



# **OPA128**

# **Difet® Electrometer-Grade OPERATIONAL AMPLIFIER**

### **FEATURES**

- ULTRA-LOW BIAS CURRENT: 75fA max
- LOW OFFSET: 500µV max
- LOW DRIFT: 5μV/°C max
- HIGH OPEN-LOOP GAIN: 110dB min
- HIGH COMMON-MODE REJECTION: 90dB min
- IMPROVED REPLACEMENT FOR AD515 AND AD549

# **APPLICATIONS**

- ELECTROMETER
- MASS SPECTROMETER
- CHROMATOGRAPH
- ION GAUGE
- PHOTODETECTOR
- RADIATION-HARD EQUIPMENT

# **DESCRIPTION**

The OPA128 is an ultra-low bias current monolithic operational amplifier. Using advanced geometry dielectrically-isolated FET (*Difet*®) inputs, this monolithic amplifier achieves a performance level exceeding even the best hybrid electrometer amplifiers.

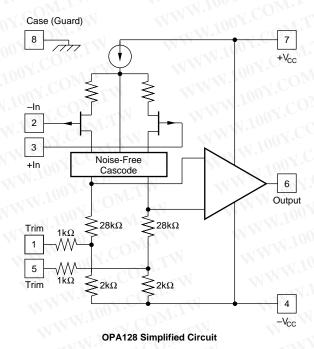
Laser-trimmed thin-film resistors give outstanding voltage offset and drift performance.

A noise-free cascode and low-noise processing give the OPA128 excellent low-level signal handling capabilities. Flicker noise is very low.

The OPA128 is an improved pin-for-pin replacement for the AD515.

Difet® Burr-Brown Corp.

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International Airport Industrial Park • Mailing Address: PO Box 11400, Tucson, AZ 85734 • Street Address: 6730 S. Tucson Blvd., Tucson, AZ 85706 • Tel: (520) 746-1111 • Twx: 910-952-1111 Internet: http://www.burr-brown.com/ • FAXLine: (800) 548-6133 (US/Canada Only) • Cable: BBRCORP • Telex: 066-6491 • FAX: (520) 889-1510 • Immediate Product Info: (800) 548-6132

# **SPECIFICATIONS**

#### **ELECTRICAL**

At  $V_{CC} = \pm 15$ VDC and  $T_A = +25$ °C, unless otherwise noted. Pin 8 connected to ground.

