM.TW	1.21	1.181/ <b>1.169</b>	1.229/ <b>1.246</b>	V
ICL	Current Limit	$R_{ADJ} = 5.6 K\Omega$ , (Note 5)	6.3	5.5/ <b>5.3</b>	7.6/ <b>8.1</b>	A
IL 100	Output Leakage Current	$V_{IN} = 40V$ , Softstart Pin = 0V $V_{SWITCH} = 0V$ $V_{SWITCH} = -1V$	1.0 6	W.100 Y.C	1.5 15	mA mA
R <sub>DS(ON)</sub>	Switch On-Resistance	I <sub>SWITCH</sub> = 5A	0.12	WW.1001	0.14/ <b>0.225</b>	Ω
fo	Oscillator Frequency	Measured at Switch Pin	260	225	280	kHz
D	Duty Cycle	Maximum Duty Cycle Minimum Duty Cycle	91 0	WWW.10	OX.COM	%
I <sub>BIAS</sub>	Feedback Bias Current	V <sub>FEEDBACK</sub> = 1.3V ADJ Version Only	85	WWW.	100Y.CON	nA
V <sub>SFST</sub>	Softstart Threshold Voltage	OM.TW WWW.100Y.COM	0.63	0.53	0.74	v
I <sub>SFST</sub>	Softstart Pin Current	Softstart Pin = 0V	3.7	WW	6.9	μA
$\theta_{JA}$	Thermal Resistance	T Package, Junction to Ambient (Note 6)	65		WW.1001	.00
$\theta_{JA}$		T Package, Junction to Ambient (Note 7)	45	WT	WWW.10	N.C
$\theta_{\text{JC}}$		T Package, Junction to Case	2	NTN .	NW AL	101.0
$\theta_{JA}$	WWW	S Package, Junction to Ambient (Note 8)	.56	WT.10	WWW.	°C/W
$\theta_{JA}$		S Package, Junction to Ambient (Note 9)	35	WT.I.MC	WWW	1.100
$\theta_{JA}$		S Package, Junction to Ambient (Note 10)	26	ON.TW	WW	N.10
$\theta_{\text{JC}}$		S Package, Junction to Case	2	WT.Mod		++

Note 1: Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings indicate conditions under which of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test condition, see the electrical Characteristics tables.

Note 2: ESD was applied using the human-body model, a 100pF capacitor discharged through a 1.5 k $\Omega$  resistor into each pin.

Note 3: Typical values are determined with  $T_A = T_J = 25^{\circ}C$  and represent the most likely norm.

**Note 4:** All limits are guaranteed at room temperature (standard type face) and at **temperature extremes (bold type face)**. All room temperature limits are 100% tested during production with  $T_A = T_J = 25^{\circ}$ C. All limits at temperature extremes are guaranteed via correlation using standard standard Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

Note 5: The peak switch current limit is determined by the following relationship: I<sub>CL</sub>=37,125/ R<sub>ADJ</sub>.

Note 6: Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads in a socket, or on a PC board with minimum copper area.

Note 7: Junction to ambient thermal resistance (no external heat sink) for the 7 lead TO-220 package mounted vertically, with ½ inch leads soldered to a PC board containing approximately 4 square inches of (1 oz.) copper area surrounding the leads.

Note 8: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.136 square inches (the same size as the TO-263 package) of 1 oz. (0.0014 in. thick) copper.

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Note 9: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board area of 0.4896 square inches (3.6 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper.

Note 10: Junction to ambient thermal resistance for the 7 lead TO-263 mounted horizontally against a PC board copper area of 1.0064 square inches (7.4 times the area of the TO-263 package) of 1 oz. (0.0014 in. thick) copper. Additional copper area will reduce thermal resistance further. See the thermal model in Switchers Made Simple<sup>®</sup> software

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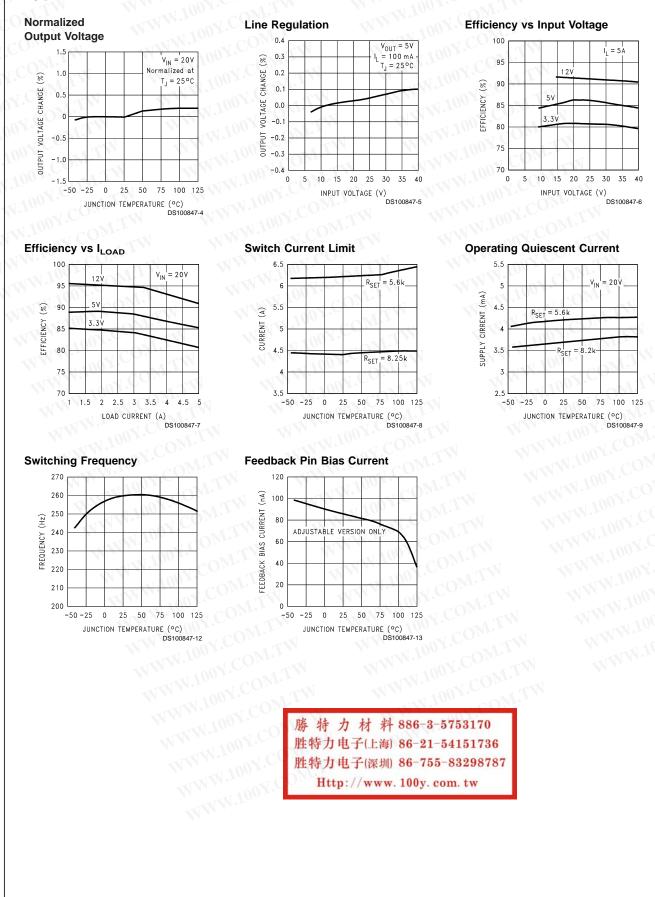
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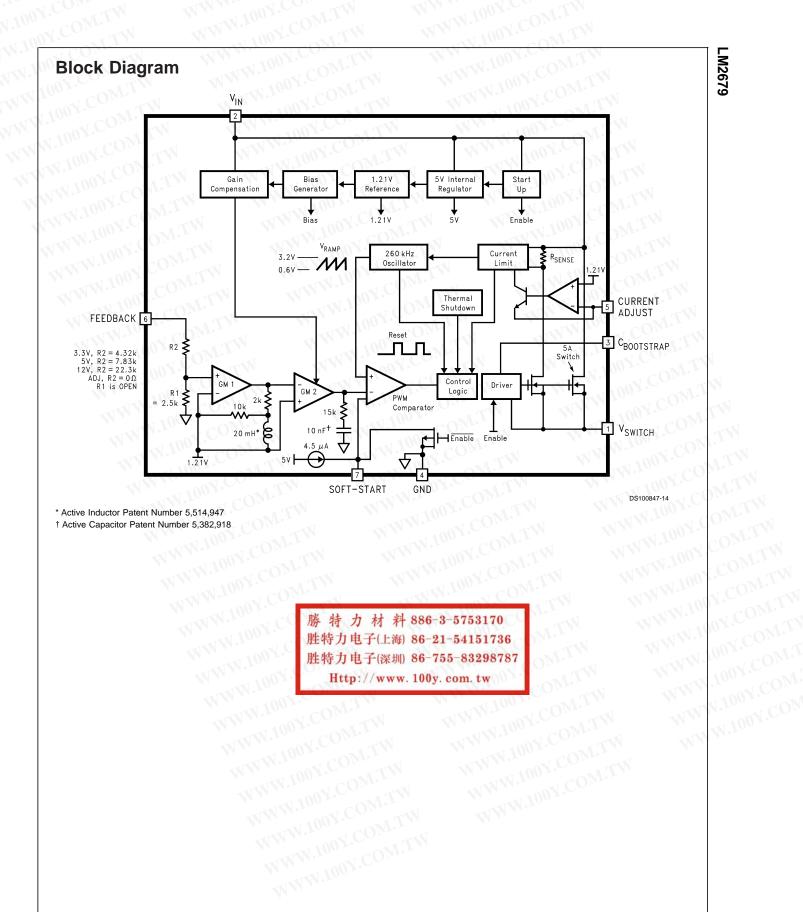
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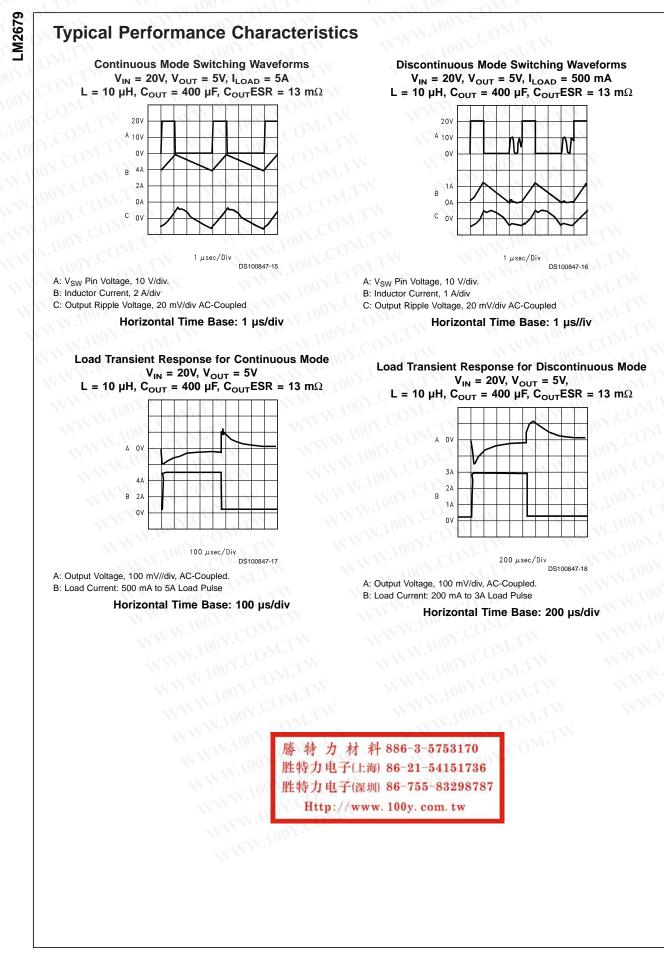
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# LM2679

