1.0 A Low-Dropout Positive Fixed and Adjustable Voltage Regulators

The NCP1117 series are low dropout positive voltage regulators that are capable of providing an output current that is in excess of 1.0 A with a maximum dropout voltage of 1.2 V at 800 mA over temperature. This series contains nine fixed output voltages of 1.5 V, 1.8 V, 1.9 V, 2.0 V, 2.5 V, 2.85 V, 3.3 V, 5.0 V, and 12 V that have no minimum load requirement to maintain regulation. Also included is an adjustable output version that can be programmed from 1.25 V to 18.8 V with two external resistors. On chip trimming adjusts the reference/output voltage to within $\pm 1.0\%$ accuracy. Internal protection features consist of output current limiting, safe operating area compensation, and thermal shutdown. The NCP1117 series can operate with up to 20 V input. Devices are available in SOT–223 and DPAK packages.

Features

- Output Current in Excess of 1.0 A
- 1.2 V Maximum Dropout Voltage at 800 mA Over Temperature
- Fixed Output Voltages of 1.5 V, 1.8 V, 1.9 V, 2.0 V, 2.5 V, 2.85 V, 3.3 V, 5.0 V, and 12 V
- Adjustable Output Voltage Option
- No Minimum Load Requirement for Fixed Voltage Output Devices
- Reference/Output Voltage Trimmed to ±1.0%
- Current Limit, Safe Operating and Thermal Shutdown Protection
- Operation to 20 V Input
- NCV Prefix for Automotive and Other Applications Requiring Site and Control Changes

Input

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• Pb–Free Packages are Available

Applications

• Consumer and Industrial Equipment Point of Regulation

Output

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μE

- Active SCSI Termination for 2.85 V Version
- Switching Power Supply Post Regulation
- Hard Drive Controllers
- Battery Chargers

Input

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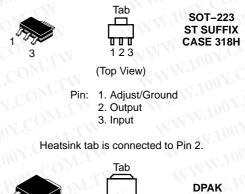
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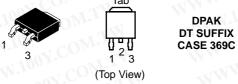
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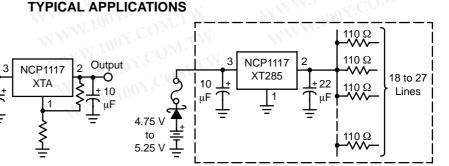
ORDERING INFORMATION

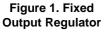
See detailed ordering and shipping information in the package dimensions section on pages 11 and 12 of this data sheet.

DEVICE MARKING INFORMATION

See general marking information in the device marking section on page 14 of this data sheet.







NCP1117

XTXX

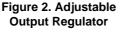


Figure 3. Active SCSI Bus Terminator

MAXIMUM RATINGS

NCP1117, NCV1117	1.WW.10		
WW.100X.COM.TW WW.100X.COM.TW			
MAXIMUM RATINGS Rating	Symbol	Value	Unit
Input Voltage (Note 1)	V _{in}	20	V
Output Short Circuit Duration (Notes 2 and 3)	Wite-	Infinite	1
Power Dissipation and Thermal Characteristics		W.100 CO	1.1
Case 318H (SOT-223)	W WI	1002.0	T.M.
Power Dissipation (Note 2)	PD	Internally Limited	W
Thermal Resistance, Junction-to-Ambient, Minimum Size Pad	R _{0JA}	160	°C/W
Thermal Resistance, Junction-to-Case	R _{θJC}	15	°C/W
Case 369A (DPAK)	NI	VIN. VINW	COw
Power Dissipation (Note 2)	PD	Internally Limited	W
Thermal Resistance, Junction-to-Ambient, Minimum Size Pad	R _{0JA}	67	°C/W
Thermal Resistance, Junction-to-Case	R _{θJC}	6.0	°C/W
Maximum Die Junction Temperature Range	CONT.	-55 to 150	°C
Storage Temperature Range	T _{stg}	-65 to 150	°C
Operating Ambient Temperature Range NCP1117 NCV1117	T _A	0 to +125 -40 to +125	0 °C

Maximum ratings are those values beyond which device damage can occur. Maximum ratings applied to the device are individual stress limit values (not normal operating conditions) and are not valid simultaneously. If these limits are exceeded, device functional operation is not implied, damage may occur and reliability may be affected.

- 1. This device series contains ESD protection and exceeds the following tests: Human Body Model 2000 V per MIL-STD-883, Method 3015. Machine Model Method 200 V.
- WW.100Y.COM 2. Internal thermal shutdown protection limits the die temperature to approximately 175°C. Proper heatsinking is required to prevent activation. $P_{D} = \frac{T_{J(max)} - T_{A}}{T_{J(max)}}$ The maximum package power dissipation is:

$$P_D = \frac{O(Hax)}{R_{0.1A}}$$

WWW.100Y.C

3. The regulator output current must not exceed 1.0 A with Vin greater than 12 V.

WWW.100Y.

