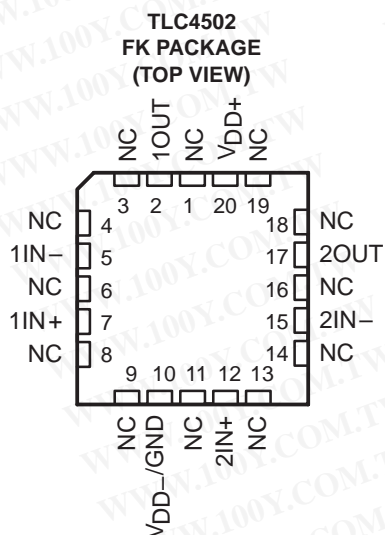
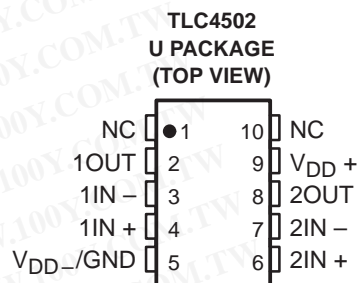
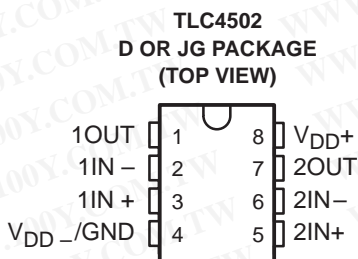
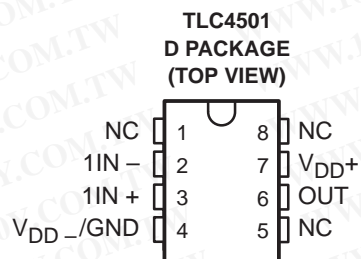


TLC4501, TLC4501A, TLC4502, TLC4502A FAMILY OF SELF-CALIBRATING (Self-Cal™) PRECISION CMOS RAIL-TO-RAIL OUTPUT OPERATIONAL AMPLIFIERS

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description (continued)

Using this technology eliminates the need for noisy and expensive chopper techniques, laser trimming, and power hungry, split supply bipolar operational amplifiers.



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NC – No internal connection

AVAILABLE OPTIONS

T _A	V _{IO} max AT 25°C	PACKAGED DEVICES			
		SMALL OUTLINE† (D)	CHIP CARRIER (FK)	CERAMIC DIP (JG)	CERAMIC FLAT PACK (U)
0°C to 70°C	40 μV	TLC4501ACD	—	—	—
	50 μV	TLC4502ACD	—	—	—
	80 μV	TLC4501CD	—	—	—
	100 μV	TLC4502CD	—	—	—
-40°C to 125°C	40 μV	TLC4501AID	—	—	—
	50 μV	TLC4502AID	—	—	—
	80 μV	TLC4501ID	—	—	—
	100 μV	TLC4502ID	—	—	—
-40°C to 125°C	50 μV	TLC4502AQD	—	—	—
	100 μV	TLC4502QD	—	—	—
-55°C to 125°C	50 μV	TLC4502AMD	TLC4502AMFKB	TLC4502AMJGB	TLC4502AMUB
	100 μV	TLC4502MD	TLC4502MFKB	TLC4502MJGB	TLC4502MUB

† The D package is also available taped and reeled.



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absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage, V_{DD+} (see Note 1)	7 V
Differential input voltage, V_{ID} (see Note 2)	± 7 V
Input voltage range, V_I (any input, see Note 1)	-0.3 V to 7 V
Input current, I_I (each input)	± 5 mA
Output current, I_O (each output)	± 100 mA
Total current into V_{DD+}	± 100 mA
Total current out of V_{DD-}/GND	± 100 mA
Electrostatic discharge (ESD)	> 2 kV
Duration of short-circuit current at (or below) 25°C (see Note 3)	unlimited
Continuous total power dissipation	See Dissipation Rating Table
Operating free-air temperature range, T_A : TLC4502C	0°C to 70°C
TLC4502I	-40°C to 125°C
TLC4502Q	-40°C to 125°C
TLC4502M	-55°C to 125°C
Storage temperature range, T_{stg}	-65°C to 150°C
Case temperature for 60 seconds, T_C : FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

