

MCP601/2/3/4

2.7V to 5.5V Single-Supply CMOS Op Amps

Features

Single-Supply: 2.7V to 5.5V

· Rail-to-Rail Output

· Input Range Includes Ground

Gain Bandwidth Product: 2.8 MHz (typ.)

· Unity-Gain Stable

Low Quiescent Current: 230 μA/amplifier (typ.)

Chip Select (CS): MCP603 only

· Temperature Ranges:

Industrial: -40°C to +85°C
 Extended: -40°C to +125°C
 Available in Single, Dual and Quad

Typical Applications

- · Portable Equipment
- · A/D Converter Driver
- · Photo Diode Pre-amp
- Analog Filters
- Data Acquisition
- · Notebooks and PDAs
- Sensor Interface

Available Tools

- · SPICE Macro Models at www.microchip.com
- FilterLab[®] Software at www.microchip.com

Description

The Microchip Technology Inc. MCP601/2/3/4 family of low-power operational amplifiers (op amps) are offered in single (MCP601), single with Chip Select ($\overline{\text{CS}}$) (MCP603), dual (MCP602) and quad (MCP604) configurations. These op amps utilize an advanced CMOS technology that provides low bias current, high-speed operation, high open-loop gain and rail-to-rail output swing. This product offering operates with a single supply voltage that can be as low as 2.7V, while drawing 230 μA (typ.) of quiescent current per amplifier. In addition, the common mode input voltage range goes 0.3V below ground, making these amplifiers ideal for single-supply operation.

These devices are appropriate for low-power, batteryoperated circuits due to the low quiescent current, for A/D convert driver amplifiers because of their wide bandwidth or for anti-aliasing filters by virtue of their low input bias current.

The MCP601, MCP602 and MCP603 are available in standard 8-lead PDIP, SOIC and TSSOP packages. The MCP601 and MCP601R are also available in a standard 5-lead SOT-23 package, while the MCP603 is available in a standard 6-lead SOT-23 package. The MCP604 is offered in standard 14-lead PDIP, SOIC and TSSOP packages.

The MCP601/2/3/4 family is available in the Industrial and Extended temperature ranges and has a power supply range of 2.7V to 5.5V.

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Package Types

MCP601 PDIP, SOIC, TSSOP	MCP602 PDIP, SOIC, TSSOP	MCP603 PDIP, SOIC, TSSOP	MCP604 PDIP, SOIC, TSSOP
NC 1 8 NC	V _{OUTA} 1 8 V _{DD}	NC 1 8 CS	V _{OUTA} 1 14 V _{OUTD}
V_{IN} – $\boxed{2}$ $\boxed{7}$ V_{D}	D V _{INA} - 2 7 V _{OUTB}	V_{IN} – 2 $\overline{7}$ V_{DD}	V _{INA} - 2 13 V _{IND} -
V _{IN} + 3 6 V _C	UT V _{INA} + 3 6 V _{INB} -	V _{IN} + 3 6 V _{OUT}	
V _{SS} 4 5 NO	V_{SS} 4 5 V_{INB} +	V _{SS} 4 5 NC	V_{DD} 4 11 V_{SS}
MCP601 SOT23-5	MCP601R SOT23-5	MCP603 SOT23-6	V _{INB} + 5 10 V _{INC} + 9 V _{INC} -
V _{OUT} 1 5 V _D	D V _{OUT} 1 5 V _{SS}	V _{OUT} 1 6 V _{DD}	V _{OUTB} 7 8 V _{OUTC}
V _{SS} 2	V _{DD} 2	V _{SS} 2 5 CS	
V _{IN} + 3 4 V _{II}	V _{IN} + 3 4 V _{IN} -	V _{IN} + 3 4 V _{IN} -	COM.TW

1.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

V _{DD} - V _{SS}	7.0V
All inputs and outputs	
Difference Input voltage	V _{DD} - V _{SS}
Output Short Circuit Current	continuous
Current at Input Pin	±2 mA
Current at Output and Supply Pins	±30 mA
Storage temperature	65°C to +150°C
Junction temperature	+150°C
ESD protection on all pins (HBM; N	MM) ≥ 3 kV; 200V

† Notice: Stresses above those listed under "Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

PIN FUNCTION TABLE

Name	Function
V_{IN} +, V_{INA} +, V_{INB} +, V_{INC} +, V_{IND} +	Non-inverting Inputs
V _{IN} -, V _{INA} -, V _{INB} -, V _{INC} -, V _{IND} -	Inverting Inputs
V_{DD}	Positive Power Supply
V_{SS}	Negative Power Supply
V _{OUT} , V _{OUTA} , V _{OUTB} , V _{OUTC} , V _{OUTD}	Outputs
CS	Chip Select
NC CO	No Internal Connection

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DC CHARACTERISTICS

Parameters	Sym	Min	Тур	Max	Units	Conditions
Input Offset			00	COMP	.41	CONT.
Input Offset Voltage	Vos	-2	±0.7	+2	mV	M. 1001.
Industrial Temperature	V _{os}	-3	±1	+3	mV	$T_A = -40^{\circ}\text{C to } +85^{\circ}\text{C (Note 1)}$
Extended Temperature	Vos	-4.5	±1	+4.5	mV	$T_A = -40^{\circ}\text{C to } +125^{\circ}\text{C (Note 1)}$
Input Offset Temperature Drift	$\Delta V_{OS}/\Delta T_{A}$	AT MA	±2.5	1.0	μV/°C	T _A = -40°C to +125°C
Power Supply Rejection	PSRR	80	88	~ €U	dB	V _{DD} = 2.7V to 5.5V
Input Current and Impedance	I.A.		-TV 11	10 ×.	MI	N.Inc. COM.
Input Bias Current	I_{B}	-4/	1	WATE.	pA	W W 1001.
Industrial Temperature	l _B	_	20	60	pA	T _A = +85°C (Note 1)
Extended Temperature	Ι _Β		450	5000	pA	T _A = +125°C (Note 1)
Input Offset Current	Ios	_	±1	<u>, m</u> Y.	pA	TW WW. 1007.0
Common Mode Input Impedance	Z _{CM}	_	10 ¹³ 6	11.70	$\Omega pF $	W. W. W. C.
Differential Input Impedance	Z _{DIFF}	_	10 ¹³ 3	W. 700	$\Omega pF $	VI.7
Common Mode		N	AN A	100	A.C.	-11 TW WW. 1007.2
Common Mode Input Range	V_{CMR}	$V_{SS} - 0.3$	-31	V _{DD} – 1.2	V	DIA WINN CON
Common Mode Rejection Ratio	CMRR	75	90	11.11	dB	$V_{DD} = 5.0V$, $V_{CM} = -0.3V$ to 3.8V
Open-loop Gain	I.Co	TW	V	NA T	007.	-11 TW WY 100
DC Open-loop Gain (large signal)	A _{OL}	100	115	WITH.	dB	R_L = 25 k Ω to $V_{DD}/2$, V_{OUT} = 100 mV to V_{DD} – 100 mV
WWW.I	A _{OL}	95	110	MAN	dB	R_L = 5 k Ω to $V_{DD}/2$, V_{OUT} = 100 mV to V_{DD} – 100 mV
Output	ON.	WILL		Mar	- T 10	DY. SMITH WY
Maximum Output Voltage Swing	V_{OL}, V_{OH}	V _{SS} + 15	<u> </u>	V _{DD} – 20	mV	$R_L = 25 \text{ k}\Omega$ to $V_{DD}/2$, Output overdrive = 0.5\
W.	V_{OL}, V_{OH}	V _{SS} + 45	_	V _{DD} – 60	mV	$R_L = 5 \text{ k}\Omega \text{ to } V_{DD}/2$, Output overdrive = 0.5V
Linear Output Voltage Swing	V _{OUT}	V _{SS} + 100	11-	V _{DD} – 100	mV	R_L = 25 k Ω to $V_{DD}/2$, $A_{OL} \ge 100$ dB
	V _{OUT}	V _{SS} + 100	1	V _{DD} – 100	mV	$R_L = 5 \text{ k}\Omega \text{ to } V_{DD}/2, A_{OL} \ge 95 \text{ dB}$
Output Short Circuit Current	I _{SC}	$M_{\overline{C}_{-}}$	±22	_	mA	V _{DD} = 5.5V
TVI	I _{SC}	V.CZ	±12	_ <	mA	V _{DD} = 2.7V
Power Supply	M.Inc	-1 CON	1.0-	í	-TIN	W. P. COM.
Supply Voltage	V_{DD}	2.7	$V_{\overline{A},\Lambda}$	5.5	V	MI JON TOWN TO
Quiescent Current per Amplifier	I _Q	A CO	230	325	μA	I _O = 0