## **Typical Performance Characteristics**







## **Application Hints**

The LM2679 provides all of the active functions required for a step-down (buck) switching regulator. The internal power switch is a DMOS power MOSFET to provide power supply designs with high current capability, up to 5A, and highly efficient operation.

The LM2679 is part of the *SIMPLE SWITCHER* family of power converters. A complete design uses a minimum number of external components, which have been pre-determined from a variety of manufacturers. Using either this data sheet or a design software program called *LM267X Made Simple* (version 2.0) a complete switching power supply can be designed quickly. The software is provided free of charge and can be downloaded from National Semiconductor's Internet site located at http://www.national.com.

#### PIN 1 - Switch Output

This is the output of a power MOSFET switch connected directly to the input voltage. The switch provides energy to an inductor, an output capacitor and the load circuitry under control of an internal pulse-width-modulator (PWM). The PWM controller is internally clocked by a fixed 260KHz oscillator. In a standard step-down application the duty cycle (Time ON/Time OFF) of the power switch is proportional to the ratio of the power supply output voltage to the input voltage. The voltage on pin 1 switches between Vin (switch ON) and below ground by the voltage drop of the external Schottky diode (switch OFF).

#### PIN 2 - Input

The input voltage for the power supply is connected to pin 2. In addition to providing energy to the load the input voltage also provides bias for the internal circuitry of the LM2679. For guaranteed performance the input voltage must be in the range of 8V to 40V. For best performance of the power supply the input pin should always be bypassed with an input capacitor located close to pin 2.

#### PIN 3 - C Boost

A capacitor must be connected from pin 3 to the switch output, pin 1. This capacitor boosts the gate drive to the internal MOSFET above Vin to fully turn it ON. This minimizes con-

#### **DESIGN CONSIDERATIONS**

duction losses in the power switch to maintain high efficiency. The recommended value for C Boost is  $0.01 \mu F.$ 

### PIN 4 - Ground

This is the ground reference connection for all components in the power supply. In fast-switching, high-current applications such as those implemented with the LM2679, it is recommended that a broad ground plane be used to minimize signal coupling throughout the circuit

#### PIN 5 - Current Adjust

A key feature of the LM2679 is the ability to tailor the peak switch current limit to a level required by a particular application. This alleviates the need to use external components that must be physically sized to accommodate current levels (under shorted output conditions for example) that may be much higher than the normal circuit operating current requirements.

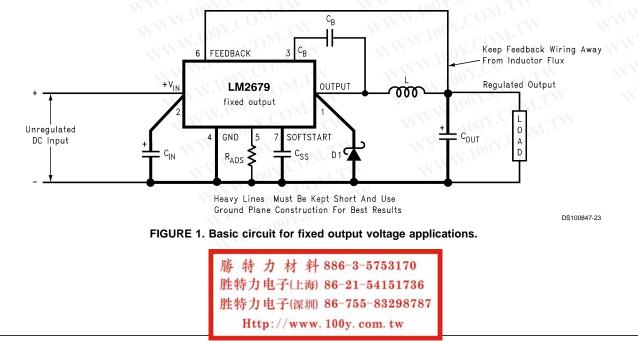
A resistor connected from pin 5 to ground establishes a current ( $I_{(pin 5)} = 1.2V / R_{ADJ}$ ) that sets the peak current through the power switch. The maximum switch current is fixed at a level of 37,125 /  $R_{ADJ}$ .

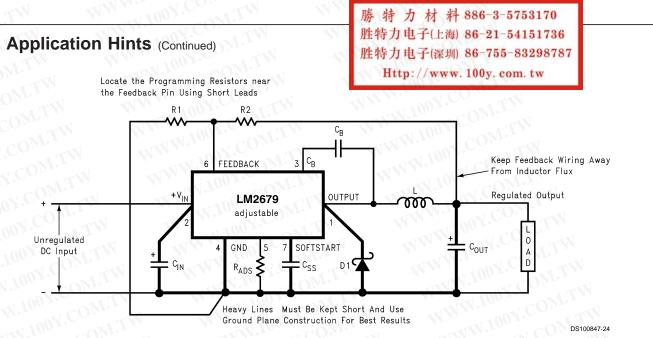
#### **PIN 6 - Feedback**

This is the input to a two-stage high gain amplifier, which drives the PWM controller. It is necessary to connect pin 6 to the actual output of the power supply to set the dc output voltage. For the fixed output devices (3.3V, 5V and 12V outputs), a direct wire connection to the output is all that is required as internal gain setting resistors are provided inside the LM2679. For the adjustable output version two external resistors are required to set the dc output voltage. For stable operation of the power supply it is important to prevent coupling of any inductor flux to the feedback input.

#### PIN 7 - Softstart

A capacitor connected from pin 7 to ground allows for a slow turn-on of the switching regulator. The capacitor sets a time delay to gradually increase the duty cycle of the internal power switch. This can significantly reduce the amount of surge current required from the input supply during an abrupt application of the input voltage. If softstart is not required this pin should be left open circuited.





#### FIGURE 2. Basic circuit for adjustable output voltage applications

Power supply design using the LM2679 is greatly simplified by using recommended external components. A wide range of inductors, capacitors and Schottky diodes from several manufacturers have been evaluated for use in designs that cover the full range of capabilities (input voltage, output voltage and load current) of the LM2679. A simple design procedure using nomographs and component tables provided in this data sheet leads to a working design with very little effort. Alternatively, the design software, *LM267X Made Simple* (version 2.0), can also be used to provide instant component selection, circuit performance calculations for evaluation, a bill of materials component list and a circuit schematic.