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WWW.100Y.C T.MOT NCP1117, NCV1117

WWW.100Y.COM	WWWW. 100X.CO.	WWW.100Y.COM.TW
	NCP1117, NCV1117	
	FICS ($C_{in} = 10 \ \mu$ F, $C_{out} = 10 \ \mu$ F, for typical value T, e that applies unless otherwise noted. (Note 4)	$_{\rm A}$ = 25°C, for min and max values T _A is the

	Characteristic	Symbol	Min	Тур	Max	Unit
Reference Voltage	e, Adjustable Output Devices	V _{ref}		N.100 3	CON	V
(V _{in} –V _{or}	$J_{tt} = 2.0 \text{ V}, \text{ I}_{out} = 10 \text{ mA}, \text{ T}_{A} = 25^{\circ}\text{C}$	N	1.238	1.25	1.262	WT.
(V _{in} –V _{or}	ut = 1.4 V to 10 V, I _{out} = 10 mA to 800 mA) (Note 4)	W	1.225	W 5	1.270	Vn.
Dutput Voltage, Fi	xed Output Devices	V _{out}		VN.10		V
1.5 V	(V _{in} = 3.5 V, I _{out} = 10 mA, T _A = 25 °C)	TYPE	1.485	1.500	1.515	T.Mo
	(V _{in} = 2.9 V to 11.5 V, I _{out} = 0 mA to 800 mA) (Note 4)	W	1.470	MJu.,	1.530	Une
		PT-T		Witten	100 -	-OV.
1.8 V	(V _{in} = 3.8 V, I _{out} = 10 mA, T _A = 25 °C)	WTA	1.782	1.800	1.818	
	(V _{in} = 3.2 V to 11.8 V, I _{out} = 0 mA to 800 mA) (Note 4)	ONL	1.755	W-W	1.845	COM
		-M.L.	-		N.100	- cO
1.9 V	$(V_{in} = 3.9 \text{ V}, I_{out} = 10 \text{ mA}, T_A = 25 ^{\circ}\text{C})$	TT	1.872	1.900	1.929	1.0-
	$(V_{in} = 3.3 \text{ V to } 11.9 \text{ V}, I_{out} = 0 \text{ mA to } 800 \text{ mA})$ (Note 4)	COM	1.862	1.900	1.938	V.C
N	1001. M.T. W. 1001	. Mo			.W.10	
2.0 V	$(V_{in} = 4.0 \text{ V}, I_{out} = 10 \text{ mA}, T_A = 25 \text{ °C})$	1.00	1.970	2.000	2.030	00X.C
	$(V_{in} = 3.4 \text{ V to } 12 \text{ V}, I_{out} = 0 \text{ mA to } 800 \text{ mA})$ (Note 4)	CONT.	1.960		2.040	.Va
0.5.1		No.			New	100 .
2.5 V	$(V_{in} = 4.5 \text{ V}, I_{out} = 10 \text{ mA}, T_A = 25 \text{ °C})$	NY.COT	2.475	2.500	2.525	11005
	$(V_{in} = 3.9 \text{ V to } 10 \text{ V}, \text{ I}_{out} = 0 \text{ mA to } 800 \text{ mA,}) \text{ (Note 4)}$		2.450	-	2.550	N.10
2.85 V	(V _{in} = 4.85 V, I _{out} = 10 mA, T _A = 25 °C)	1001.	0.004	0.050	0.070	W.100
2.05 V	$(V_{in} = 4.25 \text{ V} \text{ to } 10 \text{ V}, I_{out} = 0 \text{ mA to } 800 \text{ mA})$ (Note 4)	NOY.C	2.821	2.850	2.879	11
	$(V_{in} = 4.0 \text{ V}, I_{out} = 0 \text{ mA to 500 mA})$ (Note 5)	100	2.790	-	2.910	$\Lambda M \cdot r$
		a 100 Y.	2.790		2.910	
3.3 V	(V _{in} = 5.3 V, I _{out} = 10 mA, T _A = 25 °C)	Yook	3.267	3.300	3.333	
	$(V_{in} = 4.75 \text{ V to } 10 \text{ V}, I_{out} = 0 \text{ mA to } 800 \text{ mA})$ (Note 4)	W.100	3.235	5.500	3.365	WW
		100	5.255	17.1	5.505	
5.0 V	(V _{in} = 7.0 V, I _{out} = 10 mA, T _A = 25 °C)	WW.	4.950	5.000	5.050	WW.
	(V _{in} = 6.5 V to 12 V, I _{out} = 0 mA to 800 mA) (Note 4)	W.IU	4.900	-	5.100	
				M.TV	000	
12 V	(V _{in} = 14 V, I _{out} = 10 mA, T _A = 25 °C)	VIN VI.	11.880	12.000	12.120	W
	(V _{in} = 13.5 V to 20 V, I_{out} = 0 mA to 800 mA) (Note 4)	WW.	11.760	01.	12.240	
ne Regulation (N	Note 5) Adjustable ($V_{in} = 2.75$ V to 16.25 V, $I_{out} = 10$ mA)	Reg _{line}	100	0.04	0.1	%
Ŭ (Jine	1001		WT	-
1.5 V 1.8 V	(V _{in} = 2.9 V to 11.5 V, I _{out} = 0 mA) (V _{in} = 3.2 V to 11.8 V, I _{out} = 0 mA)	WIN	N. 1-	0.3 0.4	1.0 1.0	mV
1.8 V 1.9 V	$(V_{in} = 3.2 \text{ V to 11.8 V}, I_{out} = 0 \text{ mA})$ $(V_{in} = 3.3 \text{ V to 11.9 V}, I_{out} = 0 \text{ mA})$		W. <u>1</u> 00	0.4	2.5	
1.9 V 2.0 V	$(V_{in} = 3.3 \text{ V to } 11.9 \text{ V}, I_{out} = 0 \text{ mA})$ $(V_{in} = 3.4 \text{ V to } 12 \text{ V}, I_{out} = 0 \text{ mA})$	WW	1	0.5	2.5	
2.0 V 2.5 V	$(V_{in} = 3.4 \text{ V to } 12 \text{ V}, I_{out} = 0 \text{ mA})$ $(V_{in} = 3.9 \text{ V to } 10 \text{ V}, I_{out} = 0 \text{ mA})$		WW.10	0.5	2.5	
2.85 V	$(V_{in} = 4.25 \text{ V to 10 V}, I_{out} = 0 \text{ mA})$		1.1	0.5	3.0	
3.3 V	$(V_{in} = 4.75 \text{ V to 15 V}, I_{out} = 0 \text{ mA})$	V		0.8	4.5	N
5.0 V	$(V_{in} = 6.5 \text{ V to 15 V}, I_{out} = 0 \text{ mA})$	1	WIN	0.8	6.0	
12 V	$(V_{in} = 13.5 \text{ V to } 20 \text{ V}, I_{out} = 0 \text{ mA})$	· · ·		1.0	7.5	T.A.
		Der				0/
oad Regulation (Reg _{line}	W	0.2	0.4	%
1.5 V	(I _{out} = 0 mA to 800 mA, V _{in} = 2.9 V)	L M	12 11	2.3	5.5	mV
1.8 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 3.2 \text{ V})$	W		2.6	6.0	
1.9 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 3.3 \text{ V})$		- "	2.7	6.0	
2.0 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 3.4 \text{ V})$	MT.	-	3.0	6.0	
2.5 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 3.9 \text{ V})$		-	3.3	7.5	
2.85 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 4.25 \text{ V})$		-	3.8	8.0	
3.3 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 4.75 \text{ V})$		-	4.3	10	
5.0 V	$(I_{out} = 0 \text{ mA to } 800 \text{ mA}, V_{in} = 6.5 \text{ V})$		-	6.7	15	
12 V	(I _{out} = 0 mA to 800 mA, V _{in} = 13.5 V)		-	16	28	