† Stresses beyond those listed under “absolute maximum ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under “recommended operating conditions” is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

- NOTES: 1. All voltage values, except differential voltages, are with respect to V_{DD-}/GND .
 2. Differential voltages are at $IN+$ with respect to $IN-$. Excessive current flows when an input is brought below $V_{DD-} - 0.3$ V.
 3. The output may be shorted to either supply. Temperature and/or supply voltages must be limited to ensure that the maximum dissipation rating is not exceeded.

DISSIPATION RATING TABLE

PACKAGE	$T_A \leq 25^\circ\text{C}$ POWER RATING	DERATING FACTOR ABOVE $T_A = 25^\circ\text{C}$	$T_A = 70^\circ\text{C}$ POWER RATING	$T_A = 85^\circ\text{C}$ POWER RATING	$T_A = 125^\circ\text{C}$ POWER RATING
D	725 mW	5.8 mW/°C	464 mW	377 mW	145 mW
FK	1375 mW	11.0 mW/°C	880 mW	715 mW	275 mW
JG	1050 mW	8.4 mW/°C	672 mW	546 mW	210 mW
U	675 mW	5.4 mW/°C	432 mW	350 mW	135 mW

recommended operating conditions

	TLC4502C		TLC4502I		TLC4502Q		TLC4502M		UNIT
	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	
Supply voltage, V_{DD}	4	6	4	6	4	6	4	6	V
Input voltage range, V_I	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V
Common-mode input voltage, V_{IC}	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V_{DD-}	$V_{DD+} - 2.3$	V
Operating free-air temperature, T_A	0	70	-40	125	-40	125	-55	125	°C

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electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$, $GND = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	TLC450xC			UNIT	
			MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	Full range	TLC4501	-80	10	80	μV
			TLC4501A	-40	10	40	
			TLC4502	-100	10	100	
			TLC4502A	-50	10	50	
αV_{IO} Temperature coefficient of input offset voltage		Full range		1		$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	25°C		1		pA	
		Full range			500		
I_{IB} Input bias current		25°C		1		pA	
		Full range			500		
V_{OH} High-level output voltage	$I_{OH} = -500\ \mu\text{A}$ $I_{OH} = -5\text{ mA}$	25°C		4.99		V	
		25°C		4.9			
		Full range		4.7			
V_{OL} Low-level output voltage	$V_{IC} = 2.5\text{ V}$, $I_{OL} = 500\ \mu\text{A}$ $V_{IC} = 2.5\text{ V}$, $I_{OL} = 5\text{ mA}$	25°C		0.01		V	
		25°C		0.1			
		Full range			0.3		
A_{VD} Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$, $V_O = 1\text{ V to }4\text{ V}$, $R_L = 1\text{ k}\Omega$, See Note 4	25°C	200	1000		V/mV	
		Full range	200				
$R_{I(D)}$ Differential input resistance		25°C		10		k Ω	
R_L Input resistance	See Note 4	25°C		10 ¹²		Ω	
C_L Common-mode input capacitance	f = 10 kHz, P package	25°C		8		pF	
Z_O Closed-loop output impedance	$A_V = 10$, f = 100 kHz	25°C		1		Ω	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$, $V_O = 2.5\text{ V}$, $R_S = 1\text{ k}\Omega$	25°C	90	100		dB	
		Full range	85				
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4\text{ V to }6\text{ V}$, $V_{IC} = 0$, No load	25°C	90	100		dB	
		Full range	90				
I_{DD} Supply current	$V_O = 2.5\text{ V}$, No load	TLC4501/A	25°C	1	1.5	mA	
			Full range		2		
		TLC4502/A	25°C	2.5	3.5		
			Full range		4		
$V_{IT(CAL)}$ Calibration input threshold voltage		Full range	4			V	

† Full range is 0°C to 70°C.