The individual components from the various manufacturers called out for use are still just a small sample of the vast array of components available in the industry. While these components are recommended, they are not exclusively the only components for use in a design. After a close comparison of component specifications, equivalent devices from other manufacturers could be substituted for use in an application.

Important considerations for each external component and an explanation of how the nomographs and selection tables were developed follows.

#### INDUCTOR

LM2679

The inductor is the key component in a switching regulator. For efficiency the inductor stores energy during the switch ON time and then transfers energy to the load while the switch is OFF.

Nomographs are used to select the inductance value required for a given set of operating conditions. The nomographs assume that the circuit is operating in continuous mode (the current flowing through the inductor never falls to zero). The magnitude of inductance is selected to maintain a maximum ripple current of 30% of the maximum load current. If the ripple current exceeds this 30% limit the next larger value is selected.

The inductors offered have been specifically manufactured to provide proper operation under all operating conditions of input and output voltage and load current. Several part types are offered for a given amount of inductance. Both surface mount and through-hole devices are available. The inductors from each of the three manufacturers have unique characteristics.

Renco: ferrite stick core inductors; benefits are typically lowest cost and can withstand ripple and transient peak currents above the rated value. These inductors have an external magnetic field, which may generate EMI.

Pulse Engineering: powdered iron toroid core inductors; these also can withstand higher than rated currents and, being toroid inductors, will have low EMI.

Coilcraft: ferrite drum core inductors; these are the smallest physical size inductors and are available only as surface mount components. These inductors also generate EMI but less than stick inductors.

### **OUTPUT CAPACITOR**

The output capacitor acts to smooth the dc output voltage and also provides energy storage. Selection of an output capacitor, with an associated equivalent series resistance (ESR), impacts both the amount of output ripple voltage and stability of the control loop.

The output ripple voltage of the power supply is the product of the capacitor ESR and the inductor ripple current. The capacitor types recommended in the tables were selected for having low ESR ratings.

In addition, both surface mount tantalum capacitors and through-hole aluminum electrolytic capacitors are offered as solutions.

Impacting frequency stability of the overall control loop, the output capacitance, in conjunction with the inductor, creates a double pole inside the feedback loop. In addition the capacitance and the ESR value create a zero. These frequency response effects together with the internal frequency compensation circuitry of the LM2679 modify the gain and phase shift of the closed loop system.

As a general rule for stable switching regulator circuits it is desired to have the unity gain bandwidth of the circuit to be limited to no more than one-sixth of the controller switching frequency. With the fixed 260KHz switching frequency of the LM2679, the output capacitor is selected to provide a unity gain bandwidth of 40KHz maximum. Each recommended capacitor value has been chosen to achieve this result.

In some cases multiple capacitors are required either to reduce the ESR of the output capacitor, to minimize output ripple (a ripple voltage of 1% of Vout or less is the assumed performance condition), or to increase the output capacitance to reduce the closed loop unity gain bandwidth (to less than 40KHz). When parallel combinations of capacitors are required it has been assumed that each capacitor is the exact same part type.

The RMS current and working voltage (WV) ratings of the output capacitor are also important considerations. In a typical step-down switching regulator, the inductor ripple current (set to be no more than 30% of the maximum load current by the inductor selection) is the current that flows through the output capacitor. The capacitor RMS current rating must be greater than this ripple current. The voltage rating of the output capacitor should be greater than 1.3 times the maximum output voltage of the power supply. If operation of the system at elevated temperatures is required, the capacitor voltage rating may be de-rated to less than the nominal room temperature rating. Careful inspection of the manufacturer's specification for de-rating of working voltage with temperature is important.

#### INPUT CAPACITOR

Fast changing currents in high current switching regulators place a significant dynamic load on the unregulated power source. An input capacitor helps to provide additional current to the power supply as well as smooth out input voltage variations.

Like the output capacitor, the key specifications for the input capacitor are RMS current rating and working voltage. The RMS current flowing through the input capacitor is equal to one-half of the maximum dc load current so the capacitor should be rated to handle this. Paralleling multiple capacitors proportionally increases the current rating of the total capacitance. The voltage rating should also be selected to be 1.3 times the maximum input voltage. Depending on the unregulated input power source, under light load conditions the maximum input voltage could be significantly higher than normal operation and should be considered when selecting an input capacitor.

The input capacitor should be placed very close to the input pin of the LM2679. Due to relative high current operation with fast transient changes, the series inductance of input connecting wires or PCB traces can create ringing signals at the input terminal which could possibly propagate to the output or other parts of the circuitry. It may be necessary in some designs to add a small valued (0.1 $\mu$ F to 0.47 $\mu$ F) ceramic type capacitor in parallel with the input capacitor to prevent or minimize any ringing.

#### CATCH DIODE

When the power switch in the LM2679 turns OFF, the current through the inductor continues to flow. The path for this current is through the diode connected between the switch output and ground. This forward biased diode clamps the switch output to a voltage less than ground. This negative voltage must be greater than -1V so a low voltage drop (particularly at high current levels) Schottky diode is recommended. Total efficiency of the entire power supply is significantly impacted by the power lost in the output catch diode. The average current through the catch diode is dependent on the switch duty cycle (D) and is equal to the load current times (1-D). Use of a diode rated for much higher current than is required by the actual application helps to minimize the voltage drop and power loss in the diode.

During the switch ON time the diode will be reversed biased by the input voltage. The reverse voltage rating of the diode should be at least 1.3 times greater than the maximum input voltage.

#### BOOST CAPACITOR

The boost capacitor creates a voltage used to overdrive the gate of the internal power MOSFET. This improves efficiency by minimizing the on resistance of the switch and associated power loss. For all applications it is recommended to use a 0.01µF/50V ceramic capacitor.

### R<sub>ADJ</sub>, ADJUSTABLE CURRENT LIMIT

A key feature of the LM2679 is the ability to control the peak switch current. Without this feature the peak switch current would be internally set to 7A or higher to accommodate 5A load current designs. This requires that both the inductor (which could saturate with excessively high currents) and the catch diode be able to safely handle up to 7A which would be conducted under load fault conditions.

If an application only requires a load current of 3A or 4A the peak switch current can be set to a limit just over the maximum load current with the addition of a single programming resistor. This allows the use of less powerful and more cost effective inductors and diodes.

The peak switch current is equal to a factor of 37,125 divided by R<sub>ADJ</sub>. A resistance of 5.6K $\Omega$  sets the current limit to typically 6.3A and an R<sub>ADJ</sub> of 8.25K $\Omega$  reduces the maximum current to approximately 4.4A. For predictable control of the current limit it is recommended to keep the peak switch current greater than 3A. For lower current applications a 3A switching regulator with adjustable current limit, the LM2673, is available.

When the power switch reaches the current limit threshold it is immediately turned OFF and the internal switching frequency is reduced. This extends the OFF time of the switch to prevent a steady state high current condition. As the switch current falls below the current limit threshold, the switch will turn back ON. If a load fault continues, the switch will again exceed the threshold and switch back OFF. This will result in a low duty cycle pulsing of the power switch to minimize the overall fault condition power dissipation.

