MIC5207

180mA Low-Noise LDO Regulator

General Description

The MIC5207 is an efficient linear voltage regulator with ultralow-noise output, very low dropout voltage (typically 17mV at light loads and 165mV at 150mA), and very low ground current (720µA at 100mA output). The MIC5207 offers better than 3% initial accuracy.

Designed especially for hand-held, battery-powered devices, the MIC5207 includes a CMOS or TTL compatible enable/ shutdown control input. When shutdown, power consumption drops nearly to zero.

Key MIC5207 features include a reference bypass pin to improve its already low-noise performance, reversed-battery protection, current limiting, and overtemperature shutdown.

The MIC5207 is available in fixed and adjustable output voltage versions in a small SOT-23-5 package. Contact Micrel for details.

For low-dropout regulators that are stable with ceramic output capacitors, see the µCap MIC5245/6/7 family.

Features

- Ultra-low-noise output
- High output voltage accuracy
- Guaranteed 180mA output
- Low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
 - Current and thermal limiting
- **Reverse-battery protection**
- "Zero" off-mode current
- Logic-controlled electronic enable

Applications

- Cellular telephones
- Laptop, notebook, and palmtop computers
- Battery-powered equipment
- PCMCIA V_{CC} and V_{PP} regulation/switching
- Consumer/personal electronics
- SMPS post-regulator/dc-to-dc modules
- High-efficiency linear power supplies

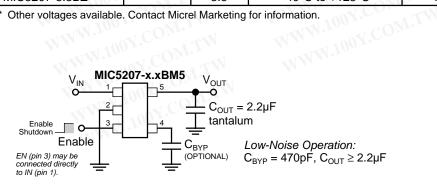
Ord	lering	Inform	ation
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Marking	Voltage	Junction Temp. Range	Package
LEAA	Adj	-40°C to +125°C	SOT-23-5
LE18	1.8	0°C to +125°C	SOT-23-5
LE25	2.5	-40°C to +125°C	SOT-23-5
LE30	3.0	-40°C to +125°C	SOT-23-5
LE33	3.3	-40°C to +125°C	SOT-23-5
LE36	3.6	–40°C to +125°C	SOT-23-5
LE38	3.8	-40°C to +125°C	SOT-23-5
LE40	4.0	-40°C to +125°C	SOT-23-5
LE50	5.0	–40°C to +125°C	SOT-23-5
TIN	3.3	–40°C to +125°C	TO-92
	LEAA LE18 LE25 LE30 LE33 LE36 LE38 LE40	LEAA Adj LE18 1.8 LE25 2.5 LE30 3.0 LE33 3.3 LE36 3.6 LE38 3.8 LE40 4.0 LE50 5.0	LEAA Adj -40°C to +125°C LE18 1.8 0°C to +125°C LE25 2.5 -40°C to +125°C LE30 3.0 -40°C to +125°C LE33 3.3 -40°C to +125°C LE36 3.6 -40°C to +125°C LE38 3.8 -40°C to +125°C LE40 4.0 -40°C to +125°C LE50 5.0 -40°C to +125°C

Other voltages available. Contact Micrel Marketing for information.

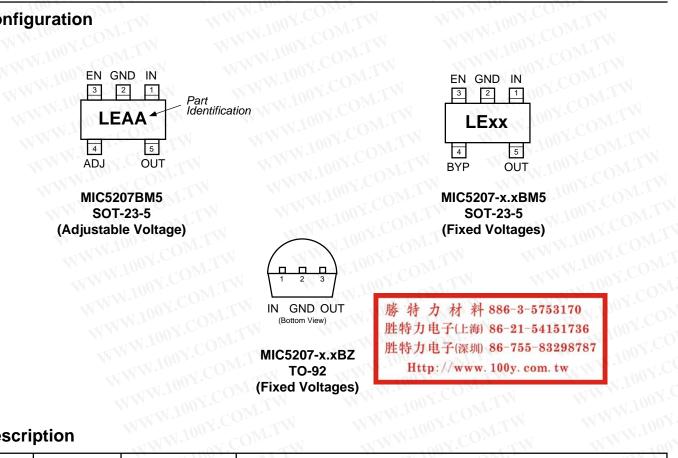
Typical Application

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Battery-Powered Regulator Application