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A^\dagger	TLC450xC, TLC450xAC			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, C_L = 100\text{ pF}$	25°C	1.5	2.5		$\text{V}/\mu\text{s}$
		Full range	1			$\text{V}/\mu\text{s}$
V_n Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C		70		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	25°C		12		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$	25°C		1		μV
	$f = 0.1\text{ to }10\text{ Hz}$	25°C		1.5		
I_n Equivalent input noise current		25°C		0.6		$\text{fA}/\sqrt{\text{Hz}}$
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}, f = 10\text{ kHz}, R_L = 1\text{ k}\Omega, C_L = 100\text{ pF}$	$A_V = 1$	25°C	0.02%		
		$A_V = 10$	25°C	0.08%		
		$A_V = 100$	25°C	0.55%		
Gain-bandwidth product	$f = 10\text{ kHz}, C_L = 100\text{ pF}$	$R_L = 1\text{ k}\Omega,$	25°C	4.7		MHz
B_{OM} Maximum output swing bandwidth	$V_{O(PP)} = 2\text{ V}, R_L = 1\text{ k}\Omega,$	$A_V = 1, C_L = 100\text{ pF}$	25°C	1		MHz
t_s Settling time	$A_V = -1, \text{ Step} = 0.5\text{ V to }2.5\text{ V}, R_L = 1\text{ k}\Omega, C_L = 100\text{ pF}$	to 0.1%	25°C	1.6		μs
		to 0.01%	25°C	2.2		
ϕ_m Phase margin at unity gain	$R_L = 1\text{ k}\Omega,$	$C_L = 100\text{ pF}$	25°C	74		
Calibration time			25°C	300		ms

† Full range is 0°C to 70°C.

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$, $GND = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	TLC450xI			UNIT	
			MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	Full range	TLC4501	-80	10	80	μV
			TLC4501A	-40	10	40	
			TLC4502	-100	10	100	
			TLC4502A	-50	10	50	
α_{VIO} Temperature coefficient of input offset voltage		Full range		1		$\mu\text{V}/^\circ\text{C}$	
I_{IO} Input offset current	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	25°C		1		pA	
		-40°C to 85°C			500		
		Full range			5	nA	
I_{IB} Input bias current	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	25°C		1		pA	
		-40°C to 85°C			500		
		Full range			10	nA	
V_{OH} High-level output voltage	$I_{OH} = -500\ \mu\text{A}$	25°C		4.99		V	
	$I_{OH} = -5\text{ mA}$	25°C		4.9			
		Full range		4.7			
V_{OL} Low-level output voltage	$V_{IC} = 2.5\text{ V}$, $I_{OL} = 500\ \mu\text{A}$	25°C		0.01		V	
	$V_{IC} = 2.5\text{ V}$, $I_{OL} = 5\text{ mA}$	25°C		0.1			
		Full range		0.3			
A_{VD} Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$, $V_O = 1\text{ V to }4\text{ V}$, $R_L = 1\text{ k}\Omega$, See Note 4	25°C	200	1000		V/mV	
		Full range	200				
$R_{I(D)}$ Differential input resistance		25°C		10		$\text{k}\Omega$	
R_L Input resistance	See Note 4	25°C		10^{12}		Ω	
C_L Common-mode input capacitance	$f = 10\text{ kHz}$, P package	25°C		8		pF	
Z_O Closed-loop output impedance	$A_V = 10$, $f = 100\text{ kHz}$	25°C		1		Ω	
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$, $V_O = 2.5\text{ V}$, $R_S = 1\text{ k}\Omega$	25°C	90	100		dB	
		Full range	85				
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4\text{ V to }6\text{ V}$, $V_{IC} = 0$, No load	25°C	90	100		dB	
		Full range	90				
I_{DD} Supply current	$V_O = 2.5\text{ V}$, No load	TLC4501/A	25°C	1	1.5	mA	
			Full range		2		
		TLC4502/A	25°C	2.5	3.5		
			Full range		4		
$V_{IT(CAL)}$ Calibration input threshold voltage		Full range	4			V	

† Full range is -40°C to 125°C.

