

### **TDA7396**

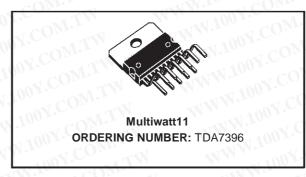
# 45W/2Ω BRIDGE CAR RADIO AMPLIFIER WITH DIAGNOSTIC FACILITY

- HIGH POWER CAPABILITY:
  - 65W/2Ω MAX
  - 60W/2Ω EIAJ
  - $-45W/2\Omega$  @ Vs = 14.4V, f = 1KHz, d = 10%
- DIFFERENTIAL INPUTS (EITHER SINGLE ENDED OR DIFFERENTIAL INPUT SIGNALS ARE ACCEPTED)
- MINIMUM EXTERNAL COMPONENT COUNT:
  - NO BOOTSTRAP CAPACITORS
  - NO BOUCHEROT CELLS
  - INTERNALLY FIXED GAIN (26dB)
  - NO SVR CAPACITOR
- ST.-BY FUNCTION (CMOS COMPATIBLE)
- AUTOMATIC MUTE DURING TURN-ON/OFF
- AUTOMUTE AT MINIMUM SUPPLY VOLTAGE DETECTION
- SYNCHRONIZING PIN FOR SIMULTANEOUS TURN-ON IN MULTI-DEVICE APPLICATIONS
- NO AUDIBLE POP DURING MUTE AND ST-BY OPERATIONS

### **Diagnostic Facilities:**

■ CLIPPING DETECTOR

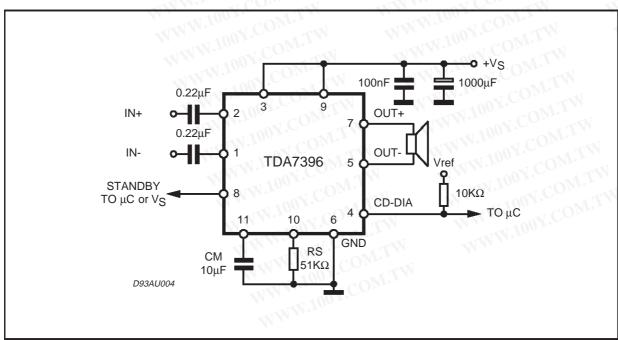
Figure 1: Test And Application Circuit



- SHORT CIRCUIT
- OPEN LOAD
- THERMAL SHUTDOWN

#### PROTECTIONS:

- SHORT CIRCUIT (TO GND, TO V<sub>S</sub>, ACROSS THE LOAD)
- VERY INDUCTIVE LOADS
- CHIP OVER-TEMPERATURE
- LOAD DUMP
- OPEN GND
- ESD



October 1998 1/11

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### DESCRIPTION

The TDA7396 is a BRIDGE class AB audio power amplifier especially intended for car radio High Power applications.

The  $2\Omega$  power capability together with the possibility to operate either in DIFFERENTIAL INPUT MODE or SINGLE ENDED INPUT MODE makes it suitable for boosters and high end car radio ap-