Pin Configuration



Pin Description

Pin No. SOT-23-5	Pin No. TO-92	Pin Name	Pin Function
1	1	N IN 100	Supply Input
2	2	GND	Ground
3		EN	Enable/Shutdown (Input): CMOS compatible input. Logic high = enable, logic low or open = shutdown.
4 (fix)		BYP	Reference Bypass: Connect external 470pF capacitor to GND to reduce output noise. May be left open. For 1.8V or 2.5V operation, see "Applications Information."
4 (adj)		ADJ M	Adjust (Input): Adjustable regulator feedback input. Connect to resistor voltage divider.
5	3	OUT	Regulator Output

WW 100	Y.COM.TY
Absolute Maximum Ratings (Note 1)	Operati
$\begin{array}{llllllllllllllllllllllllllllllllllll$	Input Volta Enable Inp Junction T all excep 1.8V onl Thermal R

Operating Ratings (Note 2)	
Input Voltage (V _{IN}) Enable Input Voltage (V _{EN}) Junction Temperature (T _I)	+2.5V to +16V 0V to V _{IN}
all except 1.8V	
1.8V only	
Thermal Resistance (θ_{JA})	Note 3

 $V_{IN} = V_{OUT} + 1V; I_L = 100\mu\text{A}; C_L = 1.0\mu\text{F}; V_{EN} \ge 2.0V; T_J = 25^\circ\text{C}, \text{ bold } \text{values indicate } -40^\circ\text{C} \le T_J \le +125^\circ\text{C} \text{ except } 0^\circ\text{C} \le T_J \le +125^\circ\text{C} \text{ for } 1.8V \text{ version; unless noted.}$

Symbol	Parameter	Conditions	Min	Typical	Max	Units
Vo	Output Voltage Accuracy	variation from specified V _{OUT}	-3 -4	L.COM	3 4	% %
ΔV _O /ΔT	Output Voltage Temperature Coefficient	Note 4	1.WV	40	M.TW	ppm/°C
ΔV _O /V _O	Line Regulation	$V_{IN} = V_{OUT} + 1V$ to 16V	WW.	0.005	0.05 0.10	%/V %/V
$\Delta V_{O}/V_{O}$	Load Regulation	I _L = 0.1mA to 150mA, Note 5	NW Y	0.05	0.5 0.7	% %
V _{IN} – V _O	Dropout Voltage, Note 6	I _L = 100μA I _L = 50mA	WW	17 115	60 80 175	mV mV mV
	WWW.100Y.COM.TW WWW.100Y.COM.TV WWW.100Y.COM.TV	$I_L = 100 \text{mA}$ $I_L = 150 \text{mA}$	N.	140 165	250 280 325 300 400	mV mV mV mV mV
I _{GND}	Quiescent Current	$V_{EN} \le 0.4V$ (shutdown) $V_{EN} \le 0.18V$ (shutdown)	I	0.01	1	μA μA
	Ground Pin Current, Note 7	$V_{EN} \ge 2.0V, I_{L} = 100\mu A$		80	130 170	μA
胜特 胜特	特力材料 886-3-5753170 行电子(上海) 86-21-54151736 方电子(深圳) 86-755-83298787 Http://www.100y.com.tw	$I_L = 50mA$ $I_L = 100mA$ $I_L = 150mA$	TW TW M.TW	350 720 1800	650 900 1100 2000 2500 3000	μΑ μΑ μΑ μΑ μΑ μΑ
PSRR	Ripple Rejection	CONTRACTION CONTRACTOR	ON.1	75	VIX	dB
I _{LIMIT}	Current Limit	V _{OUT} = 0V	0 ^{M.3}	320	500	mA
$\Delta V_O / \Delta P_D$	Thermal Regulation	Note 8	COM	0.05	11	%/W
e _{no}	Output Noise	$I_L = 50$ mA, $C_L = 2.2\mu$ F, 470pF from BYP to GND	CON	260		nV√Hz

ENABLE Input

V _{IL}	Enable Input Logic-Low Voltage	regulator shutdown	00X.CC	DM.TV	0.4 0.18	V V
V _{IH}	Enable Input Logic-High Voltage	regulator enabled	2.0	.Mo		V
I _{IL} I _{IH}	Enable Input Current	$V_{IL} \leq 0.4V$ $V_{IL} \leq 0.18V$ $V_{IH} \geq 2.0V$ $V_{IH} \geq 2.0V$	100X. N.100X	0.01 5	-1 -2 20 25	μΑ μΑ μΑ μΑ

Note 1. Exceeding the absolute maximum rating may damage the device.

Note 2. The device is not guaranteed to function outside its operating rating.