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A †	TLC450xI, TLC450xAI			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}, C_L = 100\text{ pF}$	25°C	1.5	2.5		$\text{V}/\mu\text{s}$
		Full range	1			$\text{V}/\mu\text{s}$
V_n Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C		70		$\text{nV}/\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	25°C		12		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$	25°C		1		μV
	$f = 0.1\text{ to }10\text{ Hz}$	25°C		1.5		
I_n Equivalent input noise current		25°C		0.6		$\text{fA}/\sqrt{\text{Hz}}$
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}, f = 10\text{ kHz}, R_L = 1\text{ k}\Omega, C_L = 100\text{ pF}$	$A_V = 1$	25°C	0.02%		
		$A_V = 10$	25°C	0.08%		
		$A_V = 100$	25°C	0.55%		
Gain-bandwidth product	$f = 10\text{ kHz}, C_L = 100\text{ pF}$	$R_L = 1\text{ k}\Omega,$	25°C	4.7		MHz
B_{OM} Maximum output swing bandwidth	$V_{O(PP)} = 2\text{ V}, R_L = 1\text{ k}\Omega,$	$A_V = 1, C_L = 100\text{ pF}$	25°C	1		MHz
t_s Settling time	$A_V = -1, \text{ Step} = 0.5\text{ V to }2.5\text{ V}, R_L = 1\text{ k}\Omega, C_L = 100\text{ pF}$	to 0.1%	25°C	1.6		μs
		to 0.01%	25°C	2.2		
ϕ_m Phase margin at unity gain	$R_L = 1\text{ k}\Omega,$	$C_L = 100\text{ pF}$	25°C	74		
Calibration time			25°C	300		ms

† Full range is -40°C to 125°C .

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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electrical characteristics at specified free-air temperature, $V_{DD} = 5\text{ V}$, $GND = 0$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T_A †	TLC4502Q, TLC4502M			UNIT	
			MIN	TYP	MAX		
V_{IO} Input offset voltage	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$	TLC4502 TLC4502A	Full range	-100 -50	10 10	100 50	μV
α_{VIO} Temperature coefficient of input offset voltage			Full range		1		$\mu\text{V}/^\circ\text{C}$
I_{IO} Input offset current	$V_{DD} = \pm 2.5\text{ V}$, $V_O = 0$, $V_{IC} = 0$, $R_S = 50\ \Omega$		25°C		1		nA
I_{IB} Input bias current			125°C			5	nA
			25°C		1		nA
			125°C			10	nA
V_{OH} High-level output voltage	$I_{OH} = -500\ \mu\text{A}$		25°C		4.99		V
	$I_{OH} = -5\text{ mA}$		25°C		4.9		V
			Full range		4.7		V
V_{OL} Low-level output voltage	$V_{IC} = 2.5\text{ V}$, $I_{OL} = 500\ \mu\text{A}$		25°C		0.01		V
	$V_{IC} = 2.5\text{ V}$, $I_{OL} = 5\text{ mA}$		25°C		0.1		V
			Full range			0.3	V
A_{VD} Large-signal differential voltage amplification	$V_{IC} = 2.5\text{ V}$, $V_O = 1\text{ V to }4\text{ V}$, $R_L = 1\text{ k}\Omega$, See Note 4		25°C	200	1000		V/mV
			Full range	200			V/mV
$R_{I(D)}$ Differential input resistance			25°C		10		k Ω
R_L Input resistance	See Note 4		25°C		10 ¹²		Ω
C_L Common-mode input capacitance	$f = 10\text{ kHz}$, P package		25°C		8		pF
Z_O Closed-loop output impedance	$A_V = 10$, $f = 100\text{ kHz}$		25°C		1		Ω
CMRR Common-mode rejection ratio	$V_{IC} = 0\text{ to }2.7\text{ V}$, $V_O = 2.5\text{ V}$, $R_S = 1\text{ k}\Omega$		25°C	90	100		dB
			Full range	85			dB
k_{SVR} Supply-voltage rejection ratio ($\Delta V_{DD} \pm / \Delta V_{IO}$)	$V_{DD} = 4\text{ V to }6\text{ V}$, $V_{IC} = V_{DD}/2$, No load		25°C	90	100		dB
			Full range	90			dB
I_{DD} Supply current	$V_O = 2.5\text{ V}$, No load		25°C		2.5	3.5	mA
			Full range			4	mA
$V_{IT(CAL)}$ Calibration input threshold voltage			Full range	4			V

