Advance Information PWM Buck Converter with a Very Low Iq During Low Load Conditions

The NCP1508 is a tri-mode regulator that operates either as a Synchronized PWM Buck Converter, PWM Buck Converter with internal oscillator or as a Pulsed Switching Regulator. If a synchronization signal is present, the NCP1508 operates as a current mode PWM converter with synchronous rectification. The optional external frequency input signal allows the user to control the location of the spurious frequency noise generated by a PWM converter. The Pulsed Switching Regulator mode is active when the Sync Pin is Low. The Pulsed Mode is an extremely low quiescent current Buck Converter. NCP1508 operates in a PWM mode with an internal oscillator when the Sync Pin is held high. The NCP1508 configuration allows the flexibility of efficient high power operation and low input current during system sleep modes.

Features

- Synchronous Rectification for Higher Efficiency in PWM Mode
- Pulsed Switching Mode Operation for Low Current Consumption at Low Loads
- Integrated MOSFETs and Feedback Circuits
- Cycle–by–Cycle Current Limit
- Switching Between PWM, with External or Internal Oscillator, and Pulsed Mode
- Operating Frequency Range of 450 to 1000 kHz
- Internal 1.0 MHz Oscillator
- Thermal Limit Protection
- Built-in Slope Compensation for Current Mode PWM Converter
- 1.0, 1.3, 1.5, 1.89 Fixed Output Voltages
- Shutdown Current Consumption of 0.2 µA
- Pb–Free Package is Available

Applications

- Cellular Phones and Pagers
- PDA
- Digital Cameras
- Supplies for DSP Cores
- Portable Applications



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ORDERING INFORMATION

Device	Package	Shipping†
NCP1508MNR2	10 Pin DFN	3000 Tape & Reel
NCP1508MNR2G	10 Pin DFN (Pb–Free)	3000 Tape & Reel

+For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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This document contains information on a new product. Specifications and information herein are subject to change without notice.





Figure 1. Applications Circuit





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PIN FUNCTION DESCRIPTION

Pin No.	Symbol	Туре	Description
1	GNDA	Analog Ground	Ground connection for the Analog Section of the IC. This is the GND for the Ref, Sync, CB, and SHD pins.
2	GNDP	Power Ground	Ground Connection for the NFET Power Stage.
3	LX	Analog Output	Connection from Power Pass Elements to the Inductor.
4	VCCP	Analog Input	Power Supply Input for the Switching PFET.
5	VCC	Analog Input	Power Supply Input for Power and Analog V _{CC} .
6	FB	Analog Input	Feedback Voltage from the Output of the Power Supply.
7	CB0	Analog Input	Selects Vout. This pin contains an internal pull down resistor.
8	CB1	Analog Input	Selects Vout. This pin contains an internal pull up resistor.
01.9 01.CO	SHD	Analog Input	Enable for Switching Regulator. This Pin is Active High to enable the NCP1 The SHD Pin has an internal pull down resistor to force the converter off if is not connected to the external circuit.
10,10	SYNC	Analog Input	Synchronization input for the PWM converter. If a clock signal is present, th converter uses the rising edge for the turn on. If this pin is low, the converter the Pulsed mode. If this pin is high, the converter uses the internal oscillator the PWM mode. This pin contains an internal pull down resistor.

MAXIMUM RATINGS (Note 1)

Rating	Symbol	Value	Unit
Maximum Voltage All Pins	V _{max}	5.5	V
Maximum Operating Voltage All Pins	V _{max}	5.2	V
Thermal Resistance, Junction-to-Air	Rja	68.5	°C/W
Operating Ambient Temperature Range	T _A	-30 to 85	0°C
SD Withstand Voltage Human Body Model (Note 1) Machine Model (Note 1)	V _{ESD}	> 2500 > 150	L.COVI.
Noisture Sensitivity	MSL	Level 1	N.CO.
Storage Temperature Range	T _{stg}	-55 to 150	°C
lunction Operating Temperature		-30 to 125	°C

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NCP1508
ELECTRICAL CHARACTERISTICS (V_{in} = 3.6 V, V0 = 1.5 V, T_A = 25°C, Fsyh = 600 kHz 50% Duty Cycle square wave for PWW
mode; $T_A = -30$ to 85°C for Min/Max values, unless otherwise noted.