#### **Css SOFTSTART CAPACITOR**

This optional capacitor controls the rate at which the LM2679 starts up at power on. The capacitor is charged linearly by an internal current source. This voltage ramp gradually increases the duty cycle of the power switch until it reaches the normal operating duty cycle defined primarily by the ratio of the output voltage to the input voltage. The softstart turn-on time is programmable by the selection of Css. The formula for selecting a softstart capacitor is:

$$C_{SS} \cong (I_{SST} \cdot t_{SS}) / [V_{SST} + 2.6V \cdot (\frac{V_{OUT} + V_{SCHOTTKY}}{V_{IN}})]$$

Where:

I<sub>SST</sub> = Softstart Current, 3.7µA typical

 $t_{SS}$  = Softstart time, from design requirements

V<sub>SST</sub> = Softstart Threshold Voltage, 0.63V typical

V<sub>OUT</sub> = Output Voltage, from design requirements

 $V_{\text{SCHOTTKY}}$  = Schottky Diode Voltage Drop, typically 0.5V  $V_{\text{IN}}$  = Maximum Input Voltage, from design requirements If this feature is not desired, leave the Softstart pin (pin 7) open circuited

#### SIMPLE DESIGN PROCEDURE

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胜华	侍力	电	子(深	[圳]	86-755-83298787
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Using the nomographs and tables in this data sheet (or use the available design software at http://www.national.com) a complete step-down regulator can be designed in a few simple steps.

Step 1: Define the power supply operating conditions:

Required output voltage

Maximum DC input voltage

Maximum output load current

**Step 2:** Set the output voltage by selecting a fixed output LM2679 (3.3V, 5V or 12V applications) or determine the required feedback resistors for use with the adjustable LM2679–ADJ

**Step 3:** Determine the inductor required by using one of the four nomographs, *Figure 3* through *Figure 6*. Table 1 provides a specific manufacturer and part number for the inductor.

**Step 4:** Using Table 3 (fixed output voltage) or Table 6 (adjustable output voltage), determine the output capacitance required for stable operation. Table 2 provides the specific capacitor type from the manufacturer of choice.

**Step 5:** Determine an input capacitor from Table 4 for fixed output voltage applications. Use Table 2 to find the specific capacitor type. For adjustable output circuits select a capacitor from Table 2 with a sufficient working voltage (WV) rating greater than Vin max, and an rms current rating greater than one-half the maximum load current (2 or more capacitors in parallel may be required).

**Step 6:** Select a diode from Table 5. The current rating of the diode must be greater than I load max and the Reverse Voltage rating must be greater than Vin max.

Step 7: Include a  $0.01\mu$ F/50V capacitor for Cboost in the design and then determine the value of a softstart capacitor if desired.

**Step 8:** Define a value for  $R_{ADJ}$  to set the peak switch current limit to be at least 20% greater than lout max to allow for at least 30% inductor ripple current (±15% of lout). For designs that must operate over the full temperature range the switch current limit should be set to at least 50% greater than lout max (1.5 x l<sub>out</sub> max).

#### FIXED OUTPUT VOLTAGE DESIGN EXAMPLE

A system logic power supply bus of 3.3V is to be generated from a wall adapter which provides an unregulated DC voltage of 13V to 16V. The maximum load current is 4A. A soft-start delay time of 50mS is desired. Through-hole components are preferred.

Step 1: Operating conditions are:

Vout = 3.3V

Vin max = 16V

lload max = 4A

**Step 2:** Select an LM2679T-3.3. The output voltage will have a tolerance of

 $\pm 2\%$  at room temperature and  $\pm 3\%$  over the full operating temperature range.

**Step 3:** Use the nomograph for the 3.3V device *,Figure 3.* The intersection of the 16V horizontal line ( $V_{in}$  max) and the 4A vertical line ( $I_{load}$  max) indicates that L46, a 15µH inductor, is required.

From Table 1, L46 in a through-hole component is available from Renco with part number RL-1283-15-43.

**Step 4:** Use Table 3 to determine an output capacitor. With a 3.3V output and a  $15\mu$ H inductor there are four through-hole output capacitor solutions with the number of same type capacitors to be paralleled and an identifying capacitor code given. Table 2 provides the actual capacitor characteristics. Any of the following choices will work in the circuit:

2 x 220µF/10V Sanyo OS-CON (code C5)

2 x 820µF/16V Sanyo MV-GX (code C5)

1 x 3900µF/10V Nichicon PL (code C7)

2 x 560µF/35V Panasonic HFQ (code C5)

**Step 5:** Use Table 4 to select an input capacitor. With 3.3V output and 15 $\mu$ H there are three through-hole solutions. These capacitors provide a sufficient voltage rating and an rms current rating greater than 2A (1/2 I<sub>load</sub> max). Again using Table 2 for specific component characteristics the following choices are suitable:

2 x 680µF/63V Sanyo MV-GX (code C13)

1 x 1200µF/63V Nichicon PL (code C25)

1 x 1500µF/63V Panasonic HFQ (code C16)

**Step 6:** From Table 5 a 5A or more Schottky diode must be selected. For through-hole components only 40V rated diodes are indicated and 4 part types are suitable:

1N5825 MBR745

80SQ045

6TQ045

010040

Step 7: A  $0.01\mu$ F capacitor will be used for Cboost. For the 50mS softstart delay the following parameters are to be used:

I<sub>SST</sub>: 3.7µА

t<sub>ss</sub>: 50mS

V<sub>SST</sub>: 0.63V

V<sub>OUT</sub>: 3.3V

V<sub>SCHOTTKY</sub>: 0.5V

V<sub>IN</sub>: 16V

Using Vin max ensures that the softstart delay time will be at least the desired 50mS.

Using the formula for Css a value of  $0.148\mu$ F is determined to be required. Use of a standard value  $0.22\mu$ F capacitor will produce more than sufficient softstart delay.

Step 8: Determine a value for  $R_{ADJ}$  to provide a peak switch current limit of at least 4A plus 50% or 6A.

$$R_{ADJ} = \frac{37,125}{6A} = 6.1875 \text{ k}\Omega$$

Use a value of  $6.2K\Omega$ .

#### ADJUSTABLE OUTPUT DESIGN EXAMPLE

In this example it is desired to convert the voltage from a two battery automotive power supply (voltage range of 20V to 28V, typical in large truck applications) to the 14.8VDC alternator supply typically used to power electronic equipment from single battery 12V vehicle systems. The load current required is 3.5A maximum. It is also desired to implement the power supply with all surface mount components. Softstart is not required.

Step 1: Operating conditions are:

Vout = 14.8V Vin max = 28V Iload max = 3.5A

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**Step 2:** Select an LM2679S-ADJ. To set the output voltage to 14.9V two resistors need to be chosen (R1 and R2 in *Figure 2*). For the adjustable device the output voltage is set by the following relationship:

$$V_{OUT} = V_{FB} \left( 1 + \frac{R_2}{R_1} \right)$$

Where  $V_{FB}$  is the feedback voltage of typically 1.21V.

A recommended value to use for R1 is 1K. In this example then R2 is determined to be:

$$R_{2} = R_{1} \left( \frac{V_{OUT}}{V_{FB}} - 1 \right) = 1 k\Omega \left( \frac{14.8V}{1.21V} - 1 \right)$$

 $R2 = 11.23K\Omega$ 

The closest standard 1% tolerance value to use is  $11.3K\Omega$ This will set the nominal output voltage to 14.88V which is within 0.5% of the target value.

**Step 3:** To use the nomograph for the adjustable device, *Figure 6*, requires a calculation of the inductor Volt•microsecond constant (E•T expressed in V• $\mu$ S) from the following formula:

$$E \cdot T = (V_{IN(MAX)} - V_{OUT} - V_{SAT}) \cdot \frac{V_{OUT} + V_D}{V_{IN(MAX)} - V_{SAT} + V_D} \cdot \frac{1000}{260} (V \cdot$$

where  $V_{\text{SAT}}$  is the voltage drop across the internal power switch which is  $R_{ds(ON)}$  times  $I_{load}$ . In this example this would be typically  $0.12\Omega$  x 3.5A or 0.42V and  $V_D$  is the voltage drop across the forward bisased Schottky diode, typically 0.5V. The switching frequency of 260KHz is the nominal value to use to estimate the ON time of the switch during which energy is stored in the inductor.