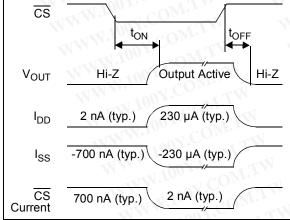
Note 1: These specifications are not tested in either the SOT-23 or TSSOP packages with date codes older than YYWW = 0408. In these cases, the minimum and maximum values are by design and characterization only.

AC CHARACTERISTICS

Parameters	Sym	Min	Тур	Max	Units	Conditions
Frequency Response	- 11 1	W	W.	- 1	001.	MITW
Gain Bandwidth Product	GBWP	CV I	2.8	$M_{\overline{A}_A}$.	MHz	O. TW
Phase Margin	PM	<u>-</u>	50	N.		G = +1 V/V
Step Response	1.00	1.7.4		11	N.100 x	COMIT
Slew Rate	SR	WFIL	2.3	M ZA	V/µs	G = +1 V/V
Settling Time (0.01%)	t _{settle}		4.5	4 0 1	μs	G = +1 V/V, 3.8V step
Noise	100 -	DIVI	~ T	-11	MW.In	COM
Input Noise Voltage	E _{ni}	TIVE	7	7//	μV _{P-P}	f = 0.1 Hz to 10 Hz
Input Noise Voltage Density	e _{ni}		29		nV/√Hz	f = 1 kHz
COM.	e _{ni}	$C_{\overline{O}_{M_I}}$	21	_	nV/√Hz	f = 10 kHz
Input Noise Current Density	J _{ni}	More	0.6	_	fA/√Hz	f = 1 kHz

MCP603 CHIP SELECT CHARACTERISTICS

Electrical Specifications: Unless otherw $V_{OUT} \approx V_{DD}/2$, R_L = 100 kΩ to $V_{DD}/2$ and			°C, V _{DD}	= +2.7V to	+5.5V,	$V_{SS} = GND, V_{CM} = V_{DD}/2,$
Parameters	Sym	Min	Тур	Max	Units	Conditions
DC Characteristics	MAN	on Y.C.	- 17	N	W	M. 100XIC TA
CS Logic Threshold, Low	V _{IL}	V _{SS}	$O_{\overline{M}^{p_{1}}}$	0.2 V _{DD}	V	INW. OV.COM
CS Input Current, Low	I _{CSL}	-1.0	Man.	_	μA	<u>CS</u> = 0.2V _{DD}
CS Logic Threshold, High	V _{IH}	0.8 V _{DD}		V_{DD}	V	M. 100x.
CS Input Current, High	I _{CSH}	-001	0.7	2.0	μA	CS = V _{DD}
Shutdown V _{SS} current	I _{Q_SHDN}	-2.0	-0.7	Mr. =	μA	CS = V _{DD}
Amplifier Output Leakage in Shutdown	I _{O_SHDN}	100 tu	1	W .,	nA	W.100 COM.
CS Threshold Hysteresis	HYST	- 10	0.3	- TT	V	Internal switch
Timing	11	MMir	anv.C	Obs	W	MMM. OUT.CO.
CS Low to Amplifier Output Turn-on Time	t _{ON}	WW.	3.1	10	μs	<u>CS</u> ≤ 0.2V _{DD} , G = +1 V/V
CS High to Amplifier Output High-Z Time	t _{OFF}	W W	100	$C_{\Omega_{D_s}}$	ns	$\overline{\text{CS}} \ge 0.8 \text{V}_{\text{DD}}, \text{ G} = +1 \text{ V/V}, \text{ No load.}$



MCP603 Chip Select (CS) FIGURE 1-1: Timing Diagram.

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TEMPERATURE CHARACTERISTICS

Electrical Specifications: Unless	otherwise	e indicat	ed, V _{DD}	= +2.7V	to +5.5V	and V _{SS} = GND.
Parameters	Sym	Min	Тур	Max	Units	Conditions
Temperature Ranges	47 CO	1.2	1	-3111	M. Lan	A COMP.
Specified Temperature Range	T _A	-40		+85	°C	Industrial temperature parts
	T _A	-40	N	+125	°C	Extended temperature part
Operating Temperature Range	TA	-40	W-	+125	°C	Note
Storage Temperature Range	T _A	-65	TIN	+150	°C	CON TW
Thermal Package Resistances	1.100	COM	. 1		VIVI	Too COMP.
Thermal Resistance, 5L-SOT23	θ_{JA}	- TOT	256	_	°C/W	N.100 F. COM. I.
Thermal Resistance, 6L-SOT23	θ_{JA}	V.C	230		°C/W	1100Y. OM.TW
Thermal Resistance, 8L-PDIP	θ_{JA}	W.C.C	85	N —	°C/W	TW. CO. TW
Thermal Resistance, 8L-SOIC	θ_{JA}	√ √ C	163	<u> </u>	°C/W	MM. TO COM
Thermal Resistance, 8L-TSSOP	θ_{JA}	00 -	124	- T	°C/W	M.Ing. COM.
Thermal Resistance, 14L-PDIP	θ_{JA}	1001.	70	$\overline{L}\overline{A_A}$	°C/W	1001. COM.
Thermal Resistance, 14L-SOIC	θ_{JA}	11107	120		°C/W	MW. 100X:00M
Thermal Resistance, 14L-TSSOP	θ_{JA}	V. <u>F</u>	100	 (\)	°C/W	MAN. O. CO.