† Full range is -40°C to 125°C for Q suffix, -55°C to 125°C for M suffix.

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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operating characteristics, $V_{DD} = 5\text{ V}$

PARAMETER	TEST CONDITIONS	T_A †	TLC4502Q, TLC4502M, TLC4502AQ, TLC4502AM			UNIT
			MIN	TYP	MAX	
SR Slew rate at unity gain	$V_O = 0.5\text{ V to }2.5\text{ V}$, $C_L = 100\text{ pF}$ See Note 4	25°C	1.5	2.5		V/ μs
		Full range	1			V/ μs
V_n Equivalent input noise voltage	$f = 10\text{ Hz}$	25°C		70		nV/ $\sqrt{\text{Hz}}$
	$f = 1\text{ kHz}$	25°C		12		
$V_{N(PP)}$ Peak-to-peak equivalent input noise voltage	$f = 0.1\text{ to }1\text{ Hz}$	25°C		1		μV
	$f = 0.1\text{ to }10\text{ Hz}$	25°C		1.5		
I_n Equivalent input noise current		25°C		0.6		fA/ $\sqrt{\text{Hz}}$
THD + N Total harmonic distortion plus noise	$V_O = 0.5\text{ V to }2.5\text{ V}$, $f = 10\text{ kHz}$, $R_L = 1\text{ k}\Omega$, $C_L = 100\text{ pF}$	$A_V = 1$	25°C	0.02%		
		$A_V = 10$	25°C	0.08%		
		$A_V = 100$	25°C	0.55%		
Gain-bandwidth product	$f = 10\text{ kHz}$, $C_L = 100\text{ pF}$	$R_L = 1\text{ k}\Omega$	25°C	4.7		MHz
BOM Maximum output swing bandwidth	$V_{O(PP)} = 2\text{ V}$, $R_L = 1\text{ k}\Omega$	$A_V = 1$, $C_L = 100\text{ pF}$	25°C	1		MHz
t_s Settling time	$A_V = -1$, Step = $0.5\text{ V to }2.5\text{ V}$, $R_L = 1\text{ k}\Omega$, $C_L = 100\text{ pF}$	to 0.1%	25°C	1.6		μs
		to 0.01%	25°C	2.2		
ϕ_m Phase margin at unity gain	$R_L = 1\text{ k}\Omega$	$C_L = 100\text{ pF}$	25°C	74		
Calibration time			25°C	300		ms

† Full range is $-40^\circ\text{C to }125^\circ\text{C}$ for Q suffix, $-55^\circ\text{C to }125^\circ\text{C}$ for M suffix.

NOTE 4: R_L and C_L values are referenced to 2.5 V.