Note 3: The maximum allowable power dissipation at any T_A (ambient temperature) is $P_{D(max)} = (T_{J(max)} - T_A) + \theta_{JA}$. Exceeding the maximum allowable power dissipation will cause excessive die temperature, and the regulator will go into thermal shutdown. The θ_{JA} of the SOT-23-5 (M5) is 235°C/W and the TO-92 (Z) is 180°C/W (0.4" leads) or 160°C/W (0.25" leads) soldered to a PC board. See "Thermal Considerations."

Note 4: Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.

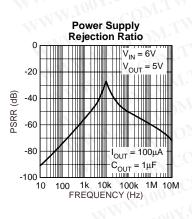
Note 5: Regulation is measured at constant junction temperature using low duty cycle pulse testing. Parts are tested for load regulation in the load range from 0.1mA to 180mA. Changes in output voltage due to heating effects are covered by the thermal regulation specification.

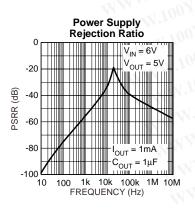
Note 6: Dropout voltage is the input to output differential at which the output voltage drops 2% below its nominal value measured at 1V differential.

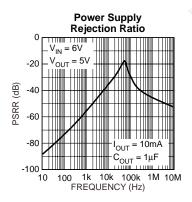
Note 7: Ground pin current is the regulator quiescent current plus pass transistor base current. The total current drawn from the supply is the sum of the load current plus the ground pin current.

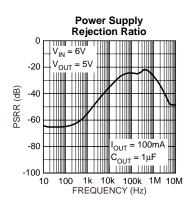
Note 8: Thermal regulation is defined as the change in output voltage at a time "t" after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 180mA load pulse at $V_{IN} = 16V$ for t = 10ms.

Typical Characteristics

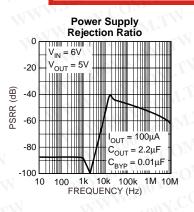


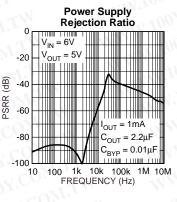












Power Supply Rejection Ratio

I_{OUT} = 10mA

C_{OUT} = 2.2μF

 $C_{BYP} = 0.01 \mu F$

100mA

C_{OUT} = 2.2μF

 $C_{BYP} = 0.01 \mu F$

ουτ

0 1k 10k 100k 1M 10M FREQUENCY (Hz)

Power Supply Rejection Ratio

V_{IN} = 6V

VOUT = 5\

-20

-40

-60

-80

-100

10 100

n

-20

-40

-60

-80

-100

10 100

PSRR (dB)

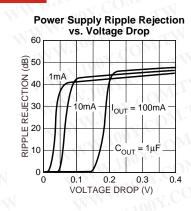
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= 5\ OU

V_{IN} =

(qB)

PSRR (



Power Supply Ripple Rejection vs. Voltage Drop 100 90 (qB) 80 1m/ REJECTION 70 60 I_{OUT} = 100mA 50 40 10m/ RIPPLE 30 20 = 2.2µF OUT $C_{BYP} = 0.01 \mu F$ 10

0.2

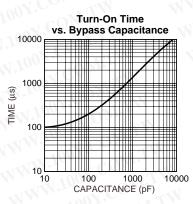
VOLTAGE DROP (V)