For this example E•T is found to be:

$$E \cdot T = (28 - 14.8 - 0.42) \cdot \frac{14.8 + 0.5}{28 - 0.42 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$$
  
$$E \cdot T = (12.78V) \cdot \frac{15.3V}{28.08V} \cdot 3.85 (V \cdot \mu s) = 26.8 (V \cdot \mu s)$$

Using *Figure 6*, the intersection of  $27V \cdot \mu S$  horizontally and the 3.5A vertical line (I<sub>load</sub> max) indicates that L48, a  $47\mu H$  inductor, or L49, a 33 $\mu H$  inductor could be used. Either inductor will be suitable, but for this example selecting the larger inductance will result in lower ripple current.

From Table 1, L48 in a surface mount component is available from Pulse Engineering with part number P0848.

**Step 4:** Use Table 6 to determine an output capacitor. With a 14.8V output the 12.5 to 15V row is used and with a 47 $\mu$ H inductor there are three surface mount output capacitor solutions. Table 2 provides the actual capacitor characteristics based on the C Code number. Any of the following choices can be used:

- 1 x 33µF/20V AVX TPS (code C6)
- 1 x 47µF/20V Sprague 594 (code C8)
- 1 x 47µF/20V Kemet T495 (code C8)

**Important Note:** When using the adjustable device in low voltage applications (less than 3V output), if the nomograph, Figure 6, selects an inductance of  $22\mu$ H or less, Table 6 does not provide an output capacitor solution. With these conditions the number of output capacitors required for stable operation becomes impractical. It is recommended to use either a  $33\mu$ H or  $47\mu$ H inductor and the output capacitors from Table 6.

**Step 5:** An input capacitor for this example will require at least a 35V WV rating with an rms current rating of 1.75A (1/2 lout max). From Table 2 it can be seen that C12, a  $33\mu$ F/35V capacitor from Sprague, has the highest voltage/current rating of the surface mount components and that two of these capacitor in parallel will be adquate.

**Step 6:** From Table 5 a 5A or more Schottky diode must be selected. For surface mount diodes with a margin of safety on the voltage rating one of two diodes can be used: MBRD1545CT

6TQ045S

Step 7: A 0.01µF capacitor will be used for Cboost.

The softstart pin will be left open circuited.

**Step 8:** Determine a value for  $R_{ADJ}$  to provide a peak switch current limit of at least 3.5A plus 50% or 5.25A.

$$ADJ = \frac{37,125}{5.2A} = 7.14 \text{ k}\Omega$$

Use a value of  $7.15 \text{K}\Omega$ .

μs

R

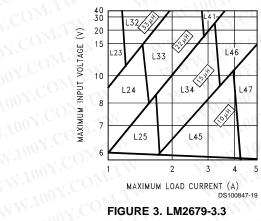
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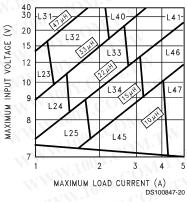
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INDUCTOR VALUE SELECTION GUIDES (For Continuous Mode Operation)









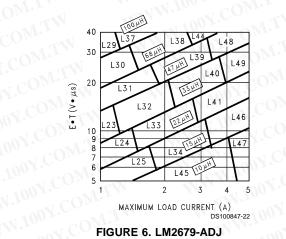
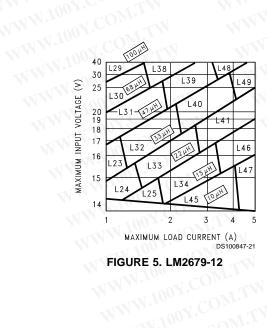


FIGURE 6. LM2679-ADJ



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Inductor	Inductance	Current	Ren	co	Pulse Er	ngineering	Coilcraft
Reference Number	(µH)	(A)	Through Hole	Surface Mount	Through Hole	Surface Mount	Surface Mount
L23	33	1.35	RL-5471-7	RL1500-33	PE-53823	PE-53823S	DO3316-333
L24	22	1.65	RL-1283-22-43	RL1500-22	PE-53824	PE-53824S	DO3316-223
L25	15	2.00	RL-1283-15-43	RL1500-15	PE-53825	PE-53825S	DO3316-153
L29	100	1.41	RL-5471-4	RL-6050-100	PE-53829	PE-53829S	DO5022P-104
L30	68	1.71	RL-5471-5	RL6050-68	PE-53830	PE-53830S	DO5022P-683
L31	47	2.06	RL-5471-6	RL6050-47	PE-53831	PE-53831S	DO5022P-473
L32	33	2.46	RL-5471-7	RL6050-33	PE-53932	PE-53932S	DO5022P-333
L33	22	3.02	RL-1283-22-43	RL6050-22	PE-53933	PE-53933S	DO5022P-223
L34	15	3.65	RL-1283-15-43	N.Y C	PE-53934	PE-53934S	DO5022P-153
L38	68	2.97	RL-5472-2	W.Inv	PE-54038	PE-54038S	WW.W
L39	47	3.57	RL-5472-3	100X.	PE-54039	PE-54039S	1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1
L40	33	4.26	RL-1283-33-43	Cont - VP	PE-54040	PE-54040S	NN -100Y
L41	22	5.22	RL-1283-22-43	WW.	PE-54041	P0841	WWW.
L44	68	3.45	RL-5473-3	N.100	PE-54044	<u> </u>	NTV.100
L45	10	4.47	RL-1283-10-43	W	01	P0845	DO5022P-103HC
L46	15	5.60	RL-1283-15-43	WA	MY.COM	P0846	DO5022P-153HC
L47	10	5.66	RL-1283-10-43	NW.	00 TC	P0847	DO5022P-103HC
L48	47	5.61	RL-1282-47-43		100 <u>-</u> . c(	P0848	War-
L49	33	5.61	RL-1282-33-43	A VI	. 10 <del>0</del> 1.0	P0849	2

## WWW.100Y **Table 1. Inductor Manufacturer Part Numbers**

## Inductor Manufacturer Contact Numbers

Coilcraft	Phone	(800) 322-2645
	FAX	(708) 639-1469
oilcraft, Europe 🔨	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
ulse Engineering	Phone	(619) 674-8100
	FAX	(619) 674-8262
ulse Engineering,	Phone	+353 93 24 107
urope	FAX	+353 93 24 459
enco Electronics	Phone	(800) 645-5828
	FAX	(516) 586-5562

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WWW.100Y.C

Capacitor		W	N.100	Su	face Mo	ount	WW	N.1-	V.C
Reference	AVX	( TPS Se	eries	Sprag	ue 594D	Series	Keme	et T495 S	Series
Code	С (µF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)	С (µF)	WV (V)	Irms (A)
C1	330	6.3	1.15	120	6.3	1.1	100	6.3	0.82
C2	100	10	1.1	220	6.3	1.4	220	6.3	1.1
C3	220	10 🖪	1.15	68	10	1.05	330	6.3	1.1
C4	47	16	0.89	150	10	1.35	100	10	1.1
C5	100	16	1.15	47	16	1	150	10	1.1
C6	33	20	0.77	100	16	1.3	220	10	1.1
C7	68	20	0.94	180	16	1.95	33	20	0.78
C8	22	25	0.77	47	20	1.15	47	20	0.94
C9	10	35	0.63	33	25	1.05	68	20	0.94
C10	22	35	0.66	68	25	1.6	10	35	0.63
C11	COL	W		15	35	0.75	22	35	0.63
C12	1 CON	1.1		33	35	v 10 <sup>1</sup>	4.7	50	0.66
C13		1.1		15	50	0.9	M.L		

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## Table 2. Input and Output Capacitor Codes (continued)