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Note: The Industrial temperature parts operate over this extended range, but with reduced performance. The Extended temperature specs do not apply to Industrial temperature parts. In any case, the internal Junction temperature (T_J) must not exceed the absolute maximum specification of 150°C. WWW.100Y.COM.

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2.0 TYPICAL PERFORMANCE CURVES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = $V_{DD}/2$, R_L = 100 k Ω to $V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and C_L = 50 pF.

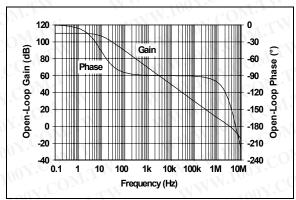


FIGURE 2-1: Open-Loop Gain, Phase vs. Frequency.

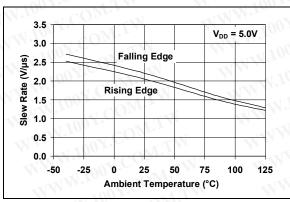


FIGURE 2-2: Slew Rate vs. Temperature.

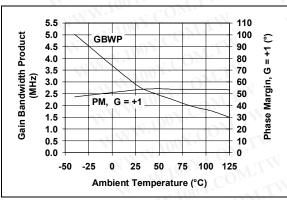


FIGURE 2-3: Gain Bandwidth Product, Phase Margin vs. Temperature.

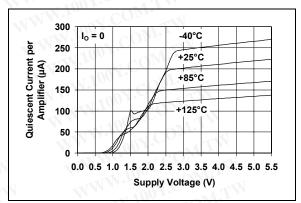


FIGURE 2-4: Quiescent Current vs. Supply Voltage.

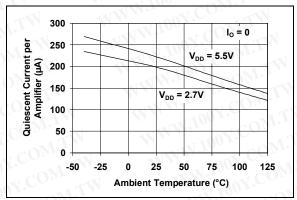


FIGURE 2-5: Quiescent Current vs. Temperature.

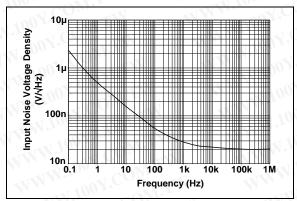


FIGURE 2-6: Input Noise Voltage Density vs. Frequency.

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = $V_{DD}/2$, R_L = 100 k Ω to $V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and C_L = 50 pF.

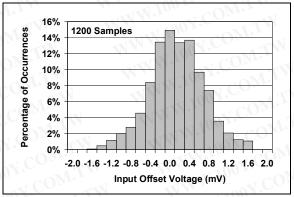


FIGURE 2-7: Input Offset Voltage.

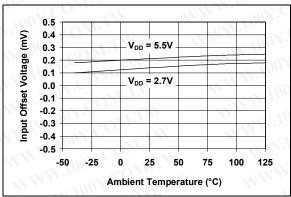


FIGURE 2-8: Input Offset Voltage vs. Temperature.

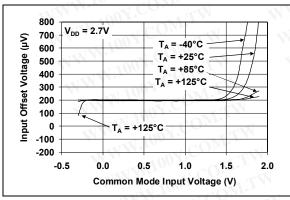


FIGURE 2-9: Input Offset Voltage vs. Common Mode Input Voltage with $V_{DD} = 2.7V$.

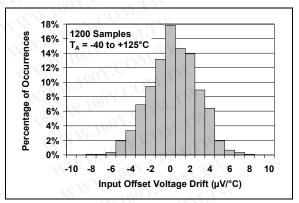


FIGURE 2-10: Input Offset Voltage Drift.

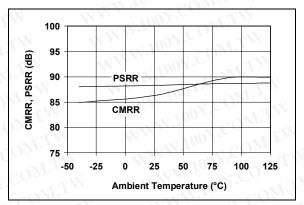


FIGURE 2-11: CMRR, PSRR vs. Temperature.

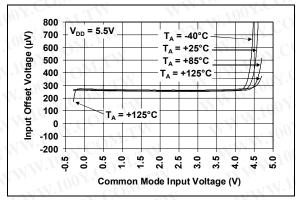


FIGURE 2-12: Input Offset Voltage vs. Common Mode Input Voltage with $V_{DD} = 5.5V$.

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = $V_{DD}/2$, R_L = 100 k Ω to $V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and C_L = 50 pF.

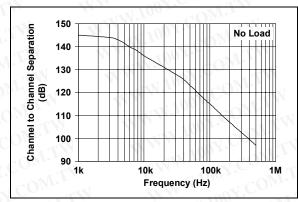


FIGURE 2-13: Channel-to-Channel Separation vs. Frequency.

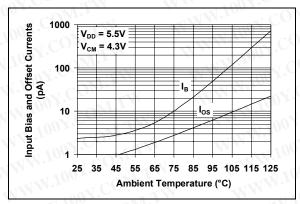


FIGURE 2-14: Input Bias Current, Input Offset Current vs. Ambient Temperature.

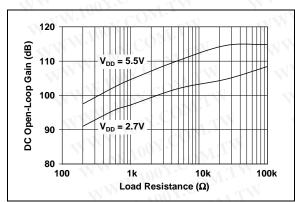


FIGURE 2-15: DC Open-Loop Gain vs. Load Resistance.

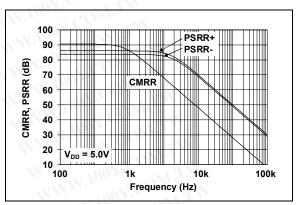


FIGURE 2-16: CMRR, PSRR vs. Frequency.

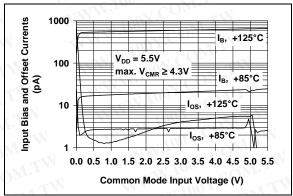


FIGURE 2-17: Input Bias Current, Input Offset Current vs. Common Mode Input Voltage.

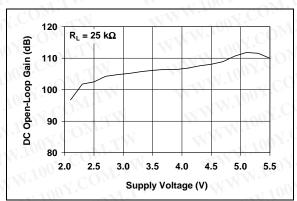


FIGURE 2-18: DC Open-Loop Gain vs. Supply Voltage.

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = V_{DD}/2, R_L = 100 k Ω to V_{DD}/2, V_{OUT} \approx V_{DD}/2 and C_L = 50 pF.

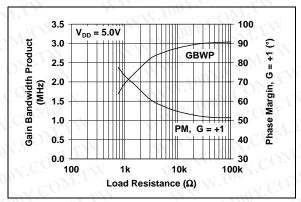


FIGURE 2-19: Gain Bandwidth Product, Phase Margin vs. Load Resistance.

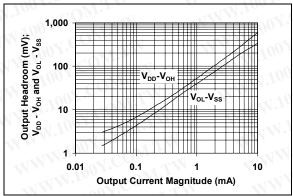


FIGURE 2-20: Output Voltage Headroom vs. Output Current.