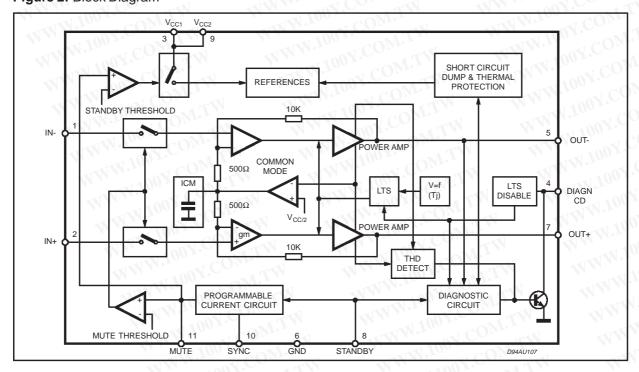
Figure 2: Block Diagram

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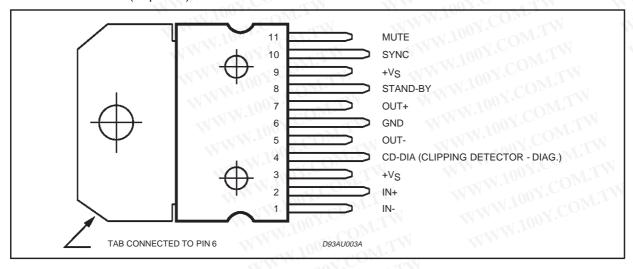
. The exclusive fully complementary output stage and the internally fixed gain configuration drop the external component count.

The on board clipping detector allows easy implementation of gain compression systems.

The diagnostics facility allows to detect any mistakes during car radio set assembly and wiring in the car.



### PIN CONNECTION (Top view)



#### THERMAL DATA

Symbol	Description		Value	Unit
R <sub>th j-case</sub>	Thermal Resistance Junction-case	Max	1.5	°C/W

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Symbol	Parameter	Value	Unit
Vs	DC Supply Voltage	28	V
V <sub>OP</sub>	Operating Supply Voltage	18	V
$V_{PEAK}$	Peak Supply Voltage (t = 50ms)	50	V
lo 10	Output Peak Current repetitive (f > 10Hz) Output Peak Current non repetitive	6 7	AA
P <sub>tot</sub>	Power Dissipation (T <sub>CASE</sub> = 85°C)	43	W
T <sub>stg</sub> , T <sub>j</sub>	Storage and Junction-Case	-40 to 150	°C

### WW.100Y.COM.T **ELECTRICAL CHARACTERISTICS** ( $V_S = 14.4V$ ; $R_L = 2\Omega$ , f = 1KHz, $T_{amb} = 25$ °C, unless otherwise specified) specified)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Uni
Vs	Supply Voltage Range	11 100 TOM	8		18	V
I <sub>q</sub>	Total Quiescent Current	M.M. 100X.C	TW	100	N 4	m/
Vos	Output Offset Voltage	MAN. OON.CO.	WT		150	m۷
I <sub>SB</sub>	ST-BY Current	Vsr-BY = 1.5V	VI.		100	μΑ
I <sub>SBin</sub>	ST-BY Input Bias Current	$V_{ST-BY} = 5V$	Mir	- 7	10	μΑ
$V_{SBon}$	ST-BY On Threshold Voltage	MW. 100X.C.	The	N	1.5	V
$V_{SBoff}$	ST-BY Off threshold Voltage	N NAMA CON C	3.5	W	W	V
ATT <sub>ST-BY</sub>	ST-BY Attenuation	IN TOO	$CO_{MT}$ .	90		dB
R <sub>EXT</sub> /R <sub>S</sub> Mute Resistor Ratio for External Mute Control		(see Application Circuit of fig. 4)	0.63	T.M.	0.69	V
I <sub>M in</sub> Mute Input Bias Current		V <sub>MUTE</sub> = 5V	47 CO	1. 1	10	μΑ
$A_M$	Mute Attenuation	1.TW W. 100	1.0	90		dB
Po	RMS Output Power	$ d = 10\% \\ d = 1\% \\ d = 10\%; R_L = 4\Omega $	OY.C	45 35 27		W W W
	EIAJ Output Power	V <sub>S</sub> = 13.7V	. Voo.	27 60 0.06		W
d	Distortion	$P_{O} = 0.1 \text{ to } 20W$ $P_{O} = 0.1 \text{ to } 15W; R_{L} = 4\Omega$	1007	0.06 0.03	WI	% %
G <sub>V</sub>	Voltage Gain	COMP	25	26	27	dB
f <sub>H</sub>	High Frequency rolloff	$P_0 = 1W; -3dB$	75	<1 CO	Mr.	KH:
R <sub>IN</sub>	Input Impedance	Differential	-TXV.10	60	$M_{II}$	KΩ
	MM.	Single Ended	-11	55	TIME	KΩ
E <sub>IN</sub>	Input Noise Voltage	$R_g = 0\Omega$ ; $f = 22Hz$ to $22KHz$	MM	47.	OF	m۱
CMRR	Input Common Mode Rejection	f = 1KHz; V <sub>IN</sub> = 1Vrms	WIN	70	$CO_{M_I}$ .	dB
SVR	Supply Voltage Rejection	$R_g = 0\Omega$ ; $V_r = 1 Vrms$	1	60	COM	dB
CDL	Clipping Detection Level	TW.Co.	MM	0.5	1.00	%
$T_{sd}$	Absolute Thermal Shutdown Junction Temperature	1.100Y.COMI.TW	WW	160	M.CO	dE