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	CONDITIONS	OPA128JM		M	0	PA128K	M	Ç	PA128L	М	OPA128SM		]	
PARAMETER		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
INPUT	WWW	Co		W		WW	1	1003		TI	N			
BIAS CURRENT(1)	. In	-1 (	Divis.	_41			MW.		J CO	Mr.				
Input Bias Current	$V_{CM} = 0VDC,$		- 1	$L_{M}$		1/1		100	7.	·Mr	1.44			
-1 COM.	$R_L \ge 10k\Omega$		±150	±300		±75	±150		±40	±75	TW	±75	±150	fA
OFFSET CURRENT(1)	77 71 10	0	-01	1.7		-	-11	NAU	0	MO	. 1	-		
Input Offset Current	$V_{CM} = 0VDC,$		Cor	- 11		-	NW		on V.		TT			
ON. IV	$R_L \ge 10k\Omega$		65	$M_{T,T}$		30		W.	30	COI	7	30		fA
OFFSET VOLTAGE(1)	MM	400	1.00	- 11			1/1/4		1007		~13	A		
Input Offset Voltage	V <sub>CM</sub> = 0VDC		±260	±1000		±140	±500	WW	±140	±500	LATO	±140	±500	μV
Average Drift	$T_A = T_{MIN}$ to $T_{MAX}$		M. C.	±20			±10		100	±5	No.		±10	μV/°C
Supply Rejection	A WIIN WAX	80	120	OM	90	120		90	120	. √1 C	90	120		. dB
1007.0			±1	±100		±1	±32	N .	±1 0	±32		±1	±32	μV/V
NOISE	TITE TO THE TENT	M.	~1	$C_{O_h}$		N		NV	44-	N.	J	- 17		
Voltage: f <sub>O</sub> = 10Hz			92	0		92			92	00 -	00	92	-7	nV/√Hz
f <sub>O</sub> = 100Hz	N XI		78	CO		78		W	78	COOL		78	N	nV/√Hz
$f_O = 1 \text{kHz}$			27	- 40		27			27	Too	0	27		nV/√Hz
$f_O = 10kHz$	W W		15	V.C		15		W	15	. 100	Y.C.	15	LAN	nV/√Hz
$f_B = 10Hz$ to $10kHz$			2.4	, (		2.4			2.4	'To	-7	2.4	-31	μVrms
$f_B = 0.1Hz$ to $10Hz$	TW		4			4		1	4	10	01.	4	IM	μVp-p
Current: $f_B = 0.1Hz$ to 10Hz			4.2	, ,		3	≪T		2.3	M.r.	=7	3		fA, p-p
$f_O = 0.1Hz$ to $20kHz$	TW		0.22	001		0.16	N		0.12	-11	00x	0.16	17.	fA/√Hz
IMPEDANCE	1 1		WW.	10		Mr.	N N		- TV	1111.		$^{1}CO_{i}$		W
Differential	TIN		10 <sup>13</sup>    1	100		1013    1			10 <sup>13</sup>    1		100	10 <sup>13</sup>    1		$\Omega \parallel pF$
Common-Mode	Mr.		1015    2			1015    2	W		1015    2	WAR	. 00	1015    2	12.	Ω    pF
VOLTAGE RANGE(4)	$M_{II}$		-11	V 10	7	M	. 7			-31	1.70.		$O_{IM}$	
Common-Mode Input Range	ULI	±10	±12	1	±10	±12	TV	±10	±12	1111	±10	±12	_ 1	V
Common-Mode Rejection	$V_{IN} = \pm 10 VDC$	80	118	$\sim 1.1$	90	118	7.7	90	118	-137	90	118		dB
OPEN-LOOP GAIN, DC	TITI		MA	-31	1007		T.M	AA		111	-TXN.	100 1.	a01	Will
Open-Loop Voltage Gain	$R_L \ge 2k\Omega$	94	128	MAA.	110	128	- K T	110	128	W	110	128		dB
FREQUENCY RESPONSE														
Unity Gain, Small Signal	(2)	0.5	1		0.5	1	. ~ 1	0.5	1	1	0.5	1.1		MHz
Full Power Response	20Vp-p, $R_L = 2kΩ$	KÍ	47	ot W	110	47	CO 21.	000	47	4	VVV	47	N.V	kHz
Slew Rate	$V_O = \pm 10V$ , $R_L = 2k\Omega$	0.5	3	<i>M</i>	1	3	-01	1	3		1	3	0	V/us
Settling Time, 0.1%	Gain = $-1$ , $R_L = 2k\Omega$		5	WIV		5	Co,	- 1	5		WV	5	on V.	μs
0.01%	10V Step		10	71		10	\ \aC	Mi	10			10	00	μs
Overload Recovery,	COP.			W		. 00					W		100	
50% Overdrive <sup>(3)</sup>	Gain = −1	1	5		TV.	5		OMr	5			5	.10	μs
RATED OUTPUT														
Voltage Output	$R_L = 2k\Omega$	±10	±13		±10	±13		±10	±13		±10	±13		V
Current Output	$V_0 = \pm 10 \text{VDC}$	±5	±10		±5	±10	00 2.	±5	±10		±5	±10	W.11	mA
Output Resistance	DC, Open Loop		100			100	. 005		100	W		100	1	Ω
Load Capacitance Stability	Gain = +1		1000			1000	$I_{\Omega\Omega}$ ,	. 00	1000			1000	WW.	pF
Short Circuit Current	NAM.	10	34	55	10	34	55	10	34	55	10	34	55	mA
POWER SUPPLY	WW.100	OM					1.Tac	~ <b>\$</b> 7 (	OM	TAN.		41		.10
Rated Voltage	100X.	~1	±15			±15	ox 10	Or.	±15	1.		±15	-411	VDC
Voltage Range,	WWW.		-	N		WW	44.	N.	COR	TI	N	4	MM,	1
Derated Performance	1007	±5	M.T.	±18	±5	44	±18	±5		±18	±5		±18	VDC
Current, Quiescent	$I_O = 0$ mADC	1.CL	0.9	1.5		0.9	1.5	.00	0.9	1.5	W	0.9	1.5	mA
TEMPERATURE RANGE	W.100	-10	OM.	- XT			WW	Tan	SI CC	) Live	-XX		XXI <sup>X</sup>	MW.
Specification	Ambient Temp.	0	MA	+70	0		+70	0	7	+70	-55		+125	°C
Operating	Ambient Temp.	-55	$\mathbb{C}\Omega_{M}$	+125	-55		+125	-55	J.V.C	+125	-55		+125	°C
Storage	Ambient Temp.	-65	1	+150	-65		+150	-65	10 .	+150	-65		+150	°C
θ Junction-Ambient	- TN . L		200			200	- 1	144.	200	1000		200	I .	°C/W