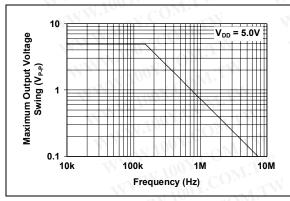


FIGURE 2-21: Maximum Output Voltage Swing vs. Frequency.

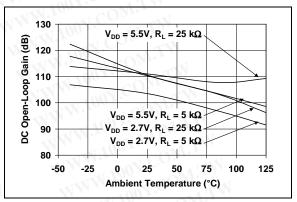


FIGURE 2-22: DC Open-Loop Gain vs. Temperature.

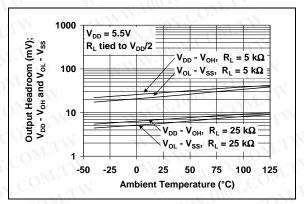


FIGURE 2-23: Output Voltage Headroom vs. Temperature.

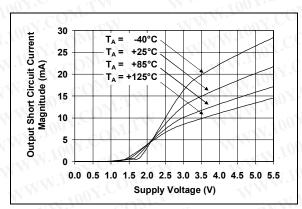


FIGURE 2-24: Output Short-Circuit Current vs. Supply Voltage.

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Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = $V_{DD}/2$, R_L = 100 k Ω to $V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and C_L = 50 pF.

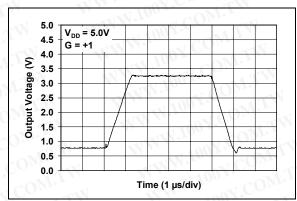


FIGURE 2-25: Large Signal Non-Inverting Pulse Response.

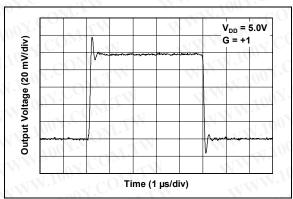


FIGURE 2-26: Small Signal Non-Inverting Pulse Response.

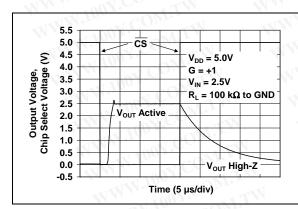


FIGURE 2-27: Chip Select Timing (MCP603).

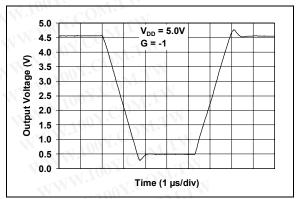


FIGURE 2-28: Large Signal Inverting Pulse Response.

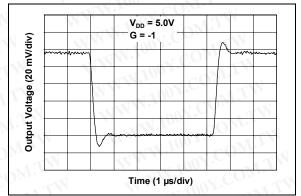


FIGURE 2-29: Small Signal Inverting Pulse Response.

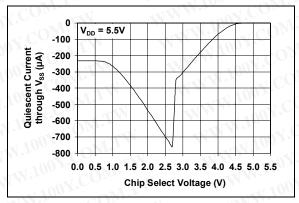


FIGURE 2-30: Quiescent Current Through V_{SS} vs. Chip Select Voltage (MCP603).

MCP601/2/3/4

Note: Unless otherwise indicated, T_A = +25°C, V_{DD} = +2.7V to +5.5V, V_{SS} = GND, V_{CM} = $V_{DD}/2$, R_L = 100 k Ω to $V_{DD}/2$, $V_{OUT} \approx V_{DD}/2$ and C_L = 50 pF.

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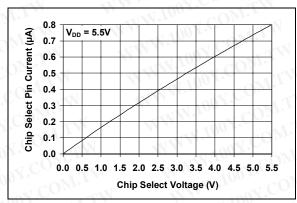
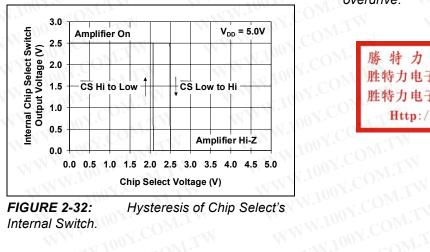


FIGURE 2-31: Chip Select Pin Input Current vs. Chip Select Voltage.



WWW.100Y.COM.TW FIGURE 2-32: Hysteresis of Chip Select's Internal Switch. WWW.100Y.COM.TW

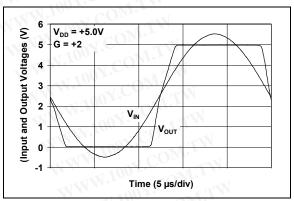


FIGURE 2-33: The MCP601/2/3/4 family of op amps shows no phase reversal under input overdrive.

3.0 APPLICATIONS INFORMATION

The MCP601/2/3/4 family of op amps are fabricated on Microchip's state-of-the-art CMOS process. They are unity-gain stable and suitable for a wide range of general purpose applications.

3.1 Input

The MCP601/2/3/4 amplifier family is designed to not exhibit phase reversal when the input pins exceed the supply rails. Figure 2-33 shows an input voltage that exceeds both supplies with no resulting phase inversion.

The Common Mode Input Voltage Range (V_{CMR}) includes ground in single-supply systems (V_{SS}), but does not include V_{DD} . This means that the amplifier input behaves linearly as long as the Common Mode Input Voltage (V_{CM}) is kept within the specified V_{CMR} limits ($V_{SS} - 0.3V$ to $V_{DD} - 1.2V$ at +25°C).

Input voltages that exceed the input voltage range ($V_{SS} - 0.3V$ to $V_{DD} - 1.2V$ at $+25^{\circ}$ C) can cause excessive current to flow into or out of the input pins. Current beyond ± 2 mA may cause reliability problems. Applications that exceed this rating must externally limit the input current with a resistor (R_{IN}), as shown in Figure 3-1.

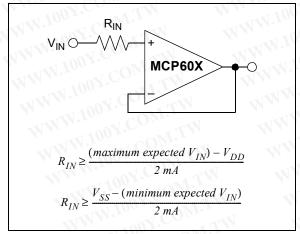


FIGURE 3-1: R_{IN} limits the current flow into an input pin.

3.2 Rail-to-Rail Output

There are two specifications that describe the output swing capability of the MCP601/2/3/4 family of op amps. The first specification (Maximum Output Voltage Swing) defines the absolute maximum swing that can be achieved under the specified load conditions. For instance, the output voltage swings to within 15 mV of the negative rail with a 25 k Ω load to V_{DD}/2. Figure 2-33 shows how the output voltage is limited when the input goes beyond the linear region of operation.

The second specification that describes the output swing capability of these amplifiers is the Linear Output Voltage Swing. This specification defines the maximum output swing that can be achieved while the amplifier is still operating in its linear region. To verify linear operation in this range, the large signal (DC Open-Loop Gain (A_{OL})) is measured at points 100 mV inside the supply rails. The measurement must exceed the specified gains in the specification table.