4. The regulator output current must not exceed 1.0 A with V_{in} greater than 12 V.
5. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.

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	FICS ($C_{in} = 10 \ \mu$ F, $C_{out} = 10 \ \mu$ F, for typical value T _A e that applies unless otherwise noted. (Note 6)	= 25°C, for min and max values T_A is the

	Symbol	Min	Тур	Max	Uni
Dropout Voltage (Measured at V _{out} – 100 mV)	V _{in} –V _{out}		V.100×	CON	V
$(I_{out} = 100 \text{ mA})$		NT.N.	0.95	1.10	T
$(I_{out} = 500 \text{ mA})$	-	-	1.01	1.15	11.
$(I_{out} = 800 \text{ mA})$	T	7.	1.07	1.20	M.
Output Current Limit (V _{in} –V _{out} = 5.0 V, T _A = 25°C, Note 6)	I _{out}	1000	1500	2200	mA
Minimum Required Load Current for Regulation, Adjustable Output Devices $(V_{in} = 15 \text{ V})$	I _{L(min)}	- 7	0.8	5.0	mA
Quiescent Current	Ι _Q		W	100	mA
1.5 V (V _{in} = 11.5 V)	WT.M.	-	3.6	10	
1.8 V (V _{in} = 11.8 V)	Dir.	-	4.2	10	V.C
1.9 V (V _{in} = 11.9 V)	M.L	_	4.3	10	1
2.0 V (V _{in} = 12 V)	. Cont	- 12	4.5	10	01.
2.5 V (V _{in} = 10 V)	COM	- T	5.2	10	1
2.85 V (V _{in} = 10 V)	Y.	<u></u>	5.5	10	00 1
3.3 V (V _{in} = 15 V)	V.CUM	4	6.0 <	10	
5.0 V $(V_{in} = 15 V)$	COM	<u> </u>	6.0	10	700
12 V $(V_{in} = 20 \text{ V})$	OV.CO	N.F.W	6.0	10	110
Thermal Regulation ($T_A = 25^{\circ}C$, 30 ms Pulse)	NOY.CO	VT-1	0.01	0.1	%/V
Ripple Rejection (V _{in} -V _{out} = 6.4 V, I _{out} = 500 mA, 10 V _{pp} 120 Hz Sinewave)	RR			NN	dB
Adjustable	100 F	67	73	-	W.
1.5 V	1004.0	66	72		
1.8 V	(. L	66	70		WW
1.9 V	N.1001.	66	72	_	
2.0 V	1005	64	70		
2.5 V	W.IVV	62	68	_	
2.85 V	100	62	68	_	
3.3 V	WW.	60	64	_	N
5.0 V	-W.10	57	61		
12 V	WW TI	50	54	<u>_</u>	N
Adjustment Pin Current (V _{in} = 11.25 V, I _{out} = 800 mA)	l _{adj}	10 9 7.	52	120	μA
Adjust Pin Current Change	ΔI_{adj}	107	0.4	5.0	μA
$(V_{in}-V_{out} = 1.4 \text{ V to 10 V}, I_{out} = 10 \text{ mA to 800 mA})$	adj	1.100	I.COM		Ĺ
Temperature Stability	ST	N 700	0.5	1. <u>-</u>	%
Long Term Stability (T _A = 25°C, 1000 Hrs End Point Measurement)	St	NN ^{+.10}	0.3	W-T	%
	N	-1	0.003	T.A.	%Vo