NOTES: (1) Offset voltage, offset current, and bias current are measured with the units fully warmed up. Bias current doubles approximately every 11°C. (2) Sample tested. (3) Overload recovery is defined as the time required for the output to return from saturation to linear operation following the removal of a 50% input overdrive. (4) If it is possible for the input voltage to exceed the supply voltage, a series protection resistor should be added to limit input current to 0.5mA. The input devices can withstand overload currents of 0.3mA indefinitely without damage.

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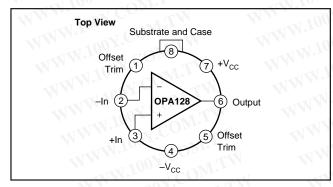
#### **ELECTRICAL (FULL TEMPERATURE RANGE SPECIFICATIONS)**

At  $V_{CC} = \pm 15$ VDC and  $T_A = T_{MIN}$  and  $T_{MAX}$ , unless otherwise noted.

10	CONDITIONS	0	PA128	JM	OPA128KM		OPA128LM			OPA128SM				
PARAMETER		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
TEMPERATURE RANGE	ON CONTRACT	TW		W	MA	- 100	Y.C.	31	TW					
Specification Range	Ambient Temp.	0		+70	0	1.2	+70	0	TW	+70	-55		+125	°C
INPUT	V.100 2	1.7			- 1	W.10	V -	CON	1.1	«T				
BIAS CURRENT <sup>(1)</sup> Input Bias Current	V <sub>CM</sub> = 0VDC	MT	±2.5	±8		±1.3	±4	$CO^{1}$	±0.7	±2		±43	±170	pA
OFFSET CURRENT <sup>(1)</sup> Input Offset Current	V <sub>CM</sub> = 0VDC	OM.	1.1		W	0.6	100	Y.C	0.6			18		pA
OFFSET VOLTAGE <sup>(1)</sup> Input Offset Voltage Average Drift Supply Rejection	V <sub>CM</sub> = 0VDC	74	114 ±2	±2.2mV ±20 ±200	80	114 ±2	±1mV ±10	80	114 ±2	±750 ±5 ±100	80	106 ±5	±1.5mV ±10	μV μV/°C dB μV/V
VOLTAGE RANGE <sup>(2)</sup> Common-Mode Input Range Commmon-Mode Rejection	V <sub>IN</sub> = ±10VDC	±10 74	±11	TW	±10 80	±11	WW	±10 80	±11	$OM_{i}$	±10 74	±11		V dB
OPEN-LOOP GAIN, DC	MM	OOY.		TTV.		1	MAG	-31 10	101.	.01	TV			
Open-Loop Voltage Gain	$R_L \ge 2k\Omega$	90	125		104	125		104	125	Co	90	122		dB
RATED OUTPUT	Wire	Jan	-7 C	Mir	-1		- 11	MW.	10-	J CO	Mr.	XX		
Voltage Output Current Output Short Circuit Current	$R_{L} = 2k \mid V_{O} = \pm 10 \text{VDC}$ $V_{O} = 0 \text{VDC}$	±10 ±5 10	22		±10 ±5 10	22	1	±10 ±5 10	22	N.C	±10 ±5 10	18		V mA mA
POWER SUPPLY	WW	1	001.	Co.	TT		1	MAN	-x11(	101.		TY	N	
Current, Quiescent	I = 0mADC	1111.7	0.9	1.8	120	0.9	1.8	WIN	0.9	1.8	$Co_k$	0.9	2	mA