3.3 MCP603 Chip Select (CS)

The MCP603 is a single amplifier with Chip Select ($\overline{\text{CS}}$). When $\overline{\text{CS}}$ is pulled high, the supply current drops to -0.7 μA (typ.), which is pulled through the $\overline{\text{CS}}$ pin to V_{SS}. When this happens, the amplifier output is put into a high-impedance state. Pulling $\overline{\text{CS}}$ low enables the amplifier and, if the $\overline{\text{CS}}$ pin is left floating, the amplifier may not operate properly. Figure 1-1 is the Chip Select timing diagram and shows the output voltage, supply currents and $\overline{\text{CS}}$ current in response to a $\overline{\text{CS}}$ pulse. Figure 2-27 shows the measured output voltage response to a $\overline{\text{CS}}$ pulse.

3.4 Capacitive Loads

Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases and the closed-loop bandwidth is reduced. This produces gain peaking in the frequency response with overshoot and ringing in the step response.

When driving large capacitive loads with these op amps (e.g., > 40 pF when G = +1), a small series resistor at the output ($R_{\rm ISO}$ in Figure 3-2) improves the feedback loop's phase margin (stability) by making the output load resistive at higher frequencies. The bandwidth will be generally lower than the bandwidth with no capacitive load.

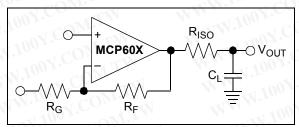


FIGURE 3-2: Output resistor R_{ISO} stabilizes large capacitive loads.

Figure 3-3 gives recommended R_{ISO} values for different capacitive loads and gains. The x-axis is the normalized load capacitance (C_L/G_N) in order to make it easier to interpret the plot for arbitrary gains. G_N is the circuit's noise gain. For non-inverting gains, G_N and the gain are equal. For inverting gains, $G_N = 1 + |Gain| (e.g., -1 V/V gives <math>G_N = +2 V/V$).

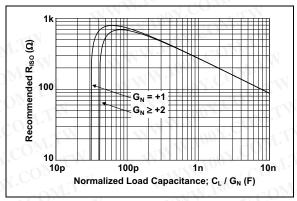


FIGURE 3-3: Recommended R_{ISO} values for capacitive loads.

Once you've selected $R_{\rm ISO}$ for your circuit, double-check the resulting frequency response peaking and step response overshoot in your circuit. Evaluation on the bench and simulations with the MCP601/2/3/4 SPICE macro model are very helpful. Modify $R_{\rm ISO}$'s value until the response is reasonable.

3.5 Supply Bypass

With this family of op amps, the power supply pin (V_{DD} for single-supply) should have a local bypass capacitor (i.e., 0.01 μ F to 0.1 μ F) within 2 mm for good high-frequency performance. It also needs a bulk capacitor (i.e., 1 μ F or larger) within 100 mm to provide large, slow currents. This bulk capacitor can be shared with other parts.

3.6 PCB Surface Leakage

In applications where low input bias current is critical, printed circuit board (PCB) surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low humidity conditions, a typical resistance between nearby traces is $10^{12}\Omega$. A 5V difference would cause 5 pA of current to flow. This is greater than the MCP601/2/3/4 family's bias current at +25°C (1 pA, typ.).

The easiest way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 3-4.

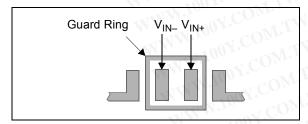


FIGURE 3-4: Example Guard Ring layout.

- Connect the guard ring to the inverting input pin (V_{IN}-) for non-inverting gain amplifiers, including unity-gain buffers. This biases the guard ring to the common mode input voltage.
- Connect the guard ring to the non-inverting input pin (V_{IN}+) for inverting gain amplifiers and transimpedance amplifiers (converts current to voltage, such as photo detectors). This biases the guard ring to the same reference voltage as the op amp (e.g., V_{DD}/2 or ground).

3.7 Typical Applications

3.7.1 ANALOG FILTERS

Figure 3-5 and Figure 3-6 show low-pass, second-order, Butterworth filters with a cutoff frequency of 10 Hz. The filter in Figure 3-5 has a non-inverting gain of +1 V/V, and the filter in Figure 3-6 has an inverting gain of -1 V/V.

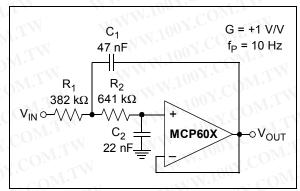


FIGURE 3-5: Second-Order, Low-Pass Sallen-Key Filter.

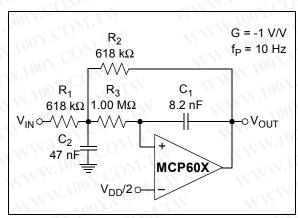


FIGURE 3-6: Second-Order, Low-Pass Multiple-Feedback Filter.

The MCP601/2/3/4 family of op amps have low input bias current, which allows the designer to select larger resistor values and smaller capacitor values for these filters. This helps produce a compact PCB layout. These filters, and others, can be designed using Microchip's FilterLab® software.

3.7.2 INSTRUMENTATION AMPLIFIER CIRCUITS

Instrumentation amplifiers have a differential input that subtracts one input voltage from another and rejects common mode signals. These amplifiers also provide a single-ended output voltage.

The three-op amp instrumentation amplifier is illustrated in Figure 3-7. One advantage of this approach is unitygain operation, while one disadvantage is that the common mode input range is reduced as R_2/R_G gets larger.

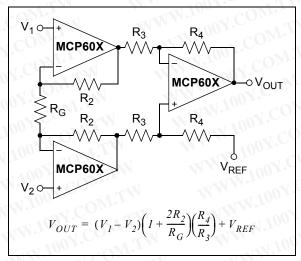


FIGURE 3-7: Three-Op Amp Instrumentation Amplifier.

The two-op amp instrumentation amplifier is shown in Figure 3-8. While its power consumption is lower than the three-op amp version, its main drawbacks are that the common mode range is reduced with higher gains and it must be configured in gains of two or higher.

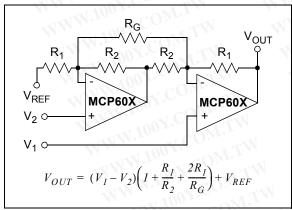


FIGURE 3-8: Two-Op Amp Instrumentation Amplifier.

Both instrumentation amplifiers should use a bulk bypass capacitor of at least 1 μ F. The CMRR of these amplifiers will be set by both the op amp CMRR and resistor matching.

3.7.3 PHOTO DETECTION

The MCP601/2/3/4 op amps can be used to easily convert the signal from a sensor that produces an output current (such as a photo diode) into a voltage (a transimpedance amplifier). This is implemented with a single resistor (R_2) in the feedback loop of the amplifiers shown in Figure 3-9 and Figure 3-10. The optional capacitor (C_2) sometimes provides stability for these circuits.

A photodiode configured in the Photovoltaic mode has zero voltage potential placed across it (Figure 3-9). In this mode, the light sensitivity and linearity is maximized, making it best suited for precision applications. The key amplifier specifications for this application are: low input bias current, low noise, common mode input voltage range (including ground) and rail-to-rail output.

