

PA98 • PA98A

FEATURES

- HIGH VOLTAGE 450V (±225V)
- HIGH SLEW RATE 1000V/μS
- HIGH OUTPUT CURRENT 200mA

APPLICATIONS

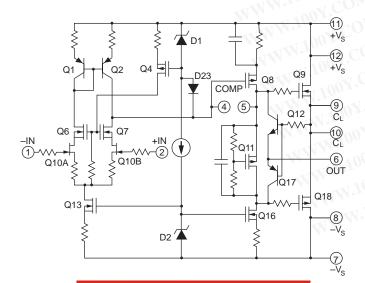
- HIGH VOLTAGE INSTRUMENTATION
- PIEZO TRANSDUCER EXCITATION
- PROGRAMMABLE POWER SUPPLIES UP TO 430V
- ELECTROSTATIC TRANSDUCERS & DEFLECTION

DESCRIPTION

The PA98 is a high voltage, high power bandwidth MOSFET operational amplifier designed for output currents up to 200mA. Output voltages can swing up to ±215V with a dual supply and up to +440 volts with a single supply. The safe operating area (SOA) has no second breakdown limitations and can be observed with all types of loads by choosing an appropriate current limiting resistor. High accuracy is achieved with a cascode input circuit configuration. All internal biasing is referenced to a bootstrapped zener-MOSFET current source. As a result, the PA98 features an unprecedented supply range and excellent supply rejection. The MOSFET output stage is biased on for linear operation. External compensation provides user flexibility.

This hybrid circuit utilizes thick film (cermet) resistors, ceramic capacitors and silicon semiconductor chips to maximize reliability, minimize size and give top performance. Ultrasonically bonded aluminum wires provide reliable interconnections at all operating temperatures. The Power SIP package is electrically isolated.

EQUIVALENT SCHEMATIC



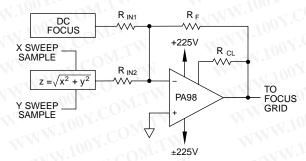
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PATENTED

12-pin SIP PACKAGE STYLE DP

TYPICAL APPLICATION

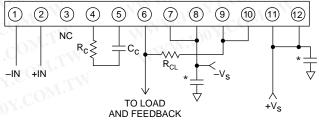


DYNAMIC FOCUSING

Dynamic focusing is the active correction of focusing voltage as a beam traverses the face of a CRT. This is necessary in high resolution flat face monitors since the distance between cathode and screen varies as the beam moves from the center of the screen to the edges. PA98 lends itself well to this function since it can be connected as a summing amplifier with inputs from the nominal focus potential and the dynamic correction. The nominal might be derived from a potentiometer, or perhaps automatic focusing circuitry might be used to generate this potential. The dynamic correction is generated from the sweep voltages by calculating the distance of the beam from the center of the display.

EXTERNAL CONNECTIONS

PACKAGE: SIP03



* Bypassing required.

PHASE COMPENSATION

Gain	C _c	R_c			
1	68pF	100Ω			
20	10pF	330Ω			
100	3.3nF	ΛO			

C_c RATED FOR FULL SUPPLY VOLTAGE

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ABSOLUTE MAXIMUM RATINGS **SPECIFICATIONS**

ABSOLUTE MAXIMUM RATINGS

SUPPLY VOLTAGE, +V $_{\! S}$ to -V $_{\! S}$ OUTPUT CURRENT, continuous within SOA POWER DISSIPATION, continuous @ $T_c = 25^{\circ}C^2$

INPUT VOLTAGE, differential INPUT VOLTAGE, common mode

TEMPERATURE, junction² TEMPERATURE RANGE, storage

OPERATING TEMPERATURE RANGE, case

TEMPERATURE, pin solder - 10s max.