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TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC4502 INPUT
 OFFSET VOLTAGE

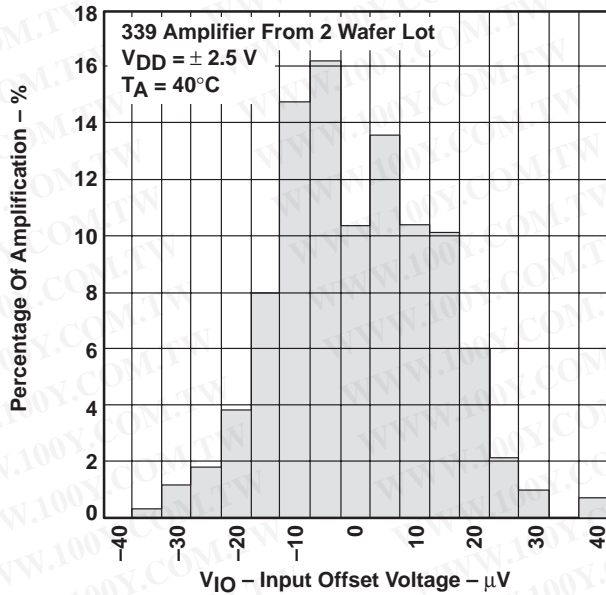


Figure 2

DISTRIBUTION OF TLC4502 INPUT
 OFFSET VOLTAGE

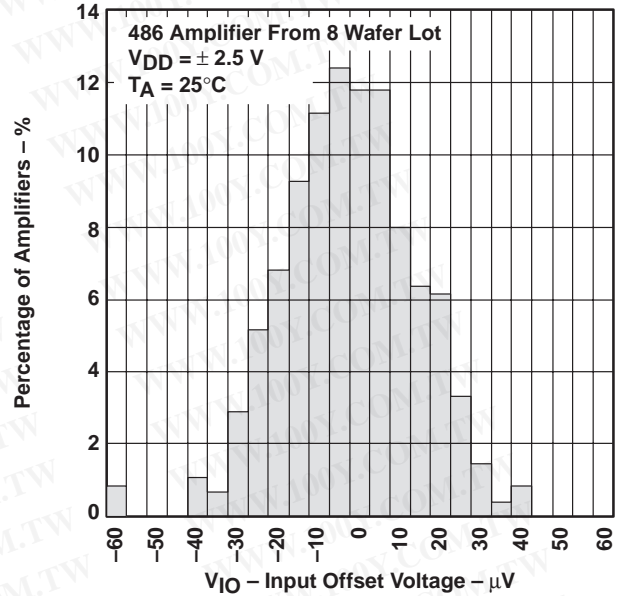


Figure 3

DISTRIBUTION OF TLC4502 INPUT
 OFFSET VOLTAGE

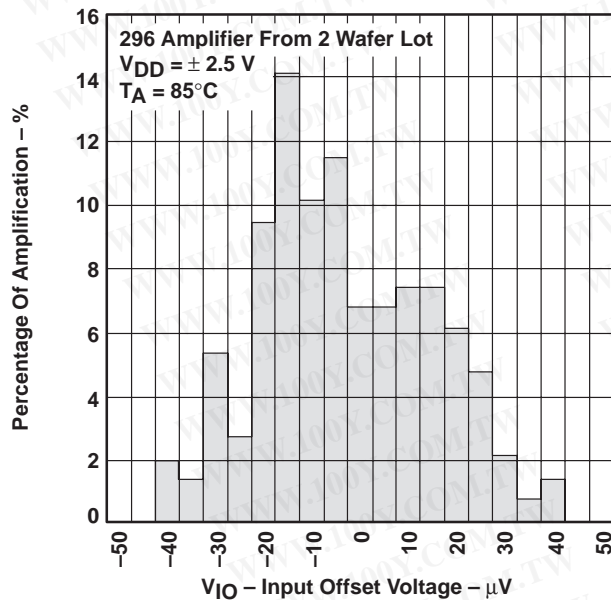


Figure 4

INPUT OFFSET VOLTAGE
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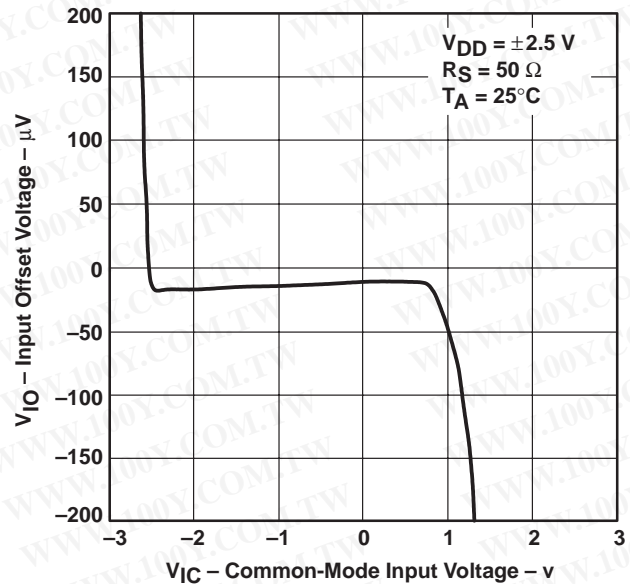