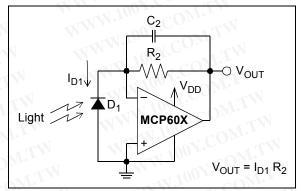


FIGURE 3-9: Photovoltaic Mode Detector.

In contrast, a photodiode that is configured in the Photoconductive mode has a reverse bias voltage across the photo-sensing element (Figure 3-10). This decreases the diode capacitance, which facilitates high-speed operation (e.g., high-speed digital communications). The design trade-off is increased diode leakage current and linearity errors. The op amp needs to have a wide Gain Bandwidth Product (GBWP).

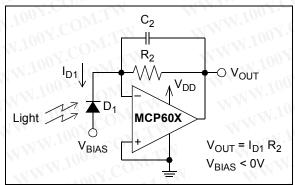


FIGURE 3-10: Photoconductive Mode Detector.

4.0 **DESIGN TOOLS**

Microchip provides the basic design tools needed for the MCP601/2/3/4 family of op amps.

4.1 SPICE Macro Model

The latest SPICE macro model of the MCP601/2/3/4 op amps is available on Microchip's web site at www.microchip.com. This model is intended as an initial design tool that works well in the op amp's linear region of operation at room temperature. See the SPICE model firmware for information on its capabilities.

Bench testing is a very important part of any design and cannot be replaced with simulations. Also, simulation results using this macro model need to be validated by comparing them to the data sheet specs and plots.

FilterLab® 2.0 4.2

FilterLab® 2.0 is an innovative software tool that simplifies analog active-filter (using op amps) design. Available at no cost from Microchip's web site at www.microchip.com, the FilterLab active-filter software design tool provides full schematic diagrams of the filter circuit with component values. It also outputs the filter circuit in SPICE format, which can be used with the macro model to simulate actual filter performance.

> 特力材料886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787 Http://www. 100y. com. tw

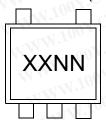
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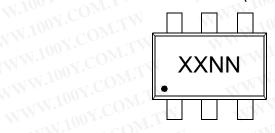
5.0 PACKAGING INFORMATION

5.1 Package Marking Information





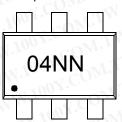
6-Lead SOT-23A (MCP603 Only)







Example:



勝 特 力 材 料 886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787 Http://www.100y.com.tw

Legend: XX...X Customer specific information*

YY Year code (last 2 digits of calendar year)
WW Week code (week of January 1 is week '01')

NNN Alphanumeric traceability code

Note: In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line thus limiting the number of available characters for customer specific information.

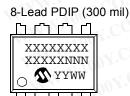
* Standard OTP marking consists of Microchip part number, year code, week code, and traceability code.

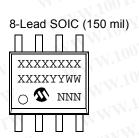
MCP601/2/3/4

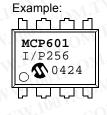
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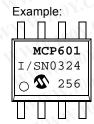
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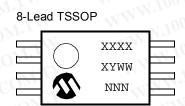
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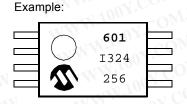


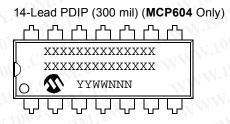


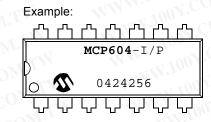


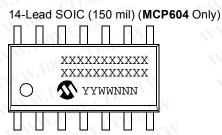


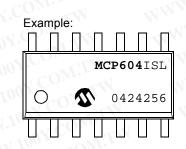


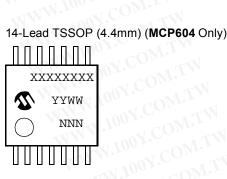






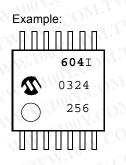




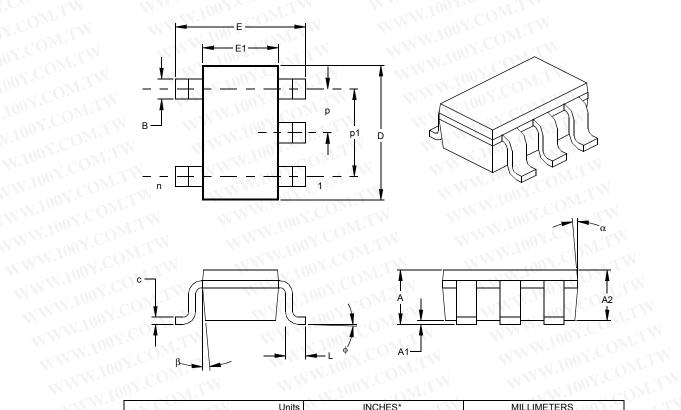


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5-Lead Plastic Small Outline Transistor (OT) (SOT-23)



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N.Co.	Units	10	INCHES*		MI	ILLIMETERS	001.
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	1	5	TIVE		5	100 1.
Pitch	р	TAN W.	.038	OLA P.	K.T	0.95	
Outside lead pitch (basic)	p1	N .	.075	7.17		1.90	x1 100°
Overall Height	Α	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff §	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	ф	0	5	10	0	5	10
_ead Thickness	С	.004	.006	.008	0.09	0.15	0.20
_ead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10
* Controlling Parameter § Significant Characteristic	VIII	- T	TAN V	1.700	$CO_{M'}$	-33	- T

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JEDEC Equivalent: MO-178

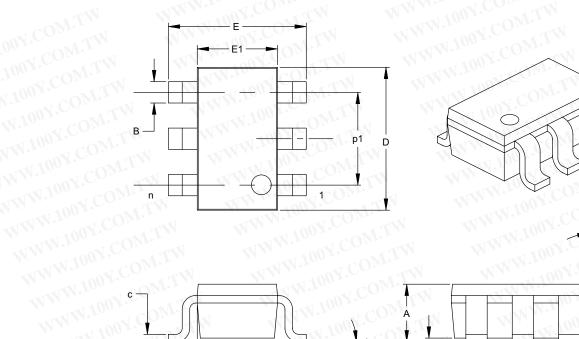
Drawing No. C04-091

材 料 886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787

Http://www.100y.com.tw

^{*} Controlling Parameter § Significant Characteristic

6-Lead Plastic Small Outline Transistor (CH) (SOT-23)



WW.100Y.COM.TW

-1007.	Linite		INCLIECT	MITH		UL IMETEDO.	$t_{00x.z}$
Dimensio	Units	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	IVIIIV	6	IVIAX	IVIIIN	1NOIVI 6	IVIAX
Pitch	р		.038		A	0.95	-1100
Outside lead pitch (basic)	p1	-111	.075	COMP	-11	1.90	W. 7.
Overall Height	A	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	E	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	LXX	.014	.018	.022	0.35	0.45	0.55
Foot Angle	φ	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