150°C -40 to +85°C

-25 to +85°C

450V

30W

±25V

260°C

±Vs

200mA

SPECIFICATIONS

SPECIFICATIONS		OOY.C	ON.COM			MM. 100X.			
PARAMETER	TEST CONDITIONS 1	MIN	PA98 TYP	MAX	MIN	PA98A TYP	MAX	UNITS	
INPUT	Om MAL	Your	CO	TW		MAN	-110	OXICO	
OFFSET VOLTAGE, initial OFFSET VOLTAGE, vs. temperature OFFSET VOLTAGE, vs. supply OFFSET VOLTAGE, vs. time BIAS CURRENT, initial ³	Full temperature range	N.100 N.100	.5 10 3 75 5	2 30 10	N	.25 5 * *	.5 10 *	mV μV/°C μV/V μV/√kh pA	
DIAS CURRENT, vs. supply OFFSET CURRENT, initial ³ INPUT IMPEDANCE, DC	M.COM.TW W	MW.1	.01 10 10 ¹¹	100	LM M	3	30	pA/V pA Ω	
INPUT CAPACITANCE COMMON MODE VOLTAGE RANGE ⁴	TOOX.COM.TW	±V _S -15	100X	CON	T*W	*		pF V	
COMMON MODE REJECTION, DC NOISE	$V_{CM} = \pm 90V$ 100kHz BW, $R_S = 1K\Omega$, $C_C = 10pf$	90	110	.00	M. T. V	*		dB μVrms	
GAIN	TW. CO. TW	MIN	100	1.00	TIM			-33	
OPEN LOOP, @ 15Hz GAIN BANDWIDTH PRODUCT at 1MHz POWER BANDWIDTH	$\begin{array}{l} R_L = 2K\Omega, \ C_C = OPEN \\ R_L = 2K\Omega, \ C_C = 3.3pf \\ C_C = 10pf \\ C_C = 3.3pf \end{array}$	96	111 100 300 500	07.C	OM.	* *		dB MHz kHz kHz	
PHASE MARGIN	Full temperature range	1	60		$C_{O_{\tilde{M}}}$	*		0	
OUTPUT	W.1001. COM.1		WIX	700 -	· c0	11.1			
VOLTAGE SWING ⁴ VOLTAGE SWING ⁴ VOLTAGE SWING ⁴ CURRENT, continuous SLEW RATE, A _V = 20 SLEW RATE, A _V = 100 CAPACITIVE LOAD, A _V = +1 SETTLING TIME to .1% RESISTANCE, no load	$\begin{split} I_{\text{O}} &= \pm 200\text{mA} \\ I_{\text{O}} &= \pm 75\text{mA} \\ I_{\text{O}} &= \pm 20\text{mA} \\ T_{\text{C}} &= 85^{\circ}\text{C} \\ C_{\text{C}} &= 10\text{pf} \\ C_{\text{C}} &= \text{OPEN} \\ \text{Full temperature range} \\ C_{\text{C}} &= 10\text{pf}, 2\text{V step} \\ R_{\text{CL}} &= 0 \end{split}$	±V-8.5	±Vs-6.5 ±Vs-6.0 ±Vs-5.5 400 1000 1	M.M.T. M.YO. M.YO.	* * * 700	(*) (*)	N PW TW LTW	V V V mA V/μs V/μs pf μs	
POWER SUPPLY	WWW. CON.	WT	-	MAL	- 100	M.Co.			
VOLTAGE ⁶ CURRENT, quiescent	Full temperature range	±15	±150	±225 25	*	***	*	V mA	
THERMAL	MAN M. TOON CO	TY		W	1	OOY.C			
RESISTANCE, AC, junction to case ⁵ RESISTANCE, DC, junction to case RESISTANCE, junction to air TEMPERATURE RANGE, case	Full temperature range, F > 60Hz Full temperature range, F < 60Hz Full temperature range Meets full range specifications	-25	30	2.5 4.2 +85	*	7.100Y	*	°C/W °C/W °C/W	

NOTES: *

- The specification of PA98A is identical to the specification for PA98 in applicable column to the left.
- 1. Unless otherwise noted: $T_C = 25^{\circ}C$, compensation = $C_C = 68$ pF, $R_C = 100\Omega$. DC input specifications are \pm value given. Power supply voltage is typical rating.
- 2. Long term operation at the maximum junction temperature will result in reduced product life. Derate internal power dissipation to achieve high MTTF. Ratings apply only to output transistors. An additional 10W may be dissipated due to quiescent power.
- 3. Doubles for every 10°C of temperature increase.
- 4. +V_s and -V_s denote the positive and negative power supply rail respectively.
- Rating applies if the output current alternates between both output transistors at a rate faster than 60Hz. 5.
- Derate max supply rating .625 V/°C below 25°C case. No derating needed above 25°C case.