NOTES: (1) Offset voltage, offset current, and bias current are measured with the units fully warmed up. (2) If it is possible for the input voltage to exceed the supply voltage, a series protection resistor should be added to limit input current to 0.5mA. The input devices can withstand overload currents of 0.3mA indefinitely without damage.

#### **CONNECTION DIAGRAM**



#### **ABSOLUTE MAXIMUM RATINGS**

Supply	±18VDC
Internal Power Dissipation <sup>(1)</sup>	500mW
Differential Input Voltage	±36VDC
Input Voltage Range	±18VDC
Storage Temperature Range	65°C to +150°C
Operating Temperature Range	55°C to +125°C
Lead Temperature (soldering, 10s)	+300°C
Output Short Circuit Duration(2)	Continuous
Junction Temperature	+175°C
NOTES: (1) Packages must be derated based 200°C/W. (2) Short circuit may be to power su	071

applies to +25°C ambient. Observe dissipation limit and  $T_{\rm J}$ .

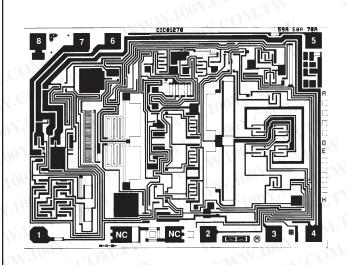
#### ORDERING INFORMATION

PRODUCT	PACKAGE	TEMPERATURE RANGE	BIAS CURRENT, max (fA)
OPA128JM	TO-99	0°C to +70°C	±300
OPA128KM	TO-99	0°C to +70°C	±150
OPA128LM	TO-99	0°C to +70°C	±75
OPA128SM	TO-99	-55°C to +125°C	±150

#### **PACKAGE INFORMATION**

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER <sup>(1)</sup>
OPA128JM	TO-99	001
OPA128KM	TO-99	001
OPA128LM	TO-99	001
OPA128SM	TO-99	001

NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.



**OPA128 DIE TOPOGRAPHY** 

PAD	FUNCTION
1	Offset Trim
2	-In
3	+In
4	-V <sub>cc</sub>
5	Offset Trim
6	Output
7 ~ 0	+V <sub>cc</sub>
8	Substrate
NC	No Connection

Substrate Bias: Isolated, normally connected to common.

#### **MECHANICAL INFORMATION**

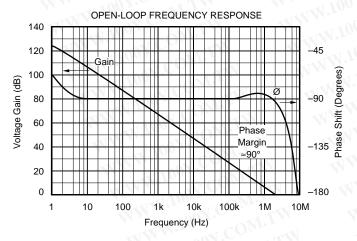
MM,	MILS (0.001")	MILLIMETERS
Die Size	96 x 71 ±5	2.44 x 1.80 ±0.13
Die Thickness	20 ±3	0.51 ±0.08
Min. Pad Size	4 x 4	0.10 x 0.10
Backing	W.100	None

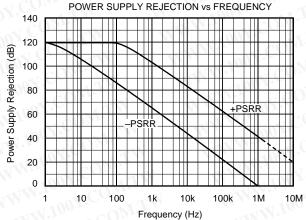
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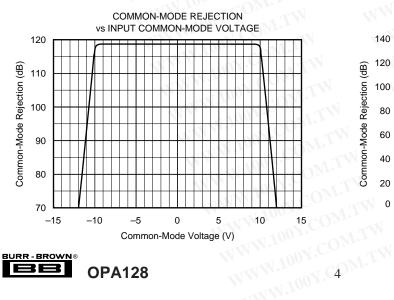
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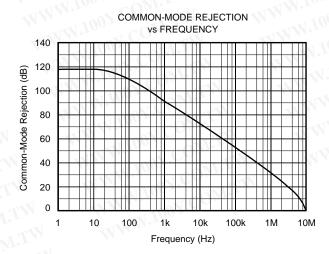
# TYPICAL PERFORMANCE CURVES

At  $T_{\Delta} = +25^{\circ}C$ ,  $\pm 15$ VDC, unless otherwise noted.