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### **FUNCTIONAL DESCRIPTION**

Pin	Function	Description
1, 2	INPUTS	The input stage is a high impedance differential type also capable of operation in single ended mode with one input capacitively coupled to the signal GND. the impedance seen by the inverting and non inverting input pins must be matched.
4	CD - DIA	The TDA7396 is equipped with a diagnostics circuitry able to detect the following events:  - Clipping in the Output Signal  - Thermal Shutdown  - Open Load (before turn-on)  - Shorted Output: to GND, to Vs, across the load (after turn-on)  The CD-DIA(open collector) pin gives out the diagnostics signal (low during clipping or output fault condition).  The device does not work as long as the faulty condition holds; the normal operation is automatically restored after the fault removal.
, 7	OUTPUTS	The output stage is a bridge type able to drive loads as high as 2Ω. It consists of two class AB fully complementary PNP/NPN stages fully protected. A rail to rail output voltage swing is achieved with no need of bootstrap capacitors. No external compensation is necessary.
8	STAND-BY	The device features a ST-BY function which shuts down all the internal bias supplies when the ST-BY input is low.  In ST-BY mode the amplifier sinks a small current (in the range of few μA).  When the ST-BY pin is high the IC becomes fully operational.
10	SYNC	A resistor ( $R_S$ ) has to be connect between pin 10 and GND in order to program the current that flows into $C_M$ capacitor (pin 11). The values of $C_M$ and $R_S$ determine the time required to bias the amplifier.
11	MUTE	a) The pin will have a capacitor ( $C_M$ ) tied to GND to set the MUTE/STAND-BY time. An automatic Mute during turn on/off is provided to prevent noisy transients b) If a independent Mute function is needed, an external transistor circuit (see fig. 4) may be connected to this pin; 1% precision resistors have to be used for $R_{EXT}/R_S$ in to order to reach the fixed limits $0.63 \le R_{EXT}/R_S \le 0.69$

Figure 3: Application Circuit with External Mute Control

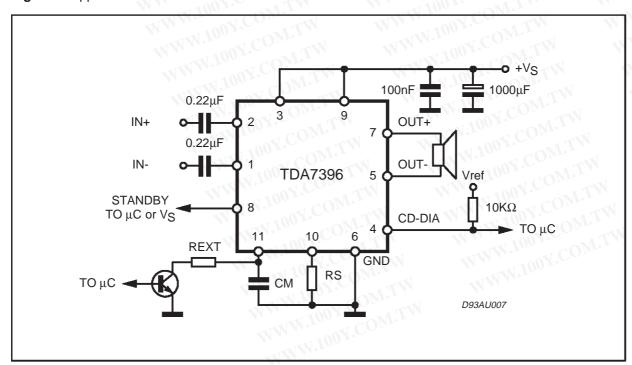
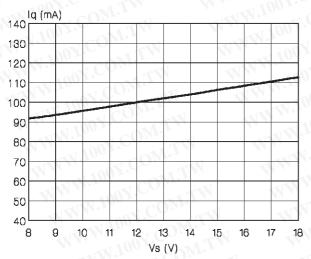