Figure 5

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TYPICAL CHARACTERISTICS

DISTRIBUTION OF TLC4502 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

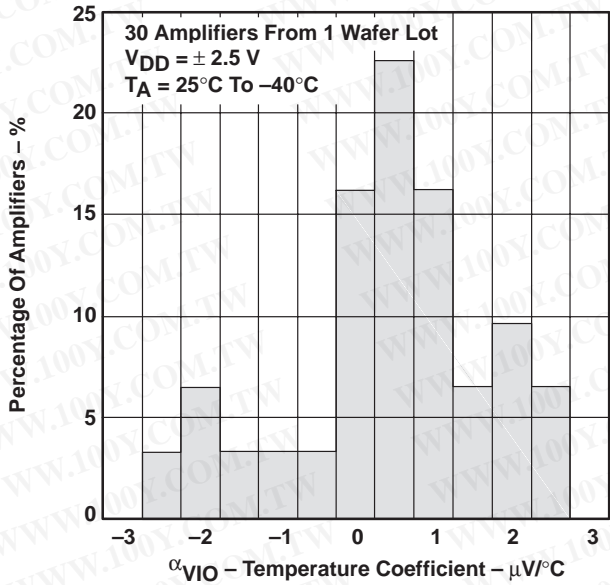


Figure 6

DISTRIBUTION OF TLC4502 INPUT OFFSET VOLTAGE TEMPERATURE COEFFICIENT

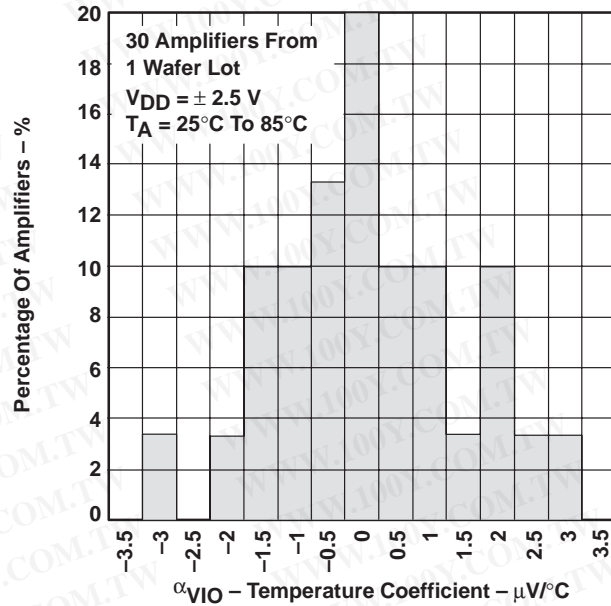


Figure 7

HIGH-LEVEL OUTPUT VOLTAGE vs HIGH-LEVEL OUTPUT CURRENT