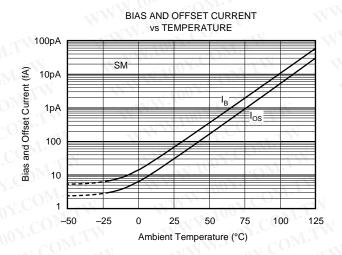


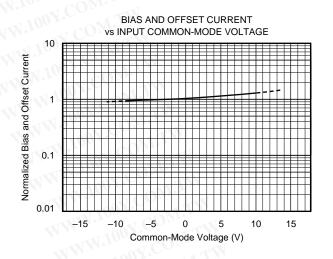


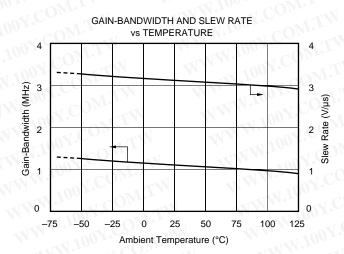
**OPA128** 

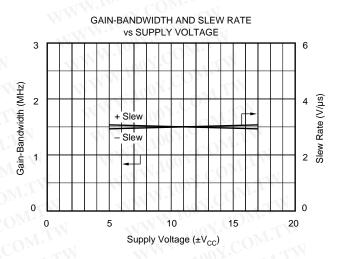
# TYPICAL PERFORMANCE CURVES (CONT)

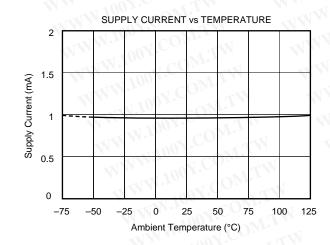
At  $T_{\Delta} = +25^{\circ}C$ , +15VDC, unless otherwise noted.

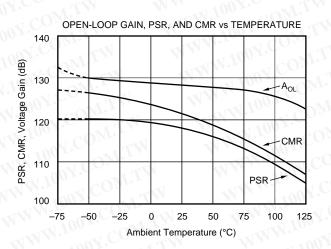










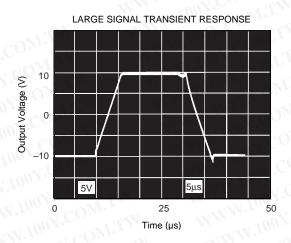


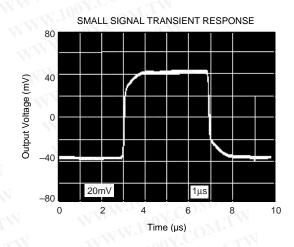
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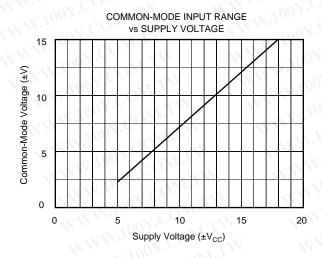
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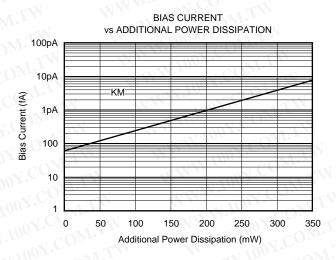
# TYPICAL PERFORMANCE CURVES (CONT)

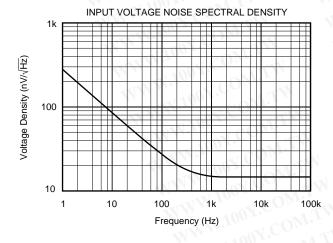
At  $T_{\Delta} = +25^{\circ}C$ , +15VDC, unless otherwise noted.