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JEITA (formerly EIAJ) equivalent: SC-74A Drawing No. C04-120

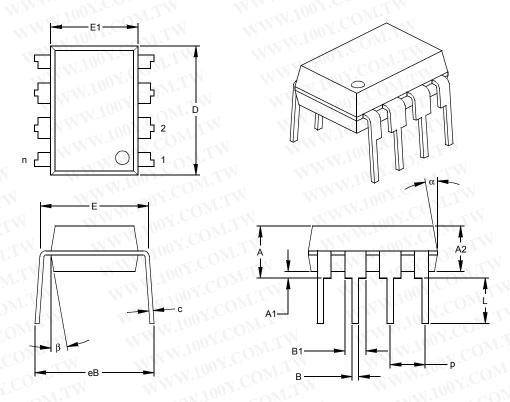
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^{*}Controlling Parameter

8-Lead Plastic Dual In-line (P) - 300 mil (PDIP)



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Y.C. TW	Units	100	INCHES*	LVV	М	ILLIMETERS	01.
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	100	8	1.7.4		8	00 .
Pitch	р	MM.	.100	TXX.		2.54	J.Vac
Top to Seating Plane	Α	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015	00-	JAI.	0.38		I.To.
Shoulder to Shoulder Width	Е	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing §	еВ	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

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.010" (0.254mm) per side. JEDEC Equivalent: MS-001

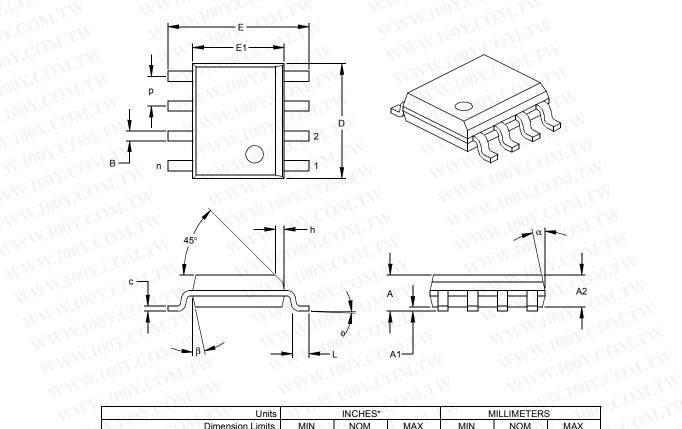
Drawing No. C04-018

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WW.100Y.COM.TW

^{*} Controlling Parameter § Significant Characteristic

8-Lead Plastic Small Outline (SN) - Narrow, 150 mil (SOIC)



WW.100Y.COM.TW

	Units		INCHES*	- 7.17	MI	LLIMETERS	
Dimension	n Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	111.	8	Mo	1.4.	8	-TXV 10
Pitch	р	-1111	.050	CO	TIN	1.27	MAL
Overall Height	Α	.053	.061	.069	1.35	1.55	1.75
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25
Overall Width	O E	.228	.237	.244	5.79	6.02	6.20
Molded Package Width	E1	.146	.154	.157	3.71	3.91	3.99
Overall Length	D	.189	.193	.197	4.80	4.90	5.00
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51
Foot Length	L	.019	.025	.030	0.48	0.62	0.76
Foot Angle	ф	0	4	8	0	4	8
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25
Lead Width	В	.013	.017	.020	0.33	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

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.010" (0.254mm) per side.

JEDEC Equivalent: MS-012

Drawing No. C04-057

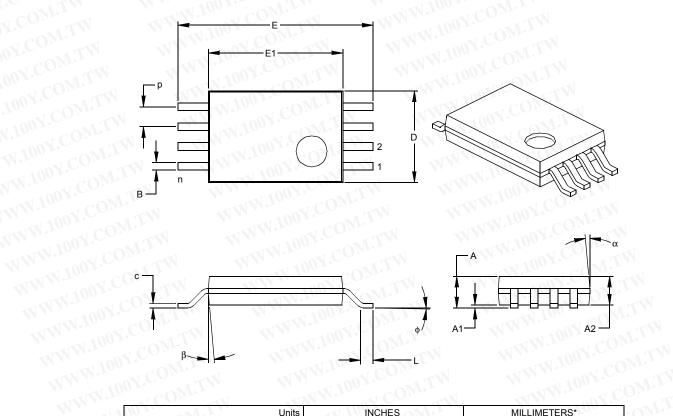
勝 特 力 材 料 886-3-5753170 胜特力电子(上海) 86-21-54151736 胜特力电子(深圳) 86-755-83298787

Http://www.100y.com.tw

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^{*} Controlling Parameter § Significant Characteristic

8-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)



WW.100Y.COM.TW

	Units		INCHES	M. T.	MI	LLIMETERS'	
Dimension	on Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	www.	8	J.Mr.	e T	8	.10
Pitch	р	N NA	.026	TI		0.65	-11007
Overall Height	Α	TAX IN	.10	.043	- 1		1.10
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95
Standoff §	A1	.002	.004	.006	0.05	0.10	0.15
Overall Width	E	.246	.251	.256	6.25	6.38	6.50
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50
Molded Package Length	D	.114	.118	.122	2.90	3.00	3.10
Foot Length	W L	.020	.024	.028	0.50	0.60	0.70
Foot Angle	ф	0	4	8	0	4	8
Lead Thickness	C	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.007	.010	.012	0.19	0.25	0.30
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10
* Controlling Parameter							
§ Significant Characteristic							

Controlling Parameter

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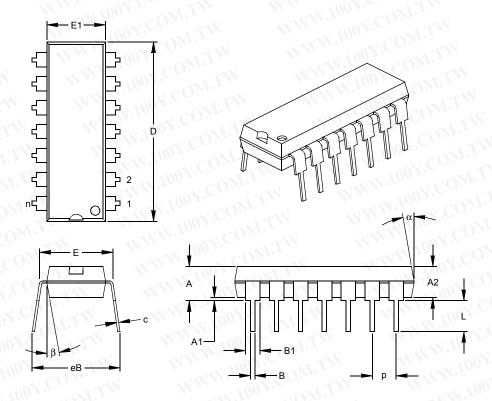
JEDEC Equivalent: MO-153 Drawing No. C04-086

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[§] Significant Characteristic

14-Lead Plastic Dual In-line (P) - 300 mil (PDIP)



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TW	MAG	INCHES*	7.17	MILLIMETERS			
Dimension	n Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	111.	14	100	1.4.	14	-xx1 10
Pitch	р	-1111	.100	1 COR	- N	2.54	Maria
Top to Seating Plane	Α	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015	-13N 1U		0.38		Wire
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.740	.750	.760	18.80	19.05	19.30
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

^{*} Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed

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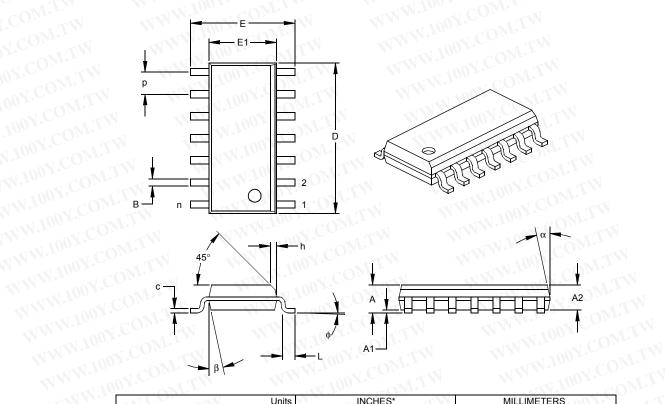
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.010" (0.254mm) per side.