Figure 4: Quiescent Current vs Supply Voltage

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**Figure 6:** Output Power vs Supply Voltage ( $R_L = 2\Omega$ )

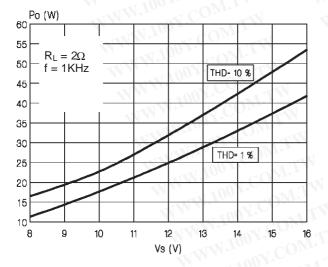


Figure 8: Output Power vs Supply Voltage ( $R_L = 4\Omega$ )

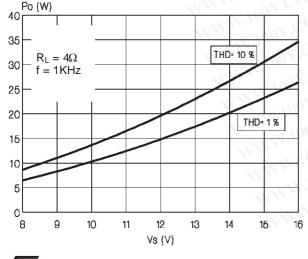
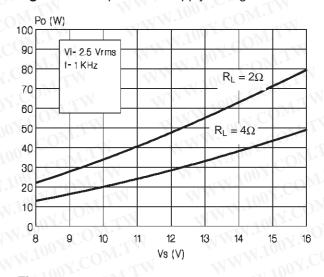


Figure 5: EIAJ power vs Supply Voltage



**Figure 7:** Distortion vs Frequency ( $R_L = 2\Omega$ )

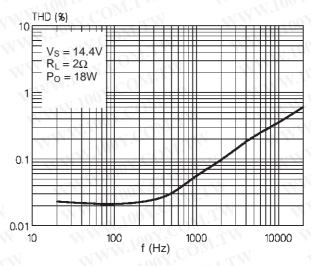
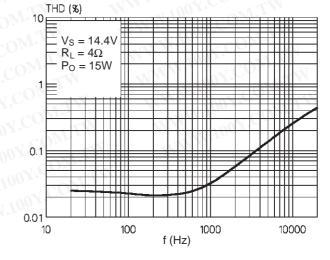
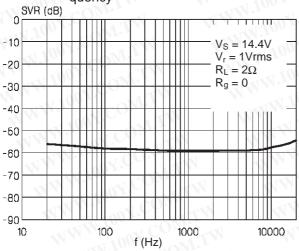


Figure 9: Distortion vs Frequency ( $R_L = 4\Omega$ )



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Figure 10: Supply Voltage Rejection vs Frequency



**Figure 12:** Total Power Dissipation & Efficiency vs. Output Power ( $R_L = 2\Omega$ )

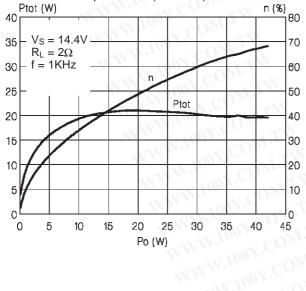


Figure 11: Common Mode Rejection vs. Frequency

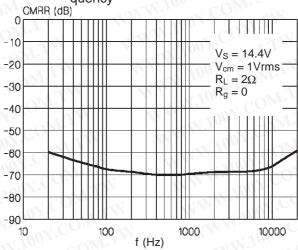
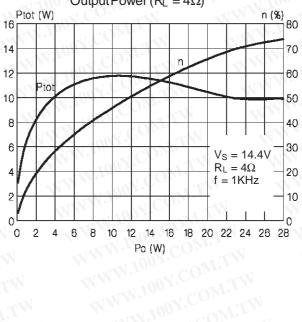


Figure 13: Total Power Dissipation & Efficiency vs. Output Power ( $R_L = 4\Omega$ )



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Figure 14: Application Circuit