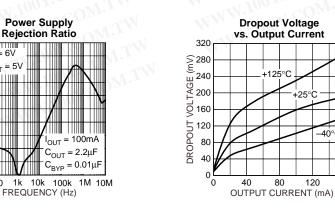
0.3

0.4

0.1

0

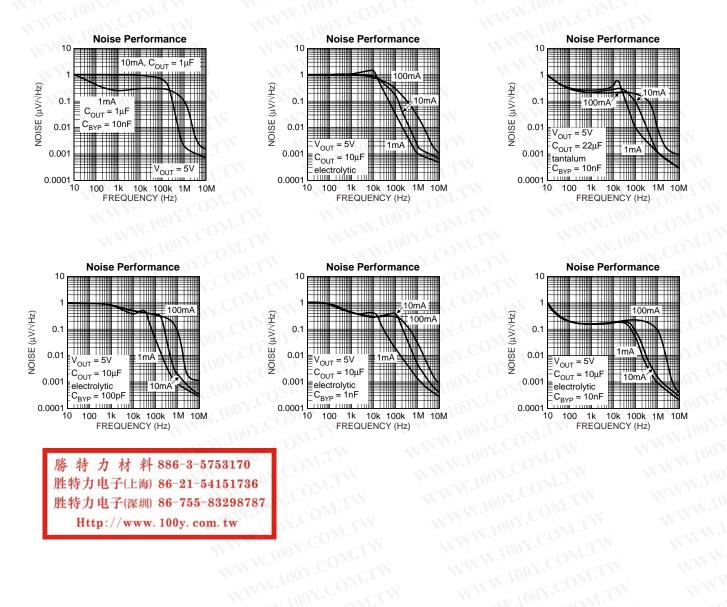




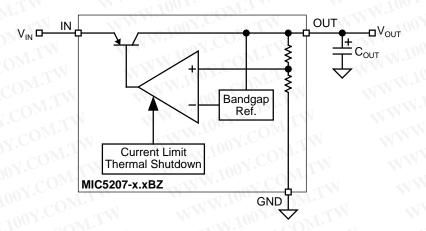
160

-40°C

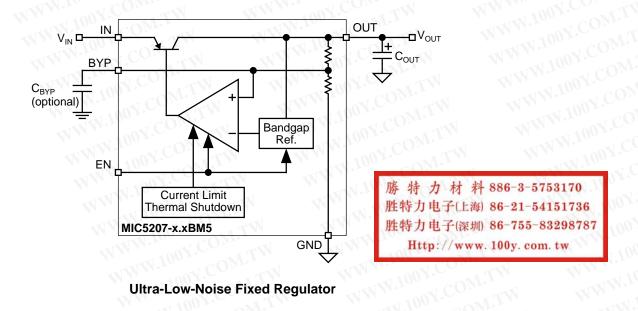
Typical Characteristics

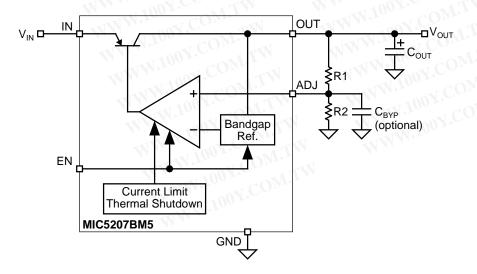


Block Diagrams



Low-Noise Fixed Regulator (TO-92 version only)







Applications Information

Enable/Shutdown

Forcing EN (enable/shutdown) high (> 2V) enables the regulator. EN is compatible with CMOS logic gates.

If the enable/shutdown feature is not required, connect EN (pin 3) to IN (supply input, pin 1). See Figure 1.

Input Capacitor

A 1μ F capacitor should be placed from IN to GND if there is more than 10 inches of wire between the input and the ac filter capacitor or if a battery is used as the input.

Reference Bypass Capacitor

BYP (reference bypass) is connected to the internal voltage reference. A 470pF capacitor (C_{BYP}) connected from BYP to GND quiets this reference, providing a significant reduction in output noise. C_{BYP} reduces the regulator phase margin; when using C_{BYP} , output capacitors of 2.2µF or greater are generally required to maintain stability.

The start-up speed of the MIC5207 is inversely proportional to the size of the reference bypass capacitor. Applications requiring a slow ramp-up of output voltage should consider larger values of $C_{\rm BYP}$. Likewise, if rapid turn-on is necessary, consider omitting $C_{\rm BYP}$.

If output noise is not a major concern, omit $\mathrm{C}_{\mathrm{BYP}}$ and leave BYP open.

Output Capacitor

An output capacitor is required between OUT and GND to prevent oscillation. The minimum size of the output capacitor is dependent upon whether a reference bypass capacitor is used. 1.0 μ F minimum is recommended when C_{BYP} is not used (see Figure 2). 2.2 μ F minimum is recommended when C_{BYP} is 470pF (see Figure 1). Larger values improve the regulator's transient response. The output capacitor value may be increased without limit.

The output capacitor should have an ESR (effective series resistance) of about 5Ω or less and a resonant frequency above 1MHz. Ultra-low-ESR capacitors can cause a low amplitude oscillation on the output and/or underdamped transient response. Most tantalum or aluminum electrolytic capacitors are adequate; film types will work, but are more expensive. Since many aluminum electrolytics have electrolytes that freeze at about -30° C, solid tantalums are recommended for operation below -25° C.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to 0.47μ F for current below 10mA or 0.33μ F for currents below 1mA.

No-Load Stability

The MIC5207 will remain stable and in regulation with no load (other than the internal voltage divider) unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications.