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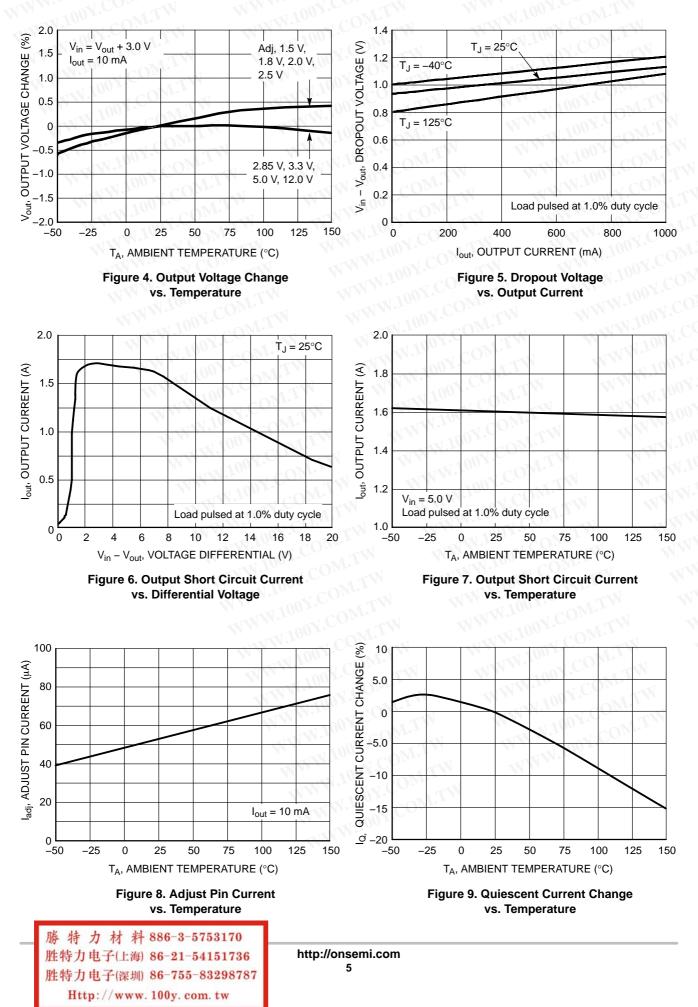
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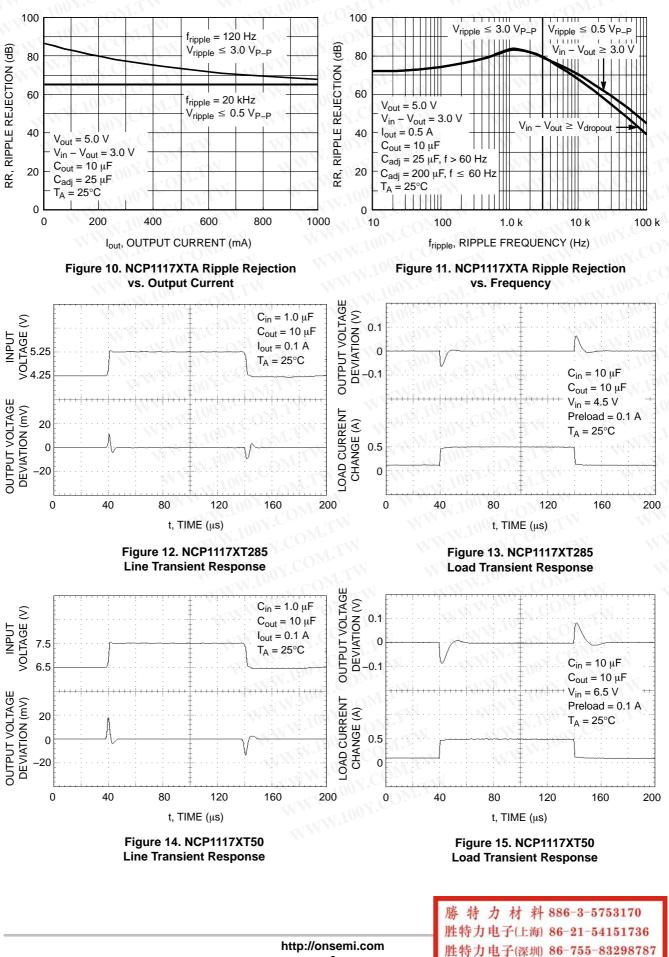
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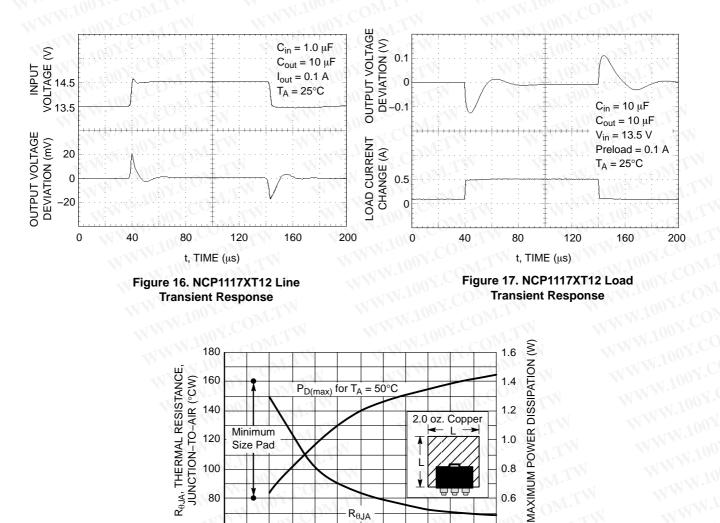
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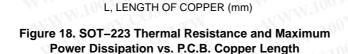
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 $R_{\theta JA}$