	Cymbol		46	inux	5.111
VCC Pin	NW.IUV C	DWr.		1	
Quiescent Current of Sync Mode, I _{out} = 0 mA	lq PWM	014.1	175	-	μA
Quiescent Current of PWM Mode, I _{out} = 0 mA	lq PWM	- M.T	185	-	μΑ
Quiescent Current of Pulsed Mode, I _{out} = 0 mA	Iq Pulsed		14	-	μΑ
Quiescent Current, SHD Low	Iq Off	V.Com	0.1	1.0	μA
Input Voltage Range	Vin	2.5	VT	5.2	V
Sync Pin					
Input Voltage	Vsync	-0.3	M-	Vcc + 0.3	V
Frequency Operational Range	Fsync	450	600	1000	kHz
Minimum Synchronization Pulse Width	Dcsync Min	1100x.c	30		%
Maximum Synchronization Pulse Width	Dcsync Max	YOUL	70	1 m -	%
SYNC "H" Voltage Threshold	Vsynch	100	920	1200	mV
SYNC "L" Voltage Threshold	Vsyncl	400	830	IFT.	mV
SYNC "H" Input Current, Vsync = 3.6 V	Isynch	WN.10	2.2	AV.	μA
SYNC "L" Input Current, Vsync = 0 V	Isyncl	-0.5	700	M	μA
Output Level Selection Pins	UN I		100.	OM.TY	-1
Input Voltage	Vcb	-0.3	1001.	Vcc + 0.3	V
CB0, CB1 "H" Voltage Threshold	Vcb h	AT.	920	1200	mV
CB0, CB1 "L" Voltage Threshold	Vcb I	400	830		mV
CB0, CB1 "H" Input Current, CB = 3.6 V	lcb h	Va-	2.2	N.COM	μA
CB0, CB1 "L" Input Current, CB = 0 V	Icb I	-0.5	NN.10	A CON	μA
Shutdown Pin	COM.T		M.W.M		N.
Input Voltage	Vshd	-0.3		Vcc + 0.3	V
SHD "H" Voltage Threshold	Vshd h	_	920	1200	mV
SHD "L" Voltage Threshold	Vshd I	400	830	100%	mV
SHD "H" Input Current. SHD = 3.6 V	Ishd h		2.2	1.01	uΑ
SHD "L" Input Current, SHD = 0 V	Ishd I	-0.5	NV a	- 001	μA
Feedback Pin	TONT COM.			NN.	V.C
Input Voltage	Vfb	-0.3	-	Vcc + 0.3	V
Input Current, Vfb = 1.5 V	lfb	<u>1. T</u>	5.0	7.5	μА
Sync PWM Mode Characteristics	1001.00	M.T.W.	0.0		
Switching P_FET Current Limit	Llim	NT2	800	MW Y	mΔ
	DC	07		100	%
Minimum On Time	Ton min	CONT	75		nsec
Rdson Switching P_FET and N_FET	Rdson	COM	0.23		0
	llook	COM	0.25	10	<u>ν</u>
	Ileak	Y	50	10	μA 0/
	VO	0.050	5.0	-	%
Feedback voltage Accuracy, v_{out} Set = 1.0 V CB0 = L, CB1 = L, I_{out} = 300 mA	Vout	0.950	1.000	1.050	V
Feedback Voltage Accuracy, Vout Set = 1.3 V CB0 = L, CB1 = H,	Vout	1.261	1.300	1.339	V