Thermal Considerations

The MIC5207 is designed to provide 180mA of continuous current in a very small package. Maximum power dissipation can be calculated based on the output current and the voltage

drop across the part. To determine the maximum power dissipation of the package, use the junction-to-ambient thermal resistance of the device and the following basic equation:

$$P_{D(max)} = \frac{\left(T_{J(max)} - T_{A}\right)}{\theta_{JA}}$$

 $T_{J(max)}$ is the maximum junction temperature of the die, 125°C, and T_A is the ambient operating temperature. θ_{JA} is layout dependent; Table 1 shows examples of junction-toambient thermal resistance for the MIC5207.

Package	θ _{JA} Recommended Minimum Footprint		θις
SOT-23-5 (M5)	235°C/W	170°C/W	130°C/W

Table 1. SOT-23-5 Thermal Resistance

The actual power dissipation of the regulator circuit can be determined using the equation:

$$\mathsf{P}_\mathsf{D} = (\mathsf{V}_\mathsf{IN} - \mathsf{V}_\mathsf{OUT}) \, \mathsf{I}_\mathsf{OUT} + \mathsf{V}_\mathsf{IN} \, \mathsf{I}_\mathsf{GND}$$

Substituting $P_{D(max)}$ for P_D and solving for the operating conditions that are critical to the application will give the maximum operating conditions for the regulator circuit. For example, when operating the MIC5207-3.3BM5 at room temperature with a minimum footprint layout, the maximum input voltage for a set output current can be determined as follows:

$$\begin{split} \mathsf{P}_{\mathsf{D}(\mathsf{max})} &= \frac{125^\circ \mathsf{C} - 25^\circ \mathsf{C}}{235} \\ \mathsf{P}_{\mathsf{D}(\mathsf{max})} &= 425 \mathsf{mW} \end{split} \begin{matrix} \texttt{B} & \texttt{H} & \texttt{D} & \texttt{H} & \texttt{H} & \texttt{886-3-5753170} \\ \texttt{B} & \texttt{H} & \texttt{P} & \texttt{D} & \texttt{H} & \texttt{H} & \texttt{D} & \texttt{H} & \texttt{H}$$

The junction-to-ambient thermal resistance for the minimum footprint is 220°C/W, from Table 1. The maximum power dissipation must not be exceeded for proper operation. Using the output voltage of 3.3V and an output current of 150mA, the maximum input voltage can be determined. From the Electrical Characteristics table, the maximum ground current for 150mA output current is 3000µA or 3mA.

$$455\text{mW} = (V_{\text{IN}} - 3.3\text{V}) \ 150\text{mA} + V_{\text{IN}} \cdot 3\text{mA}$$
$$455\text{mW} = V_{\text{IN}} \cdot 150\text{mA} - 495\text{mW} + V_{\text{IN}} \cdot 3\text{mA}$$
$$920\text{mW} = V_{\text{IN}} \cdot 153\text{mA}$$

$$V_{IN(max)} = 6.01$$

Therefore, a 3.3V application at 150mA of output current can accept a maximum input voltage of 6V in a SOT-23-5 package. For a full discussion of heat sinking and thermal effects on voltage regulators, refer to the Regulator Thermals section of Micrel's *Designing with Low-Dropout Voltage Regulators* handbook.

Low-Voltage Operation

The MIC5207-1.8 and MIC5207-2.5 require special consideration when used in voltage-sensitive systems. They may momentarily overshoot their nominal output voltages unless appropriate output and bypass capacitor values are chosen.

During regulator power up, the pass transistor is fully saturated for a short time, while the error amplifier and voltage reference are being powered up more slowly from the output (see "Block Diagram"). Selecting larger output and bypass capacitors allows additional time for the error amplifier and reference to turn on and prevent overshoot.

To ensure that no overshoot is present when starting up into a light load (100µA), use a 4.7µF output capacitance and 470pF bypass capacitance. This slows the turn-on enough to allow the regulator to react and keep the output voltage from exceeding its nominal value. At heavier loads, use a 10µF output capacitance and 470pF bypass capacitance. Lower values of output and bypass capacitance can be used, depending on the sensitivity of the system.

Applications that can withstand some overshoot on the output of the regulator can reduce the output capacitor and/ or reduce or eliminate the bypass capacitor. Applications that are not sensitive to overshoot due to power-on reset delays can use normal output and bypass capacitor configurations.