15

20

1.2

1.0

0.8

0.6

0.4

30

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2.0 oz. Copper

25

140

120

100

80

60

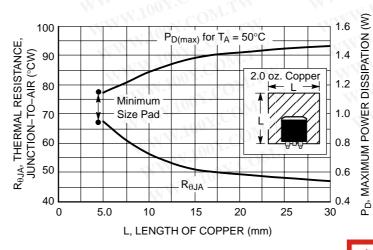
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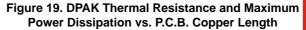
Minimum

Size Pad

5.0

10





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APPLICATIONS INFORMATION

Introduction

The NCP1117 features a significant reduction in dropout voltage along with enhanced output voltage accuracy and temperature stability when compared to older industry standard three-terminal adjustable regulators. These devices contain output current limiting, safe operating area compensation and thermal shutdown protection making them designer friendly for powering numerous consumer and industrial products. The NCP1117 series is pin compatible with the older LM317 and its derivative device types.

Output Voltage

The typical application circuits for the fixed and adjustable output regulators are shown in Figures 20 and 21. The adjustable devices are floating voltage regulators. They develop and maintain the nominal 1.25 V reference voltage between the output and adjust pins. The reference voltage is programmed to a constant current source by resistor R1, and this current flows through R2 to ground to set the output voltage. The programmed current level is usually selected to be greater than the specified 5.0 mA minimum that is required for regulation. Since the adjust pin current, Iadj, is significantly lower and constant with respect to the programmed load current, it generates a small output voltage error that can usually be ignored. For the fixed output devices R1 and R2 are included within the device and the ground current Ignd, ranges from 3.0 mA to 5.0 mA depending upon the output voltage.

External Capacitors

Input bypass capacitor C_{in} may be required for regulator stability if the device is located more than a few inches from the power source. This capacitor will reduce the circuit's sensitivity when powered from a complex source impedance and significantly enhance the output transient response. The input bypass capacitor should be mounted with the shortest possible track length directly across the regulator's input and ground terminals. A 10 μ F ceramic or tantalum capacitor should be adequate for most applications.

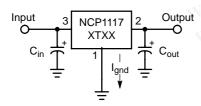


Figure 20. Fixed Output Regulator

Frequency compensation for the regulator is provided by capacitor C_{out} and its use is mandatory to ensure output stability. A minimum capacitance value of 4.7 μ F with an equivalent series resistance (ESR) that is within the limits of 0.25 Ω to 2.2 Ω is required. The capacitor type can be ceramic, tantalum, or aluminum electrolytic as long as it meets the minimum capacitance value and ESR limits over the circuit's entire operating temperature range. Higher values of output capacitance can be used to enhance loop stability and transient response with the additional benefit of reducing output noise.

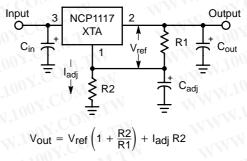


Figure 21. Adjustable Output Regulator

The output ripple will increase linearly for fixed and adjustable devices as the ratio of output voltage to the reference voltage increases. For example, with a 12 V regulator, the output ripple will increase by 12 V/1.25 V or 9.6 and the ripple rejection will decrease by 20 log of this ratio or 19.6 dB. The loss of ripple rejection can be restored to the values shown with the addition of bypass capacitor C_{adj} , shown in Figure 21. The reactance of C_{adj} at the ripple frequency must be less than the resistance of R1. The value of R1 can be selected to provide the minimum required load current to maintain regulation and is usually in the range of 100 Ω to 200 Ω .

$$C_{adj} > \frac{1}{2 \pi fripple R1}$$

The minimum required capacitance can be calculated from the above formula. When using the device in an application that is powered from the AC line via a transformer and a full wave bridge, the value for C_{adj} is:

 $f_{ripple} = 120 \text{ Hz}, \text{ R1} = 120 \Omega$, then $C_{adj} > 11.1 \mu F$

The value for C_{adj} is significantly reduced in applications where the input ripple frequency is high. If used as a post regulator in a switching converter under the following conditions:

 $f_{ripple} = 50 \text{ kHz}, \text{ R1} = 120 \Omega, \text{ then } C_{adi} > 0.027 \mu\text{F}$

Figures 10 and 11 shows the level of ripple rejection that is obtainable with the adjust pin properly bypassed.

Protection Diodes

The NCP1117 family has two internal low impedance diode paths that normally do not require protection when used in the typical regulator applications. The first path connects between V_{out} and V_{in} , and it can withstand a peak surge current of about 15 A. Normal cycling of V_{in} cannot generate a current surge of this magnitude. Only when V_{in} is shorted or crowbarred to ground and C_{out} is greater than 50 µF, it becomes possible for device damage to occur. Under these conditions, diode D1 is required to protect the device. The second path connects between C_{adj} and V_{out} , and it can withstand a peak surge current of about 150 mA. Protection diode D2 is required if the output is shorted or crowbarred to ground and C_{adj} is greater than 1.0 µF.

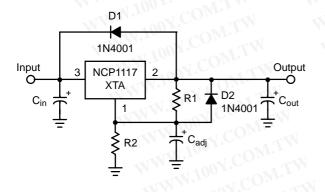


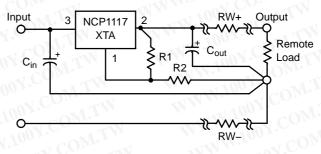
Figure 22. Protection Diode Placement

A combination of protection diodes D1 and D2 may be required in the event that V_{in} is shorted to ground and C_{adj} is greater than 50 µF. The peak current capability stated for the internal diodes are for a time of 100 µs with a junction temperature of 25°C. These values may vary and are to be used as a general guide.