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Characteristic	Symbol	WIN	тур	wax	Unit
Sync PWM Mode Characteristics (continued)	W.100 F	COM.1			
Feedback Voltage Accuracy, V _{out} Set = 1.5 V CB0 = H, CB1 = H, I_{out} = 300 mA	V _{out}	1.450	1.500	1.550	V
Feedback Voltage Accuracy, V _{out} Set = 1.89 V CB0 = H CB1 = L, I_{out} = 300 mA	V _{out}	1.833	1.890	1.947	V
Line Regulation, V _{in} = 2.7 V–3.6 V, I _{out} = 100 mA	WW-	-15	N.T.W	+15	mV
Line Regulation, V _{in} = 3.6 V–5.2 V, I _{out} = 100 mA	AM M	-15	WEI	+15	mV
Load Regulation, I _{out} = 100 mA-300 mA	Mallar.	-15	175	+15	mV
Load Transient Response 10 to 100 mA Load Step	V _{out}	1.100Y.	25	- 17	mV
Line Transient Response, I _{out} = 100 mA 3.0 to 3.6 Vin Line Step	V _{out}	N.1001	±5.0	<u>un</u>	mVpp
PWM Mode with Internal Oscillator Characteristics		WW.100	V.CON	W	
Switching P-FET Current Limit	l lim	10/4/10	800		mA
Duty Cycle	DC	L.W.T	10×	100	%
Minimum On Time	Ton min		75	OVERY	nsec
Internal Oscillator Frequency	Fosc	800	1000	1240	kHz
Rdson Switching P–FET and N_FET	Rdson	Atm	0.23		Ω
Switching P–FET and N–FET Leakage Current	lleak	WW.	0	10	μA
Output Overvoltage Threshold	Vo	-11	5.0	A'COM	%
Feedback Voltage Accuracy, V_{out} Set = 1.0 V CB0 = L, CB1 = L, I_{out} = 300 mA	V _{out}	0.950	1.000	1.050	V
Feedback Voltage Accuracy, V_{out} Set = 1.3 V CB0 = L, CB1 = H, I_{out} = 300 mA	V _{out}	1.261	1.300	1.339	O V V
Feedback Voltage Accuracy, V_{out} Set = 1.5 V CB0 = H, CB1 = H, I_{out} = 300 mA	Vout	1.450	1.500	1.550	O V
Feedback Voltage Accuracy, V_{out} Set = 1.89 V CB0 = H CB1 = L, I_{out} = 300 mA	V _{out}	1.833	1.890	1.947	V
Line Regulation, V _{in} = 2.7 V–3.6 V, I _{out} = 100 mA	100 OM	-15	_	+15	mV
Line Regulation, V _{in} = 3.6 V–5.2 V, I _{out} = 100 mA	100X	-15	_ 1	+15	mV
Load Regulation, I _{out} = 100 mA–300 mA	A. MOX-CON	-15	- 1	+15	mV
Load Transient Response 10 to 100 mA Load Step	Vout	ON.TW	25	WW.	mV
Line Transient Response, I _{out} = 100 mA 3.0 to 3.6 Vin Line Step	V _{out}	$0^{\underline{\Lambda},\Gamma}$	±5.0	WW	mVpp
Pulsed Mode Characteristics	NWW.	COM	W	WW	100
Minimum On Time	Ton min	V.COM	50	-01	nsec
Output Ripple Voltage, I _{out} = 100 μA	V _{out}	COM	40		mV
Feedback Voltage Accuracy, Vout Set = 1.0 V CB0 = L, CB1 = L	V _{out}	0.950	1.000	1.050	V.
Feedback Voltage Accuracy, Vout Set = 1.3 V CB0 = L, CB1 = H	V _{out}	1.261	1.300	1.339	V
Feedback Voltage Accuracy, Vout Set = 1.5 V CB0 = H, CB1 = H	V _{out}	1.450	1.500	1.550	V
Feedback Voltage Accuracy, V _{out} Set = 1.89 V CB0 = H CB1 = L	V _{out}	1.833	1.890	1.947	V
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ELECTRICAL CHARACTERISTICS (continued) (Vin = 3.6 V, Vo = 1.5 V, TA = 25°C, Fsyn = 600 kHz 50% Duty Cycle square wave