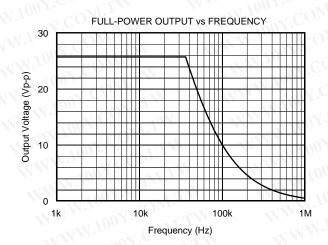












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BURR-BROWN® OPA128

# APPLICATIONS INFORMATION

#### OFFSET VOLTAGE ADJUSTMENT

The OPA128 offset voltage is laser-trimmed and will require no further trim for most applications. As with most amplifiers, externally trimming the remaining offset can change drift performance by about  $0.3\mu V/^{\circ}C$  for each  $100\mu V$  of adjusted effort. Note that the trim (Figure 1) is similar to operational amplifiers such as HA-5180 and AD515. The OPA128 can replace many other amplifiers by leaving the external null circuit unconnected.

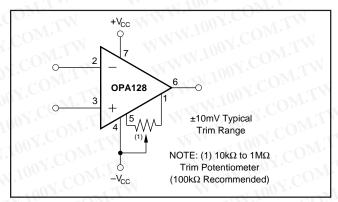


FIGURE 1. Offset Voltage Trim.

#### INPUT PROTECTION

Conventional monolithic FET operational amplifiers' inputs must be protected against destructive currents that can flow when input FET gate-to-substrate isolation diodes are forward-biased. Most BIFET® amplifiers can be destroyed by the loss of  $-V_{\rm CC}$ .

Because of its dielectric isolation, no special protection is needed on the OPA128. Of course, the differential and common-mode voltage limits should be observed.

Static damage can cause subtle changes in amplifier input characteristics without necessarily destroying the device. In precision operational amplifiers (both bipolar and FET types), this may cause a noticeable degradation of offset voltage and drift.

Static protection is recommended when handling any precision IC operational amplifier.

#### **GUARDING AND SHIELDING**

As in any situation where high impedances are involved, careful shielding is required to reduce "hum" pickup in input leads. If large feedback resistors are used, they should also be shielded along with the external input circuitry. Leakage currents across printed circuit boards can easily exceed the bias current of the OPA128. To avoid leakage problems, it is recommended that the signal input lead of the OPA128 be wired to a Teflon standoff. If the input is to be soldered directly into a printed circuit board, utmost care must be used in planning the board layout. A "guard" pattern should completely surround the high impedance input leads and should be connected to a low impedance point which is at the signal input potential.

The amplifier case should be connected to any input shield or guard via pin 8. This insures that the amplifier itself is fully surrounded by guard potential, minimizing both leakage and noise pickup (see Figure 2).

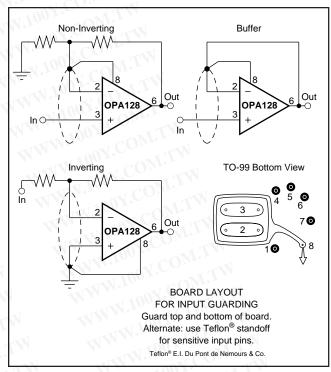


FIGURE 2. Connection of Input Guard.

Triboelectric charge (static electricity generated by friction) can be a troublesome noise source from cables connected to the input of an electrometer amplifier. Special low-noise cable will minimize this effect but the optimum solution is to mount the signal source directly at the electrometer input with short, rigid, wiring to preclude microphonic noise generation.

#### **TESTING**

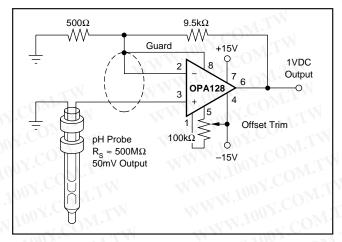
Accurately testing the OPA128 is extremely difficult due to its high level of performance. Ordinary test equipment may not be able to resolve the amplifier's extremely low bias current.

Inaccurate bias current measurements can be due to:

- 1. Test socket leakage
- 2. Unclean package
- 3. Humidity or dew point condensation
- 4. Circuit contamination from fingerprints or anti-static treatment chemicals
- 5. Test ambient temperature
- 6. Load power dissipation

BIFET® National Semiconductor Corp







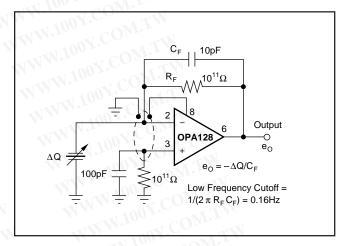


FIGURE 4. Piezoelectric Transducer Charge Amplifier.