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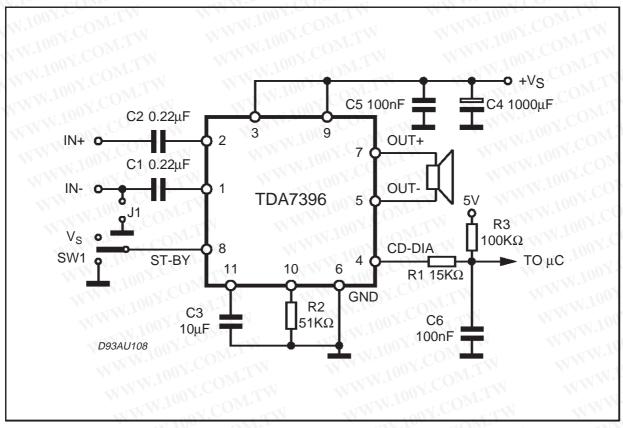
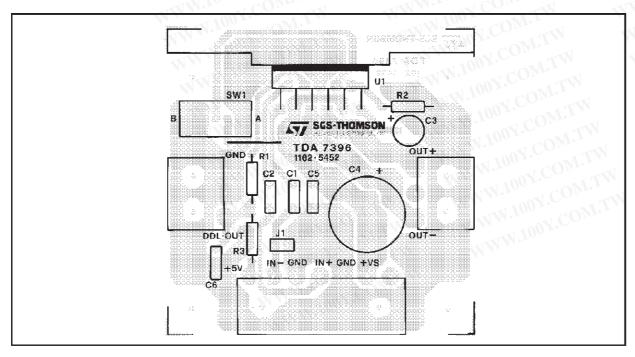


Figure 15: P.C. Board and Component Layout (1:1 scale)



### **CLIPPING DETECTION & DIAGNOSTICS** (see fig. 16)

An active pull-down circuit is provided to signal out the occurrence of any of the conditions described later. In order to use this function, the CD-DIA pin (#4) has to be resistively connected to a positive voltage reference (between 5V and Vs).

### A) Clipping Detection

Current is sunk whenever the output clipping distortion level reaches a fixed 0.5% threshold; this function allows gain compression facility when the amplifier is overdriven.

### **B) Output Fault Diagnostics**

Current is sunk as soon as one of the following output faults is recognized:

- short-circuit to GND
- short-circuit to Vs
- short-circuit across the load (after turn-on)
- open load (before turn-on)

The diagnostics signal remains steadily on until the fault is removed.

All the output fault conditions (listed above) can be distinguished from the clipping detection (A) because of their different time duration.

### THERMAL WARNING (see fig. 17)

Thermal protection has been implemented in accordance to a new principle involving different steps:

- 1) Thermal foldback (Linear Thermal Shutdown)
- Shutdown with soft restart (Absolute Thermal Shutdown)

As long as the junction temperature remains below the preset threshold (140°C) the IC delivers the full power. Once the threshold has been reached, a thermal foldback starts limiting the output signal level; the output power is then reduced, thus decreasing also the temperature without output signal interruption (LTS). Supposing the thermal foldback does not reduce the junction temperature to a safe level, a complete shutdown will occur at 160°C (ATS).

Soft restart avoids large voltage disturbance across the loudspeaker, due to the presence of high input signals when the IC comes out of thermal shutdown.

Current is sunk from the CD-DIA pin (#4)when the thermal protection is acting. The voltage at pin #4 bounces back and forth (depending on the amplifier input signal magnitude) during the linear thermal intervention (LTS) and stays low (sinking current) after the amplifier has been eventually shut-down (Absolute Thermal Shutdown)