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INTRODUCTION

The NCP1508 is a tri-mode regulator intended for use in baseband supplies for portable equipment. Its unique features provide an efficient power supply for a portable device at full operating current, while also providing extremely low standby current for idle mode operation. When the system is idle, the user can activate the pulsed mode function. In this mode, the regulator provides a regulated low current output voltage keeping the system biased. When the device is in its normal operating mode, the regulator synchronizes to the system clock or uses an internal 1.0 MHz clock and turns into a switching regulator. This allows the regulator to provide efficient power to the system. This circuit is patent pending.

Operation Description

The Buck regulator is a synchronous rectifier PWM regulator with integrated MOSFETs. This regulator has a Pulsed function for low power modes to conserve power. The Tri PWM with external or internal oscillator/pulsed mode is an exclusive Patent Pending circuit.

For the PWM Synchronization mode, the operating frequency range for the NCP1508 is 450 to 1000 kHz. The output current of the PWM is optimized for 100 mA with a maximum current supply in excess of 300 mA for the 2.5 to 5.2 input voltage range.

If the Sync Pin is held low, the NCP1508 changes into the Pulsed mode. The Pulsed function assures the user of an extremely low input current and greatly reduced quiescent current when the users system is in a sleep mode. Internally to the NCP1508, the Synchronization pin has a pull down resistor to force the part into Pulsed mode when a clock signal is not present. The Pulsed mode supplies a current in excess of 30 mA.

If the Sync Pin is held high, NCP1508 enters a PWM mode with an internal 1.0 MHz oscillator. The PWM mode has the same operational characteristics (current limit, maximum output current, etc.) as the synchronized PWM mode. The Sync Pin threshold is fixed as noted in the Electrical Characteristics table.

Sync Pin State	Operational Mode			
Low	Low Iq Pulsed Mode Operation			
High	PWM Using Internal Oscillator for the Clock			
Clock	PWM Using Rising Edge of Clock Signal to Turn On PFET Pass Element			

Table 1. Sync Pin Input with CorrespondingOperational Mode of NCP1508

PWM Mode with External Synchronization Signal

During normal operation, a synchronization pulse acts as the clock for the DC/DC controller. The rising edge of the clock pulls the gate of Q1 low allowing the inductor to charge. When the current through Q1 reaches either the current limit or feedback voltage reaches its limit, Q1 will turn off and Q2 will turn on. Q2 replaces the free wheeling diode typically associated with Buck Converters. Q2 will turn off when either a rising edge sync pulse is present or all the stored energy is depleted from the inductor.

The output voltage accuracy in the PWM mode is well within 3% of the nominal set value. An overvoltage protection circuit is present in the PWM mode to limit the positive voltage spike due to fast load transient conditions. If the OVP comparator is activated, the duty cycle will be 0% until the output voltage falls to the nominal level. The PWM also has the ability to go to 100% duty cycle for transient conditions and low input to output voltage differentials.

The PWM mode operates as a forced–PWM converter. Each switching cycle has a typical on–time of 75 nsec. NCP1508 has two protection circuits that can eliminate the minimum on time for the cycle. When tripped, the overvoltage protection or the thermal shutdown overrides the gate drive of the high side MOSFET.

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PWM Mode with Internal Oscillator

If a synchronization signal is not available, the converter has a 1.0 MHz internal oscillator available. The Sync Pin must be held high to enter this mode. The characteristics of the PWM mode with internal oscillator are similar to the Sync PWM Mode.

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Pulsed Mode

During low–level current output, NCP1508 can enter a low current consumption mode when the Sync Pin is held low. This mode will typically have a free running frequency and an output voltage ripple similar to a PFM mode. The advantage of the Pulsed mode is much lower Iq (14 μ A) and drastically higher efficiency compared with PWM and PFM modes in low output loads.