CO1	Through Hole											
Capacitor Reference		OS-CO Series	N SA	Sanyo	Sanyo MV-GX Series		Nichio	Nichicon PL Series			nic HFQ	Series
Code	С (µF)	N WV (V)	Irms (A)	С (µF)	WV (V)	Irms (A)	(μF)	WV (V)	Irms (A)	C (µF)	WV (V)	Irms (A)
C1	47	6.3		1000	6.3	0.8	680	10	0.8	82	35	0.4
C2	150	6.3	1.95 <	270	16	0.6	820	10	0.98	120	35	0.44
C3	330	6.3	2.45	470	16	0.75	1000	10	1.06	220	35	0.76
C4	100	10	1.87	560	16	0.95	1200	10	1.28	330	35	1.01
C5	220	10	2.36	820	16	1.25	2200	10	1.71	560	35	1.4
C6	33	16	0.96	1000	16	1.3	3300	10	2.18	820	35	1.62
C7	100	16	1.92	150	35	0.65	3900	10	2.36	1000	35	1.73
C8	150	16	2.28	470	35	1.3	6800	10	2.68	2200	35	2.8
C9	100	20	2.25	680 🔨	35	1.4	180	16	0.41	56	50	0.36
C10	47	25	2.09	1000	35	1.7	270	16	0.55	100	50	0.5
C11	W.100 *	100	1.1	220	63	0.76	470	16	0.77	220	50	0.92
C12	100		M.T.W	470	63	1.2	680	16	1.02	470	50	1.44
C13	N. C.	N.CU	T	680	63	1.5	820	16	1.22	560	50	1.68
C14	WW.IC	NC C	DAT	1000	63	1.75	1800	16	1.88	1200	50	2.22
C15	.W.I	00 -	·0M.			WW.	220	25	0.63	330	63	1.42
C16		100X.	Mo	L.M.			220	35	0.79	1500	63	2.51
C17	MWN	LOON	CO	WT		NN.	560	35	1.43		M.	110
C18	VWI	1.102	<sup>a</sup> CON	N/m		WW	2200	35	2.68		WW	VY
C19		N.100	- c0	V.L	т		150	50	0.82	KT.		W.r
C20	AL.	110	01.0	M.TV	N		220	50	1.04			AN.
C21	W	14	NY.C	1	N	W	330	50	1.3		N	
C22		MM'	<b>N</b>	ON.	N		100	63	0.75	W	<	NN N
C23		W	100	COM.			390	63	1.62			W
C24	1		100%		TN		820	63	2.22	UIN		NY .
C25		WWW		I.COM	Wn.		1200	63	2.51	WT .		N.V.

## LM2679

## **Capacitor Manufacturer Contact Numbers**

Nichicon	Phone	(847) 843-7500
	FAX	(847) 843-2798
Panasonic	Phone 🔨	(714) 373-7857
	FAX	(714) 373-7102
AVX	Phone	(845) 448-9411
	FAX	(845) 448-1943
Sprague/Vishay	Phone	(207) 324-4140
	FAX	(207) 324-7223
Sanyo	Phone	(619) 661-6322
	FAX	(619) 661-1055
Kemet	Phone	(864) 963-6300
	FAX	(864) 963-6521

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## Application Hints (Continued)

## Table 3. Output Capacitors for Fixed Output Voltage Application

Output		NN.10	COM.	Surfac	e Mount	N	V.COm		
Voltage	Inductance (µH)	AVX TPS Series		Sprague 594D Series		Kemet T	495 Series		
(V)		No.	C Code	No.	C Code	No.	C Code		
coM.	10	5	C1	5	C1	5	C2		
2.2	15	4	C1	4	C1	4	C3		
3.3	22	3	C2	2	C7	3	C4		
A CON	33	1	C1	2	C7	3	C4		
oy.col	10	4	C2	4	C6	4	C4		
	15	3	C3	2	C7	3	C5		
5 C	22	3	C2	2	C7	3	C4		
	33	2	C2	2	C3	2	C4		
	47	2	C2	1	C7	2	C4		
1001	10	4	C5	3	C6	5	C9		
	15	3	C5	2	C7	4	C9		
	22	2	C5	2	C6	3	C8		
12	33	2	C5	1.1.4	C7	3	C8		
	47	2	C4	10	C6	2	C8		
	68	1	C5	1	C5	2	C7		
	100	1	C4	1	C5	1	C8		
	1002.0	M		M.W.	Throug	gh Hole		NN.10	~1 CO1
Output Voltage	Inductance (µH)	Sanyo OS-CON SA Series		Sanyo MV-GX Series		Nichicon PL Series		Panasonic HFQ Series	
(V)				NET	0.0.1	No.		No.	C Code
	W.Ive	No.	C Code	No.	C Code	INO.	C Code	110.	
	10	<b>No.</b> 2	C Code C5	<b>NO.</b> 2	C Code C6	NO. 1	C Code C8	2	C6
WW WW	10 15						1		C6 C5
3.3		2	C5	2	C6	01	C8	2	
WW WW	15	2	C5 C5	2 2	C6 C5		C8 C7	2 2	C5
WW WW	15 22	2 2 1	C5 C5 C5	2 2 1	C6 C5 C10	1 1 1	C8 C7 C5	2 2 1	C5 C7
WW WW	15 22 33	2 2 1 1	C5 C5 C5 C5 C5	2 2 1 1	C6 C5 C10 C10	1 1 1 1	C8 C7 C5 C5	2 2 1 1	C5 C7 C7
WW WW	15 22 33 10	2 2 1 1 2	C5 C5 C5 C5 C5 C4	2 2 1 1 2 2	C6 C5 C10 C10 C5 C5	1 1 1 1 1	C8 C7 C5 C5 C5 C6	2 2 1 1 2 2	C5 C7 C7 C5
3.3	15 22 33 10 15	2 2 1 1 2 1 2 1	C5 C5 C5 C5 C5 C4 C5	2 2 1 1 2 1 1	C6 C5 C10 C10 C5 C10	1 1 1 1 1 1 1	C8 C7 C5 C5 C6 C5 C5	2 2 1 1 2 1 2 1	C5 C7 C7 C5 C7
3.3	15           22           33           10           15           22	2 2 1 1 2 1 2 1 1	C5 C5 C5 C5 C5 C4 C5 C5 C5	2 2 1 1 2 1 2 1 1	C6 C5 C10 C10 C5 C10 C9	1 1 1 1 1 1 1 1	C8 C7 C5 C5 C6 C5 C5 C5 C5	2 2 1 1 2 1 2 1 1	C5 C7 C7 C5 C5 C7 C5
3.3	15           22           33           10           15           22           333	2 2 1 1 2 1 1 1 1 1	C5 C5 C5 C5 C4 C5 C5 C5 C5 C4	2 2 1 1 2 1 1 1 1 1	C6           C5           C10           C9           C5	1 1 1 1 1 1 1 1 1	C8 C7 C5 C5 C6 C5 C5 C5 C5 C4	2 2 1 1 2 1 1 1 1 1	C5 C7 C7 C5 C7 C5 C5 C4
3.3	15           22           33           10           15           22           33           47	2 2 1 1 2 1 1 1 1 1 1	C5           C5           C5           C5           C5           C5           C5           C5           C4           C5           C4           C4           C4           C4           C4	2 2 1 1 2 1 1 1 1 1 1	C6           C5           C10           C5           C10           C5           C10           C5           C10           C5           C10           C5           C10           C9           C5           C4	1 1 1 1 1 1 1 1 1 1	C8 C7 C5 C5 C6 C5 C5 C5 C5 C4 C2	2 2 1 1 2 1 1 1 1 2	C5 C7 C7 C5 C7 C5 C7 C5 C4 C4
3.3	15           22           33           10           15           22           33           47           10	2 2 1 1 2 1 1 1 1 1 2	C5           C5           C5           C5           C5           C4           C5           C5           C4           C5           C4           C5           C4           C7	2 2 1 1 2 1 1 1 1 1 1 1 1	C6           C5           C10           C5           C10           C5           C10           C5           C10           C5           C10           C9           C5           C4           C10	1 1 1 1 1 1 1 1 1 1 1	C8           C7           C5           C5           C6           C5           C5           C5           C4           C2           C14	2 2 1 1 2 1 1 1 1 2 2 2	C5 C7 C7 C5 C7 C5 C7 C5 C4 C4 C4
3.3	15           22           33           10           15           22           33           47           10           15	2 2 1 1 2 1 1 1 1 1 2 1 2 1	C5           C5           C5           C5           C5           C4           C5           C5           C4           C5           C4           C5           C4           C5           C4           C7           C8	2 2 1 1 2 1 1 1 1 1 1 1 1 1	C6           C5           C10           C5           C10           C5           C10           C5           C10           C9           C5           C4           C10           C6	1 1 1 1 1 1 1 1 1 1 1 1 1	C8           C7           C5           C6           C5           C6           C5           C4           C2           C14           C17	2 2 1 1 2 1 1 1 2 2 2 1	C5 C7 C7 C5 C7 C5 C4 C4 C4 C4 C4 C5
3.3	15           22           33           10           15           22           33           47           10           15           22           33           47           10           15           22	2 2 1 1 2 1 1 1 1 1 2 1 1 1 1	C5           C5           C5           C5           C4           C5           C4           C4           C4           C4           C4           C4           C4           C4           C7           C8           C7	2 2 1 1 2 1 1 1 1 1 1 1 1 1 1	C6           C5           C10           C5           C10           C5           C10           C9           C5           C4           C10           C6           C5	1 1 1 1 1 1 1 1 1 1 1 1 1 1	C8           C7           C5           C5           C6           C5           C5           C4           C2           C14           C17           C13	2 2 1 1 2 1 1 1 2 2 2 1 1 1	C5 C7 C7 C5 C7 C5 C4 C4 C4 C4 C4 C5 C5
3.3	15           22           33           10           15           22           33           47           10           15           22           33           33           33	2 2 1 1 2 1 1 1 1 2 1 1 2 1 1 1 1	C5           C5           C5           C5           C5           C4           C5           C4           C4           C4           C4           C4           C7           C8           C7           C7           C7	2 2 1 1 2 1 1 1 1 1 1 1 1 1 1 1 1	C6           C5           C10           C5           C10           C5           C10           C9           C5           C4           C10           C6           C5           C4	1 1 1 1 1 1 1 1 1 1 1 1 1 1	C8           C7           C5           C5           C6           C5           C4           C2           C14           C17           C13           C12	2 2 1 1 2 1 1 1 2 2 2 1 1 1 1 1	C5           C7           C5           C7           C5           C7           C5           C4           C4           C5           C5           C5           C4           C4           C5           C5           C5           C5           C5           C5           C5           C5           C5           C4