Load Regulation

The NCP1117 series is capable of providing excellent load regulation; but since these are three terminal devices, only partial remote load sensing is possible. There are two conditions that must be met to achieve the maximum available load regulation performance. The first is that the top side of programming resistor R1 should be connected as close to the regulator case as practicable. This will minimize the voltage drop caused by wiring resistance RW + from appearing in series with reference voltage that is across R1.

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Thermal Considerations

This series contains an internal thermal limiting circuit that is designed to protect the regulator in the event that the maximum junction temperature is exceeded. When activated, typically at 175°C, the regulator output switches off and then back on as the die cools. As a result, if the device is continuously operated in an overheated condition, the output will appear to be oscillating. This feature provides protection from a catastrophic device failure due to accidental overheating. It is not intended to be used as a substitute for proper heatsinking. The maximum device power dissipation can be calculated by:

$$P_{D} = \frac{T_{J(max)} - T_{A}}{R_{\theta}JA}$$

The devices are available in surface mount SOT–223 and DPAK packages. Each package has an exposed metal tab that is specifically designed to reduce the junction to air thermal resistance, $R_{\theta JA}$, by utilizing the printed circuit board copper as a heat dissipater. Figures 18 and 19 show typical $R_{\theta JA}$ values that can be obtained from a square pattern using economical single sided 2.0 ounce copper board material. The final product thermal limits should be tested and quantified in order to insure acceptable performance and reliability. The actual $R_{\theta JA}$ can vary considerably from the graphs shown. This will be due to any changes made in the copper aspect ratio of the final layout, adjacent heat sources, and air flow.

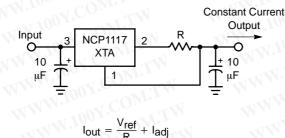
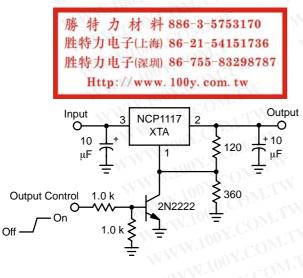


Figure 24. Constant Current Regulator



Vout(Off) = Vref



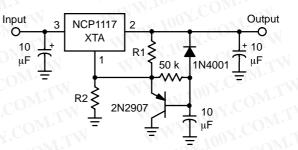
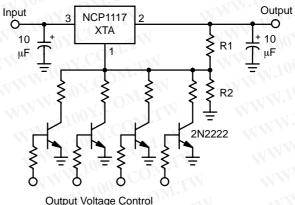


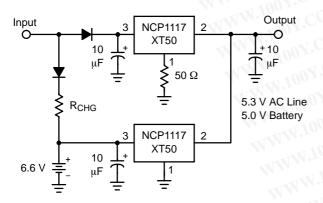
Figure 25. Slow Turn-On Regulator



Culput voltage Control

Resistor R2 sets the maximum output voltage. Each transistor reduces the output voltage when turned on.

Figure 27. Digitally Controlled Regulator



The 50 Ω resistor that is in series with the ground pin of the upper regulator level shifts its output 300 mV higher than the lower regulator. This keeps the lower regulator off until the input source is removed.

Figure 28. Battery Backed–Up Power Supply

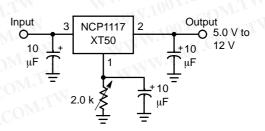


Figure 29. Adjusting Output of Fixed Voltage Regulators

WWW.100Y.C NCP1117, NCV1117

WWW.100Y.COM.TW 100Y.COM.TW ORDERING INFORMATION

RDERING INFORMATION				
Device	Nominal Output Voltage	Package	Shipping [†]	
NCP1117DTA	WWW.Ioo	DPAK	WWW.200.YOU	
NCP1117DTAG	TN WWW.100	DPAK (Pb-Free)	75 Units / Rail	
NCP1117DTARK	WWW.	DPAK	WWWWWWWWWWWWWWWWWWWWWWWWW	
NCP1117DTARKG	NITH MWWW.	DPAK (Pb-Free)	TTO TO DUI OM	
NCP1117DTAT5	Adjustable	DPAK	- 2500 / Tape & Reel	
NCP1117DTAT5G	WWW WILLIAM	DPAK (Pb-Free)	WWW.100Y.COM	
NCP1117STAT3	WILL WILL	SOT-223	W 1002.00	
NCP1117STAT3G	WT.MO	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT12	I. WILL M.	DPAK	75 Units / Rail	
NCP1117DT12G	ON COMPT	DPAK (Pb–Free)		
NCP1117DT12RK	OOT. COM.I'M	DPAK	2500 / Tape & Reel	
NCP1117DT12RKG	12	DPAK (Pb–Free)		
NCP1117ST12T3	V.IUV COM. L	SOT-223	4000 / Tape & Reel	
NCP1117ST12T3G	W1001.COM.TW	SOT-223 (Pb-Free)		
NCP1117DT15	NN. LOUV.COM	DPAK	WWWWWWWWW	
NCP1117DT15G	WW.100 P.COM.I	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT15RK	NYVN. LOOY.COMP. TY	DPAK	N.CO. IN WWW	
NCP1117DT15RKG	1.5	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117ST15T3	100X.COMPT	SOT-223	100Y.COM.TW	
NCP1117ST15T3G	WWW.100Y.COM	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT18	W 1002. COM	DPAK	N.1001. COM.1 V	
NCP1117DT18G	WWW.100Y.COM	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT18RK	WW.1001.200	DPAK	NW.100 COM. I	
NCP1117DT18RKG	1.8	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117DT18T5	1.0	DPAK		
NCP1117DT18T5G	WWW.100	DPAK (Pb–Free)	WWW.1003.COM.IW	
NCP1117ST18T3	WWW.IL	SOT-223	WWW. CON.COM	
NCP1117ST18T3G	WWW.Wo	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT19RKG	1.9	DPAK (Pb–Free)	2500 / Tape & Reel	