Please note the junction temperature range of the regulator at 1.8V output (fixed and adjustable) is 0°C to +125°C.

Fixed Regulator Applications

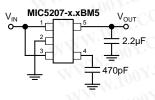


Figure 1. Ultra-Low-Noise Fixed Voltage Regulator

Figure 1 includes a 470pF capacitor for ultra-low-noise operation and shows EN (pin 3) connected to IN (pin 1) for an application where enable/shutdown is not required. $C_{OUT} = 2.2\mu$ F minimum.

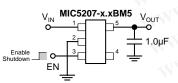


Figure 2. Low-Noise Fixed Voltage Regulator

Figure 2 is an example of a basic low-noise configuration. $C_{OUT} = 1\mu F$ minimum.

Adjustable Regulator Applications

The MIC5207BM5 can be adjusted to a specific output voltage by using two external resistors (figure 3). The resistors set the output voltage based on the following equation:

$$V_{OUT} = V_{REF} \left(1 + \frac{R2}{R1} \right), \quad V_{REF} = 1.242V$$

This equation is correct due to the configuration of the bandgap reference. The bandgap voltage is relative to the output, as seen in the block diagram. Traditional regulators normally have the reference voltage relative to ground; therefore, their equations are different from the equation for the MIC5207BM5.

Resistor values are not critical because ADJ (adjust) has a high input impedance, but for best results use resistors of $470k\Omega$ or less. A capacitor from ADJ to ground provides greatly improved noise performance.

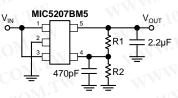


Figure 3. Ultra-Low-Noise Adjustable Voltage Regulator

Figure 3 includes the optional 470pF noise bypass capacitor from ADJ to GND to reduce output noise.

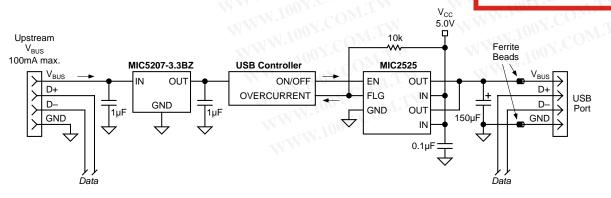
Dual-Supply Operation

When used in dual-supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

USB Application

Figure 4 shows the MIC5207-3.3BZ (3-terminal, TO-92) in a USB application. Since the V_{BUS} supply may be greater than 10 inches from the regulator, a 1µF input capacitor is included.

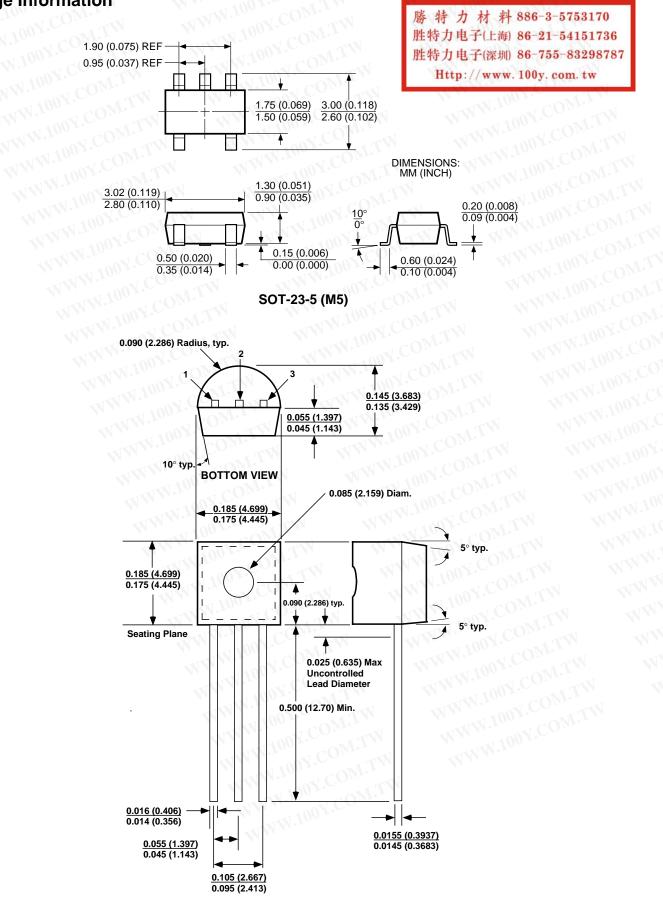
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TO-92 (Z)

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