CAUTION

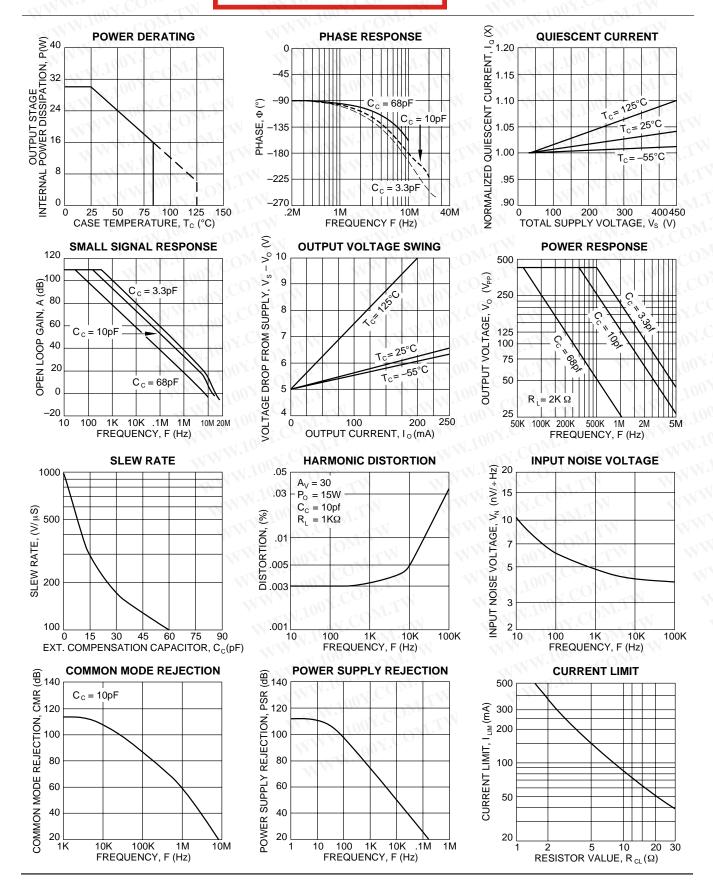
The PA98 is constructed from MOSFET transistors. ESD handling procedures must be observed. The exposed substrate contains beryllia (BeO). Do not crush, machine, or subject to temperatures in excess of 850°C to avoid generating toxic fumes.

TYPICAL PERFORMANCE GRAPHS

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OPERATING CONSIDERATIONS

GENERAL

Please read Application Note 1 "General Operating Considerations" which covers stability, supplies, heat sinking, mounting, current limit, SOA interpretation, and specification interpretation. Visit www.apexmicrotech.com for design tools that help automate tasks such as calculations for stability, internal power dissipation, current limit; heat sink selection; Apex's complete Application Notes library; Technical Seminar Workbook; and Evaluation Kits.

CURRENT LIMIT

For proper operation, the current limit resistor (R_{CL}) must be connected as shown in the external connection diagram. The minimum value is 1.4 ohm, however for optimum reliability the resistor value should be set as high as possible. The value is calculated as follows; with the maximum practical value of 30 ohms

$$R_{CL} = \frac{.7}{I_{LIM} - .016}$$

SAFE OPERATING AREA (SOA)

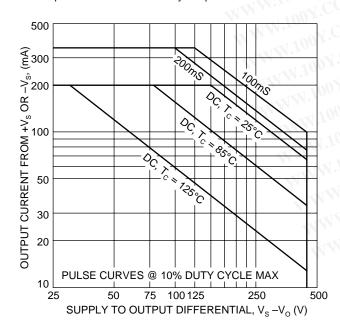
The MOSFET output stage of this power operational amplifier has two distinct limitations:

- 1. The current handling capability of the MOSFET geometry and the wire bonds.
- 2. The junction temperature of the output MOSFETs.

NOTE: The output stage is protected against transient flyback. However, for protection against sustained, high energy flyback, external fast-recovery diodes should be used.

SAFE OPERATING CURVES

The safe operating area curves define the maximum additional internal power dissipation the amplifier can tolerate when it produces the necessary output to drive an external



load. This is not the same as the absolute maximum internal power dissipation listed elsewhere in the specification since the quiescent power dissipation is significant compared to the

INPUT PROTECTION

Although the PA98 can withstand differential voltages up to ±25V, additional external protection is recommended. Since the PA98 is a high speed amplifier, low leakage, low capacitance JFETs connected as diodes are recommended (e.g. 2N4416, Q1-Q4 in Figure 1). The differential input voltage will be clamped to ±1.4V. This is sufficient overdrive to produce maximum power bandwidth.

POWER SUPPLY PROTECTION

Unidirectional zener diode transient suppressors are recommended as protection on the supply pins. The zeners clamp transients to voltages within the power supply rating and also clamp power supply reversals to ground. Whether the zeners are used or not, the system power supply should be evaluated for transient performance including power-on overshoot and power-off polarity reversals as well as line regulation.

Conditions which can cause open circuits or polarity reversals on either power supply rail should be avoided or protected against. Reversals or opens on the negative supply rail is known to induce input stage failure. Unidirectional transzorbs prevent this, and it is desirable that they be both electrically and physically as close to the amplifier as possible.

STABILITY

The PA98 is externally compensated and performance can be tailored to the application. Use the graphs of small signal response and power response as a guide. The compensation capacitor C_c must be rated at 500V working voltage. An NPO capacitor is recommended. The compensation network C_cR_c must be mounted closely to the amplifier pins 7 and 8 to avoid spurious oscillation.