No. represents the number of identical capacitor types to be connected in parallel

C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.

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## Table 4. Input Capacitors for Fixed Output Voltage Application

(Assumes worst case maximum input voltage and load current for a given inductance value)

NY.C	MILIA		100	1003					
Output Voltage (V)	Inductance (µH)	AVX TPS Series			ue 594D eries	Kemet 1	495 Series		
(v)	WTT	No.	C Code	No.	C Code	No.	C Code		
V. Lou	10	3	C7	2	C10	3	C9		
N100 1	15	*	*	3	C13	4	C12		
3.3	22	*	*	2	C13	3	C12		
	33	*	*	2	C13	3	C12		
WW.L	10	3	C4	2	C6	3	C9		
	15	4	C9	3	C12	4	C10		
5	22	*	*	3	C13	4	C12		
	33	*	*	2	C13	3	C12		
	47	*	*	1	C13	2	C12		
VI	10	4	C9	2	C10	4	C10		
	15	4	C8	2	C10	4	C10		
	22	4	C9	3	C12	4	C10		
12	33	<b>*</b>	*	3	C13	4	C12		
	47	*	*	2	C13	3	C12		
	68	*	*77	2	C13	2	C12		
	100	*	*	1 🔨	C13	2	C12		
	WWW.I	02.VC	V	WW.L					
Output Voltage	Inductance (µH)	Sanyo OS-CON SA Series		Sanyo MV-GX Series		Nichicon PL Series		Panasonic HFQ Series	
(V)	WWW	No.	C Code	No.	C Code	No.	C Code	No.	C Code
	10	2	C9	2	C8	1.	C18	1	C8
	15	*	*	2	C13	1	C25	1	C16
3.3	22	* 00	*	1	C14	1103	C24	1	C16
	33 🔨	*	N. (*	11	C14	1.00	C24	1	C16
	10	2	C7	2	C8	1	C25	~1	C8
	15	*	*	2	C8	.IV	C25	1	C8
5	22	*	*	2	C13	1	C25	1.1	C16
	33	*	*	1	C14	1	C23	1	C13
	47	*	*	01	C12	1	C19	1	C11
	10	2	C10	2	C8	1	C18	ONT	C8
	15	2	C10	2	C8	1	C18	1.1	C8
	22	*	*	2	C8	1	C18	1	C8
12	33	*	*	2	C12	1	C24	CO	C14
	47	*	*	1	C14	1	C23		C13
	68	*	*	1	C13	1	C21	1	C15
	4.0.0				011	1 .	000		011

\* Check voltage rating of capacitors to be greater than application input voltage.

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No. represents the number of identical capacitor types to be connected in parallel

C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.

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C11

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C22

1

LM2679

C Code C8 C16 C16 C16 C8 C8 C16 C13 C11 C8 C8 C8 C14 C13 C15 C11

## NWW.1007 Table 5. Schottky Diode Selection Table

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Reverse	Surfa	ace Mount	Throug	gh Hole
Voltage (V)	3A	5A or More	3A	5A or More
20V	SK32	WWW.100	1N5820 SR302	LM
30V	SK33 30WQ03F	MBRD835L	1N5821 31DQ03	WT.
40V	SK34 30BQ040 30WQ04F MBRS340 MBRD340	MBRD1545CT 6TQ045S	1N5822 MBR340 31DQ04 SR403	1N5825 MBR745 80SQ045 6TQ045
0V or Nore	SK35 30WQ05F		MBR350 31DQ05 SR305	COM.

## **Diode Manufacturer Contact Numbers** DOY.COM.TW

International Rectifier	Phone	(310) 322-3331
	FAX	(310) 322-3332
Motorola	Phone	(800) 521-6274
	FAX	(602) 244-6609
General Semiconductor	Phone	(516) 847-3000
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850

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## Table 6. Output Capacitors for Adjustable Output Voltage Applications