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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DERING INFORMATION	N Nominal Output Voltage	Package	Shipping [†]	
NCP1117DT20		DPAK	Compping	
NCP1117DT20G	TR WWW.100	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT20RK	NWW WWW	DPAK	WWW. COM	
NCP1117DT20RKG	2.0	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117ST20T3	WWW WT	SOT-223	WW 100Y.	
NCP1117ST20T3G	COL. WWW	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT25	NT WILLING	DPAK	W	
NCP1117DT25G	NW WILLOW	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT25RK	OT. ONLTW W	DPAK	W.100 .	
NCP1117DT25RKG	00) COM TW	DPAK (Pb–Free)	2500 / Tana & Daal	
NCP1117DT25T5	2.5	DPAK	2500 / Tape & Reel	
NCP1117DT25T5G	N.10N.COM.TW	DPAK (Pb–Free)	LIW WWW.I	
NCP1117ST25T3	W.JOOL COM.I.	SOT-223	MAN WWW.	
NCP1117ST25T3G	NW 100X.COM.TW	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT285	WW.LOW.COM	DPAK	NW WT	
NCP1117DT285G	WWW.100 P.COM. I	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT285RK	WYW. LOOY.COMPTY	DPAK	Y.CO. TW W	
NCP1117DT285RKG	2.85	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117ST285T3	TAN TOOX.COMPT	SOT-223	OOY.COM.TW	
NCP1117ST285T3G	WWW.100Y.COM.T	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCP1117DT33	WWW 100Y.COM	DPAK	V.1001.COM.TW	
NCP1117DT33G	WWW.100Y.CO	DPAK (Pb–Free)	75 Units / Rail	
NCP1117DT33RK	WW.1001.00	DPAK	WW.100 COM. L	
NCP1117DT33RKG	3.3	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117DT33T5		DPAK		
NCP1117DT33T5G	WWW.100	DPAK (Pb-Free)	WWW.100Y.COM.1	
NCP1117ST33T3	WWW.IO	SOT-223	WWW. LOOY.COM	
NCP1117ST33T3G	100 MW.100	SOT-223 (Pb-Free)	4000 / Tape & Reel	

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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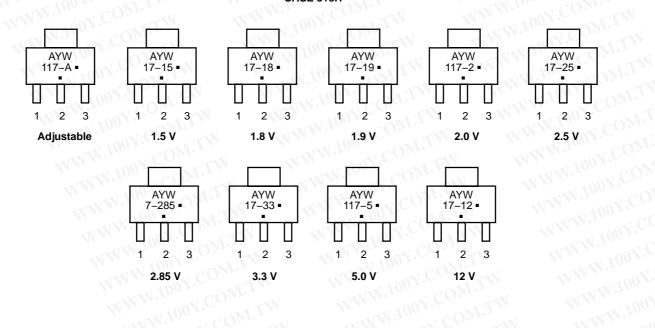
RDERING INFORMATION				
Device	Nominal Output Voltage	Package	Shipping [†]	
NCP1117DT50	NWW.Look	DPAK	WWW. COM. TW	
NCP1117DT50G	N WWW.100	DPAK (Pb-Free)	75 Units / Rail	
NCP1117DT50RK	WWW.L	DPAK	WWWWWWWWWWWWWW	
NCP1117DT50RKG	5.0	DPAK (Pb–Free)	2500 / Tape & Reel	
NCP1117ST50T3	TW WW	SOT-223	WW 100X.COM.	
NCP1117ST50T3G	MIN WWW	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCV1117DTARK*	WILM WW	DPAK	W W. 1001.CO	
NCV1117DTARKG*	Adjustable	DPAK (Pb–Free)	2500 / Tape & Reel	
NCV1117STAT3*	Aujustable	SOT-223	.1 WW.100 F	
NCV1117STAT3G*	CONTRACTION N	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCV1117ST12T3*	COMIT	SOT-223		
NCV1117ST12T3G*	12	SOT-223 (Pb-Free)	TW WWW.100	
NCV1117DT15RK*	JUO COM.	DPAK	ON WWW.re	
NCV1117DT15RKG*	1.5	DPAK (Pb-Free)	2500 / Tape & Reel	
NCV1117ST15T3*	V. LOUIS IN	SOT-223	4000 / Tape & Reel	
NCV1117ST15T3G*	W.100 COM.LTW	SOT-223 (Pb-Free)		
NCV1117DT18T5*	N. M. MOY.COM TV	DPAK	MT.W.	
NCV1117DT18T5G*	1.8	DPAK (Pb-Free)	2500 / Tape & Reel	
NCV1117DT20RK*	MM 100Y.COM	DPAK	-1001.0 _001.1 kpc	
NCV1117DT20RKG*	2.0	DPAK (Pb–Free)	W.100Y.COM.TW W	
NCV1117ST20T3*	W.1001.COM	SOT-223	V.100 P. CONCL	
NCV1117ST20T3G*	WWW.100X.COM	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCV1117DT25RK*	WW.100 1 (0	DPAK	WW.100 COM. A	
NCV1117DT25RKG*	2.5	DPAK (Pb–Free)	2500 / Tape & Reel	
NCV1117ST25T3*	WWW.Ioo	SOT-223	WWW. COM	
NCV1117ST25T3G*	WWW.100	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCV1117DT33T5*	- WWW	DPAK	WWW. 20Y.COM	
NCV1117DT33T5G*	3.3	DPAK (Pb-Free)	2500 / Tape & Reel	
NCV1117ST33T3*		SOT-223		
NCV1117ST33T3G*	3.3	SOT-223 (Pb-Free)	4000 / Tape & Reel	
NCV1117DT50RK*		DPAK		
NCV1117DT50RKG*	5.0	DPAK (Pb-Free)	2500 / Tape & Reel	