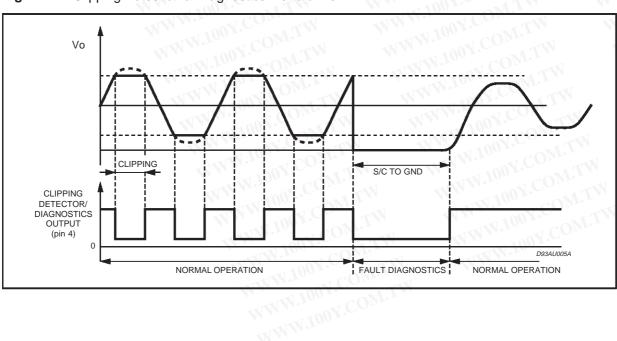


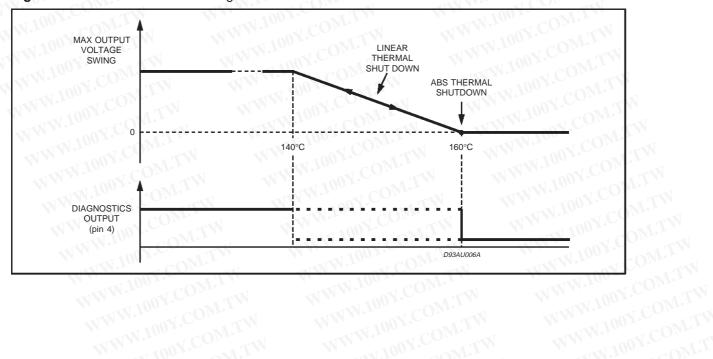
Figure 16: Clipping Detection & Diagnostics Waveforms

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Figure 17: Thermal Protection & Diagnostic Waveforms

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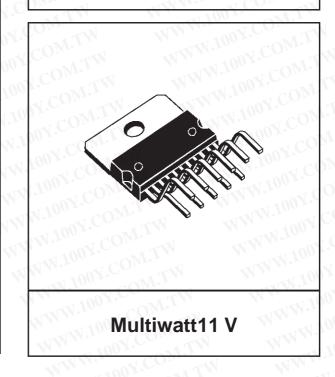


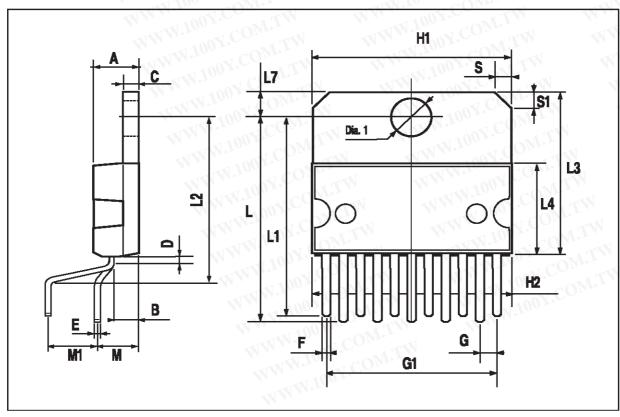
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### **TDA7396**

DIM.		mm	<b>≪</b> Ĭ	inch		
AT 10	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
Α	N.C.	1	5		144	0.197
В		$O_{Mr}$	2.65		NW	0.104
С	100 =	COM	1.6		TIN	0.063
D	1007	1	TIM		0.039	TXI.
E	0.49	CO	0.55	0.019	W	0.022
F	0.88	J CO	0.95	0.035	N.	0.037
G	1.45	1.7	1.95	0.057	0.067	0.077
G1	16.75	17	17.25	0.659	0.669	0.679
H1	19.6	ooy.	V	0.772		MAN
H2	WW.	<b>V</b>	20.2			0.795
L	21.9	22.2	22.5	0.862	0.874	0.886
L1	21.7	22.1	22.5	0.854	0.87	0.886
L2	17.4	14.5	18.1	0.685	N	0.713
L3	17.25	17.5	17.75	0.679	0.689	0.699
L4	10.3	10.7	10.9	0.406	0.421	0.429
L7	2.65		2.9	0.104	IM	0.114
М	4.25	4.55	4.85	0.167	0.179	0.191
M1	4.73	5.08	5.43	0.186	0.200	0.214
S	1.9	WY T	2.6	0.075	$M_{IJ}$	0.102
S1	1.9	MA	2.6	0.075	Time	0.102
Dia1	3.65	TV	3.85	0.144	Ob.	0.152

## OUTLINE AND MECHANICAL DATA





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