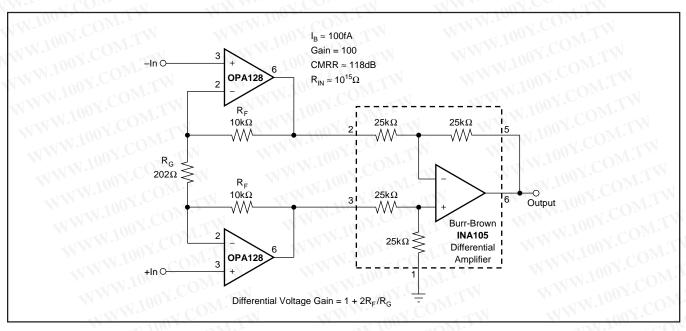


FIGURE 5. FET Input Instrumentation Amplifier for Biomedical Applications.

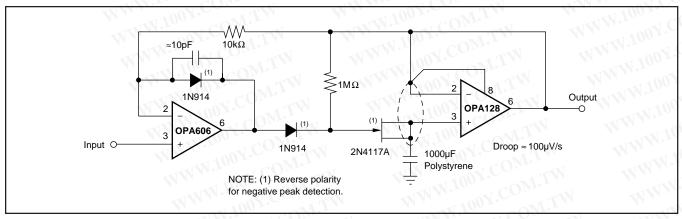
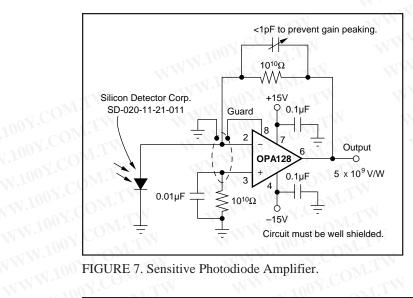


FIGURE 6. Low-Droop Positive Peak Detector.





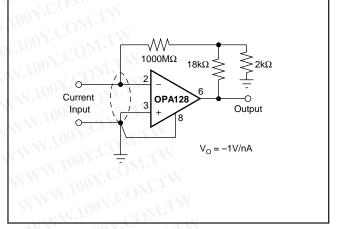


FIGURE 8. Current-to-Voltage Converter.

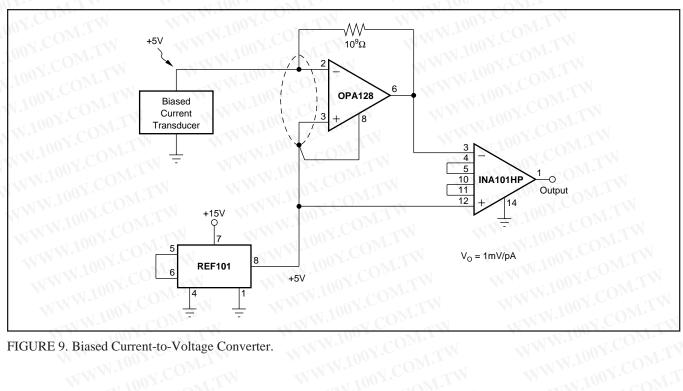


FIGURE 9. Biased Current-to-Voltage Converter. WWW.100Y.COM

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#### PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp <sup>(3)</sup>
OPA128JM	NRND	TO-99	LMC	8	20	TBD	Call TI	Level-NA-NA-NA
OPA128KM	NRND	TO-99	LMC	8	20	TBD	Call TI	Level-NA-NA-NA
OPA128LM	NRND	TO-99	LMC	8	20	TBD	Call TI	Level-NA-NA-NA
OPA128SM	NRND	TO-99	LMC	8	20	TBD	Call TI	Level-NA-NA-NA

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in

a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS) or Green (RoHS & no Sb/Br) - please check <a href="http://www.ti.com/productcontent">http://www.ti.com/productcontent</a> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

**Pb-Free** (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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Mailing Address: Texas Instruments

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