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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MARKING DIAGRAMS – NCP PREFIX

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DPAK DT SUFFIX CASE 369C

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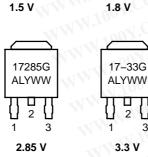
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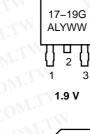
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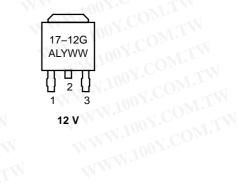
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А	= Assembly Location
L	= Wafer Lot
Y	= Year
WW, N	N = Work Week

• or G = Pb-Free Package

(Note: Microdot may be in either location)

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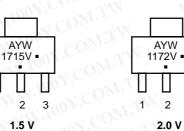
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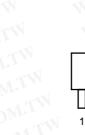
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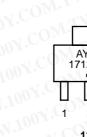
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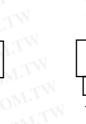
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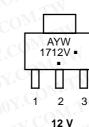
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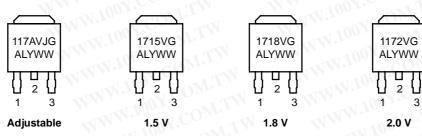
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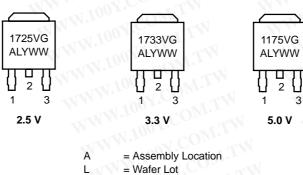


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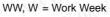


DPAK DT SUFFIX CASE 369C









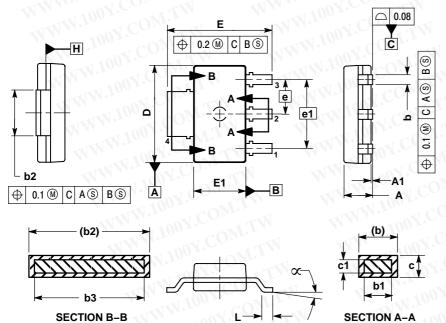


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MILLIMETERS DIM MIN MAX 1.80 Α 0.02 0.11 A1 b 0.60 0.88 b1 0.60 0.80 b2 2.90 3.10 b3 2.90 3.05 0.24 0.35 C c1 0.24 0.30 D 6.30 6.70 Ε 6.70 7.30 E1 3.30 3.70 е 2.30 e1 4.60 0.25 x 10° 0 °

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I. DIMENSIONS ARE IN MILLIMETERS.
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 PER ASME Y14.5M, 1994.

DIMENSION E1 DOES NOT INCLUDE INTERLEAD FLASH OR PROTRUSION. INTERLEAD FLASH OR

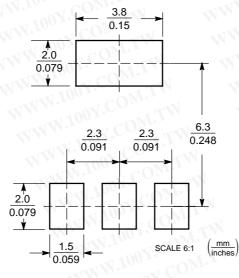
PROTRUSION SHALL NOT EXCEED 0.23 PER

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DAMDAR PROTROSION. ALLOWABLE DAMDAR PROTRUSION SHALL BE 0.08 TOTAL IN EXCESS OF THE 6 AND 62 DIMENSIONS AT MAXIMUM MATERIAL CONDITION. TERMINAL NUMBERS ARE SHOWN FOR

REFERENCE ONLY. DIMENSIONS D AND E1 ARE TO BE DETERMINED AT DATUM PLANE H.

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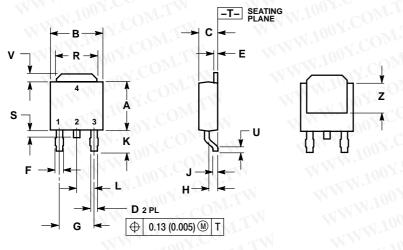


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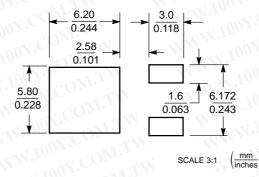


NOTES:

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 CONTROLLING DIMENSION: INCH.

		INC	HES	MILLIN	ETERS			
	DIM	MIN	MAX	MIN	MAX			
	Α	0.235	0.245	5.97	6.22			
	В	0.250	0.265	6.35	6.73			
	С	0.086	0.094	2.19	2.38			
	D	0.027	0.035	0.69	0.88			
1	Е	0.018	0.023	0.46	0.58			
N	F	0.037	0.045	0.94	1.14			
	G	0.180	0 BSC 4.58 I		BSC			
\leq	Н	0.034	0.040	0.87	1.01			
	J	0.018	0.023	0.46	0.58			
	K	0.102	0.114	2.60	2.89			
N	L	L 0.090 BS		2.29	BSC			
	R	0.180	0.215	4.57	5.45			
	S	0.025	0.040	0.63	1.01			
	U	0.020		0.51	. 41 .			
	V	0.035	0.050	0.89	1.27			
	Ζ	0.155		3.93				

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