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7



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9

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Voltage Output Selection The CB1 and CB0 pins control the output voltage selection. The output voltages are listed in Table 2. The CB pins contain internal resistors to force the NCP1508 to 1.3 V_{out} if they are not connected to an external circuit. The CB0 has a pull down resistor and the CB1 has a pull up resistor. The CB Pin thresholds are fixed as noted in the Electrical Characteristics table.

Shutdown Pin

The Shutdown Pin enables the operation of the device. The Shutdown Pin has an internal pull down resistor to force the NCP1508 into the off mode if this pin is floating due to the external circuit. The Shutdown Pin threshold is fixed as noted in the Electrical Characteristics table. During Startup, the NCP1508 has a soft start function to limit fast dV/dt and eliminate overshoot on the output.

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Thermal Shutdown

Internal Thermal Shutdown circuitry is provided to protect the integrated circuit in the event at the maximum junction temperature is exceeded. When activated, typically at 160°C, the PWM latch is reset and the linear regulator control circuitry is disabled. The thermal shutdown circuit is designed with 25°C of hysteresis. This means that the PWM latch and the regulator control circuitry cannot be re–enabled until the die temperature drops by this amount. This feature is provided to prevent catastrophic failures from accidental device overheating. It is not intended as a substitute for proper heat–sinking.

Table 2. Truth Table for CB0 and CB1 with theCorresponding Output Voltage

CB0	CB1	V _{out} (V)
0	0	1.0
0	100 100M.	1.3
1	100 1 COM	1.5
1	N.1000.	1.89

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Design Example

First, determine the normal operating conditions of the NCP1508. If one assumes that the NCP1508 is used in a single lithium–ion battery application, the input voltage, V_{in} , is 3.0 V to 4.2 V. Next determine which output voltage is required. Output conditions for this example will be V_{out} at 1.5 V with a typical load current of 120 mA and a maximum of 300 mA.

Since the compensation is fixed internally in the IC, the inductance as well as the input and output capacitors have a predetermined value. The input and output capacitors are 10 μ F. The ESR of the capacitors needs to be as low as possible, therefore, it is recommended that one uses a ceramic capacitor. The inductor must be 6.8 μ H. The series resistance of the inductor only factors into the overall efficiency of the converter. The inductor needs to be selected by the peak current required.

Equation 1 is the basic equation for an inductor and can derive Equations 2 and 3. The equation describes the voltage across the inductor and the inductance value determines the slope of the current of the inductor.

$$\frac{V}{L} = \frac{di}{dt}$$
 (eq. 1)

Equation 1 is rearranged to solve for the change in current for the on-time of the converter in Continuous Conduction Mode.

$$iLpk - pk = \frac{(V_{in} - V_{out}) * t_{on}}{L} * (eq. 2)$$

The only unknown of Equation 2 is the on-time. For a steady state condition at maximum load, the di and L are equal during the on-time and off-time. Manipulating Equation 1 yields the following.

$$t_{\text{ON}} = \frac{V_{\text{in}}}{V_{\text{out}}} * \frac{1}{f}$$
 (eq. 3)

Utilizing Equations 2 and 3, the peak inductor current is calculated. The following worst case conditions are used for the calculations.

$$V_{IN MAX} = 4.2 V$$

$$V_{OUT} = 1.5 V$$
Freq = 1.0 MHz - 20%
$$L = 6.8 \mu H - 10\%$$
luce result in the following

These values result in the following.

$$t_{on} = 446$$
 nsec

The maximum current in the inductor is half of the peak to peak value plus the maximum DC current in the load. The inductor must have a maximum current exceeding 399 mA.

PC Board Layout Criteria

iL

When laying out the printed circuit board, the following checklist should be used to ensure proper operation of the NCP1508.

- Place C_{in} as close as possible to the V_{CC} and GND pins. This will minimize the AC noise into the Reference and Analog Circuitry of the IC generated by the switching in the MOSFETs.
- 2. Keep the GND connections of the capacitors as close as possible.
- 3. The power traces should be short and wide as possible.



PACKAGE DIMENSIONS



For additional information on our Pb–Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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