		N W	V.CU		e Mount		MY.C.	
Output Voltage (V)	Inductance (µH)	AVX TPS Series			ue 594D eries	Kemet T495 Series		
.Io. COM	A.	No.	C Code	No.	C Code	No.	C Code	
1.21 to 2.50	33*	7	C1	6	C2	7	C3	
1.21 to 2.50	47*	5	C1	4	C2	5	C3	
2.5 to 3.75	33*	4	C1	3	C2	4	C3	
2.0 10 0.10	47*	3	C1	2	C2	3	C3	
	22	4	C1	3	C2	4	C3	
3.75 to 5	33	3	C1	2	C2	3	C3	
WW.	47	2	C1	2	C2	2	C3	
	22	3	C2	3	C3	3	C4	
5 to 6.25	33	2	C2	2	C3	2	C4	
5 10 0.25	47	2	C2	2	C3	2	C4	
WWW.10	68	1	C2	1.00	C3	1	C4	
6.25 to 7.5	22	3	C2	1	C4	3	C4	
	33	2	C2	1.10	C3	2	C4	
	47	1	C3 🔨	1	C4	1	C6	
	68	1.1	C2	1	C3	1	C4	
	33	2	C5	1	C6	2	C8	
7.5 to 10	47	11.1	C5	1	C6	2	C8	
	68	1	C5	1	C6	1	C8	
	100	LCG	C4	11	C5		C8	
	33	- CDN	C5	1	C6	2	C8	
10 to 12.5	47	1	C5	1	C6	2	C8	
10 10 12.5	68	1.1	C5	1 📢	C6	1	C8	
	100	100	C5	1	C6	1.1	C8	
	33	1	C6	1	C8	100	C8	
12.5 to 15	47	1001	C6	1	C8	001	C8	
12.5 to 15	68	11	C6	1	C8	1001	C8	
	100	1	C6	1	C8	1	C8	
	33	1.10	C8	1	C10	2	C10	
15 to 20	47	1,00	C8	1	C9	2	C10	
15 10 20	68	1	C8	1	C9	2	C10	
	100	1.	C8	1	C9	1	C10	
	33	2	C9	2	C11	2	C11	
20 to 30	47	1	C10	1	C12	1	C11	
20 10 30	68	1	C9	1	C12	1	C11	
	100	1	C9	01	C12	1	C11	
	10		W.100	4	C13	8	C12	
	15		1001	3	C13	5	C12	
30 to 37	22	No Value	es Available	2	C13	4	C12	
00 10 01	33		WW.Ivv		C13	3	C12	
	47		W.100	1	C13	2	C12	
	68		NN	1	C13	2	C12	

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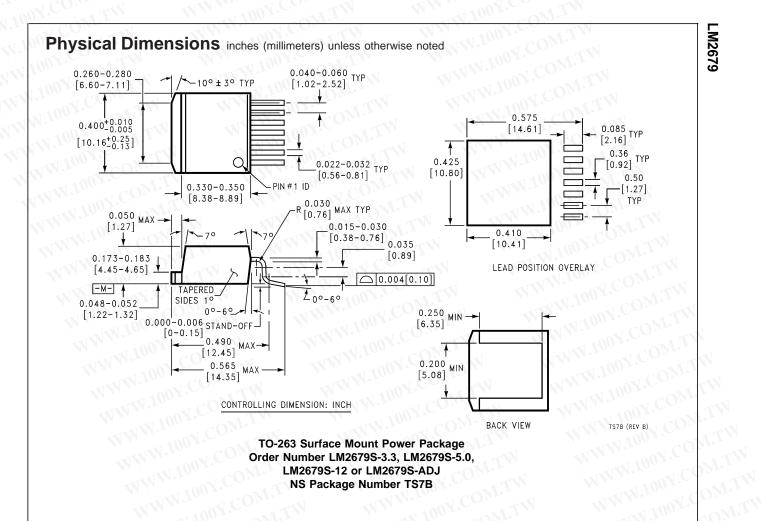
## Table 6. Output Capacitors for Adjustable Output Voltage Applications (continued)

	Inductance (µH)	Through Hole								
Output Voltage (V)		Sanyo OS-CON SA Series		Sanyo MV-GX Series		Nichicor	PL Series	Panasonic HFQ Series		
COM.1		No.	C Code	No.	C Code	No.	C Code	No.	C Code	
1.21 to 2.50	33*	2	C3	5	C1	5	C3	3	С	
1.21 10 2.50	47*	2	C2	4	C1	3	C3	2	C5	
2.5 to 3.75	33*	1	C3	3	C1 🔨	3	C1	2	C5	
2.5 10 3.75	47*	11.	C2	2	C1	2	C3	1	C5	
T.M.	22	1	C3	3	C1	3	C1	2	C5	
3.75 to 5	33	1	C2	2	C1	2	C1	1	C5	
	47	1	C2	2	C1	1	C3	1	C5	
1001.00	22	1	C5	2	C6	2	C3	2	C5	
E to C DE	33	1	C4	1	C6	2	C1	1	C5	
5 to 6.25	47	1	C4	. 1	C6	1	C3	1	C5	
	68	1	C4	$\sqrt{10^{N}}$	C6	1	C1	CY	C5	
-W.1001.	22	1	C5	1.0	C6	2	C1	10	C5	
6.25 to 7.5	33	1	C4	0 1	C6	1	C3	1	C5	
	47	1	C4	1.C	C6	1	C1	01	C5	
	68	1	C4	1	C2	1	C1	-1 C	C5	
WWW.100	33	1	C7	101	C6	1	C14	1	C5	
	47	1	C7	- 10Y	C6	1	C14	101	C5	
7.5 to 10	68	1	C7	1	C2	1	C14	101	C2	
	100	1	C7	11	C2	1	C14	1	C2	
N W	33	1	C7	110	C6	1	C14	1.1	C5	
10 1 10 5	47	1	C7 🔨	1	C2	1	C14	110	C5	
10 to 12.5	68	1	C7	1	C2	1	C9 🔨	1	C2	
	100	011	C7	1	C2	1	C9	11	C2	
N.A.	33	1.1	C9	1	C10	1	C15	1.	C2	
	47	1	C9	1	C10	11	C15	1	C2	
12.5 to 15	68	CO1	C9	1	C10		C15	1	C2	
	100	_1\	C9	1	C10		C15	1	C2	
1	33	1	C10	1	C7	1	C15	1	C2	
45 44 00	47	1.9	C10	1	C7	1	C15	1	C2	
15 to 20	68	100	C10	1	C7	V.00	C15	1	C2	
	100	1	C10	1	C7	10	C15	1	C2	
	33	1004.0	WI.In	1	C7	001	C16	1	C2	
00 ( 00	47	No	Values	1	C7		C16	1	C2	
20 to 30	68	Ava	ailable	1	C7	1	C16	1	C2	
	100		. MO	1	C7	1.101	C16	1	C2	
	10	100	Y.U	1	C12	1	C20	1	C10	
	15		V.CONT	1	C11	1	C20	1	C11	
00 ( ) 07	22	No	Values	1	C11	1	C20	1	C10	
30 to 37	33	Ava	ailable	1	C11	1	C20	1	C10	
	47		F	1	C11	1	C20	1	C10	
	68		F	1	C11	1	C20	1	C10	

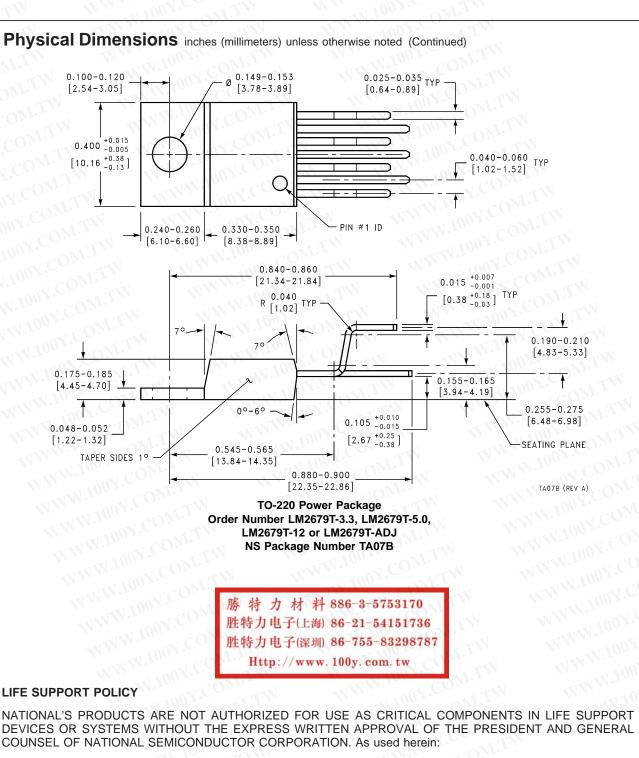
\* Set to a higher value for a practical design solution. See Applications Hints section

No. represents the number of identical capacitor types to be connected in parallel

C Code indicates the Capacitor Reference number in Table 2 for identifying the specific component from the manufacturer.







- Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.
- A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

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