JEDEC Equivalent: MS-001 Drawing No. C04-005

[§] Significant Characteristic

14-Lead Plastic Small Outline (SL) - Narrow, 150 mil (SOIC)



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ON CO.	Units	N	INCHES*		MI	LLIMETERS	OUX.
Dimension	on Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n	MA .	14	TV		14	100 x.
Pitch	р	-TANNO	.050	Olar.	ĸT	1.27	0.5
Overall Height	Α	.053	.061	.069	1.35	1.55	1.75
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25
Overall Width	E	.228	.236	.244	5.79	5.99	6.20
Molded Package Width	E1	.150	.154	.157	3.81	3.90	3.99
Overall Length	D	.337	.342	.347	8.56	8.69	8.81
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51
Foot Length	√\ L	.016	.033	.050	0.41	0.84	1.27
Foot Angle	ф	0	4	8	0	4	8
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25
Lead Width	В	.014	.017	.020	0.36	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15
* Controlling Parameter § Significant Characteristic	Mil	<1	TAN V	1.100	CO_{M+r}	- XXI	

^{*} Controlling Parameter

Notes:

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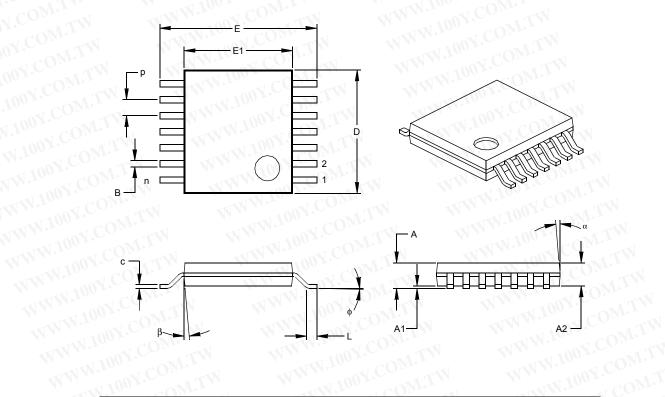
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JEDEC Equivalent: MS-012

Drawing No. C04-065

[§] Significant Characteristic

14-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)



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	Units	Maria.	INCHES	- 1	MI	LLIMETERS	* _ <u> </u>
Dimensio	n Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		14	Ma	7.	14	- xx 1 1 1
Pitch	р	-XIV	.026	1 COm	TV	0.65	M. A.
Overall Height	Α	44.	-XX 100	.043			1.10
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95
Standoff §	A1	.002	.004	.006	0.05	0.10	0.15
Overall Width	(E	.246	.251	.256	6.25	6.38	6.50
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50
Molded Package Length	D	.193	.197	.201	4.90	5.00	5.10
Foot Length	L	.020	.024	.028	0.50	0.60	0.70
Foot Angle	ф	0	4	8	0	4	8
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.007	.010	.012	0.19	0.25	0.30
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10
* Controlling Parameter § Significant Characteristic	M	LM	MV	100	Y.C.	T.IV	1

^{*} Controlling Parameter

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. JEDEC Equivalent: MO-153

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Drawing No. C04-087

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[§] Significant Characteristic

PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office

DARTHO	MAN TO MAN TO THE TOWN TO	Exa	amples:	
	X /XX 	a) b)	MCP601-I/P: MCP601-E/SN:	Single Op Amp, Industrial Temperature, 8LD PDIP package. Single Op Amp,
Device	MCP601 Single Op Amp MCP601T Single Op Amp (Tape and Reel for SOT23, SOIC and TSSOP)	c)	MCP601T-I/OT:	Extended Temperature, 8LD SOIC package. Tape and Reel, Industrial Temperature, Single Op Amp,
	MCP601RT Single Op Amp (Tape and Reel for SOT23-5) MCP602 Dual Op Amp	d)	MCP601T-E/ST:	5-LD SOT23 package. Tape and Reel, Extended Temperature,
	MCP602T Dual Op Amp (Tape and Reel for SOIC and TSSOP) MCP603 Single Op Amp with Chip Select			Single Op Amp, 8LD TSSOP package
	MCP603T Single Op Amp with Chip Select (Tape and Reel for SOT23, SOIC and TSSOP) MCP604 Quad Op Amp MCP604T Quad Op Amp	e)	MCP601RT-E/OT:	Tape and Reel, Extended Temperature, Single Op Amp, Rotated, 5-LD SOT23 package.
Townsorture Bongo	(Tape and Reel for SOIC and TSSOP)	a)	MCP602-I/SN:	Dual Op Amp, Industrial Temperature, 8LD SOIC package.
Temperature Range	= -40°C to +85°C E = -40°C to +125°C	b)	MCP602-E/P:	Dual Op Amp, Extended Temperature, 8LD PDIP package.
Package	OT = Plastic SOT23, 5-lead (MCP601 only) CH = Plastic SOT23, 6-lead (MCP603 only) P = Plastic DIP (300 mil Body), 8, 14-lead SN = Plastic SOIC (150 mil Body), 8-lead SL = Plastic SOIC (150 mil Body), 14-lead	c)	MCP602T-E/ST:	Tape and Reel, Extended Temperature, Dual Op Amp, 8LD TSSOP package.
N.100X.CC	ST = Plastic TSSOP (4.4mm Body), 8, 14-lead	a)	MCP603-I/SN:	Industrial Temperature, Single Op Amp with Chip Select,8LD SOIC package.
		b)	MCP603-E/P:	Extended Temperature, Single Op Amp with Chip Select, 8LD PDIP package.
M.W. 100X	CON.TW WWW.100X.COM	c)	MCP603T-E/ST:	Tape and Reel, Extended Temperature, Single Op Amp with Chip Select, 8LD TSSOP package
胜特力	力材料 886-3-5753170 力电子(上海) 86-21-54151736 力电子(深圳) 86-755-83298787	d)	MCP603T-I/SN:	Tape and Reel, Industrial Temperature, Single Op Amp with Chip Select, 8LD SOIC package.
	ttp://www. 100y. com. tw	a)	MCP604-I/P:	Industrial Temperature, Quad Op Amp, 14LD PDIP package.
		b)	MCP604-E/SL:	Extended Temperature, Quad Op Amp, 14LD SOIC package.
		(c)	MCP604T-I/ST:	Tape and Reel, Industrial Temperature, Quad Op Amp, 14LD TSSOP package.

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