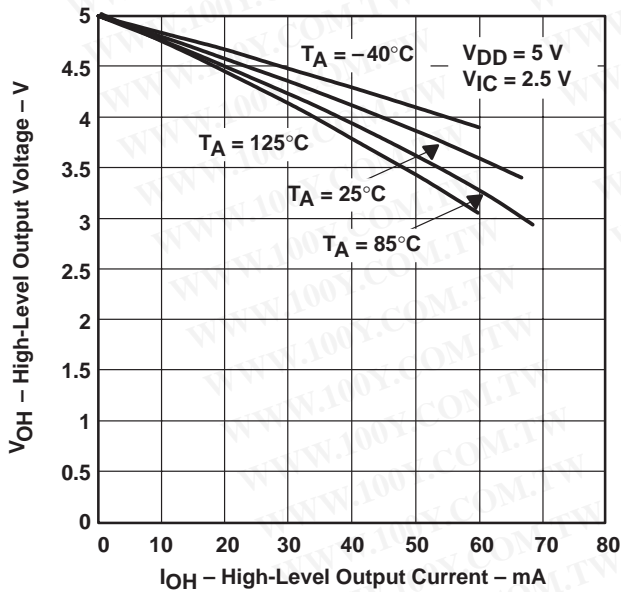


Figure 8

LOW-LEVEL OUTPUT VOLTAGE vs LOW-LEVEL OUTPUT CURRENT

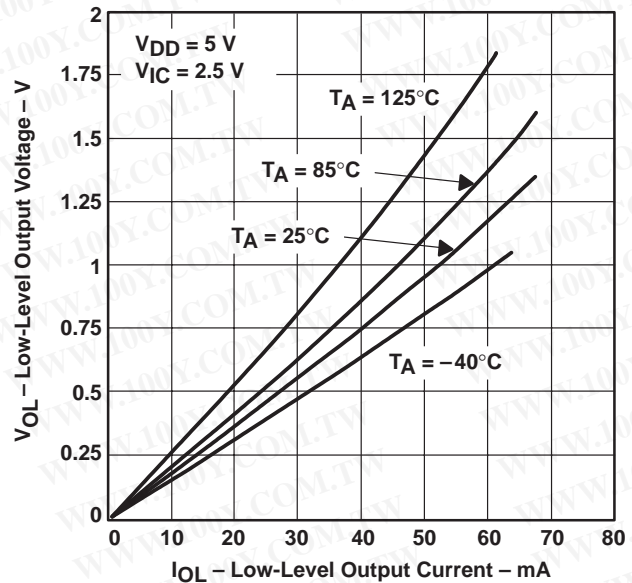


Figure 9

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MAXIMUM PEAK-TO-PEAK OUTPUT VOLTAGE
 vs
 FREQUENCY

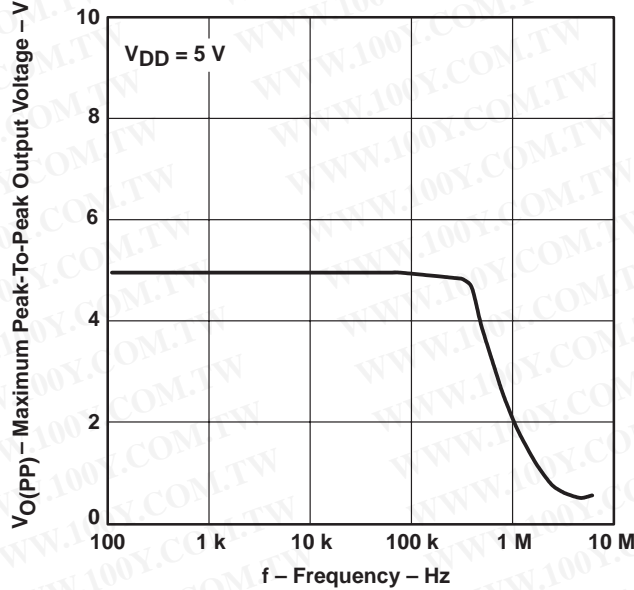


Figure 10

SHORT-CIRCUIT OUTPUT CURRENT
 vs
 FREE-AIR TEMPERATURE

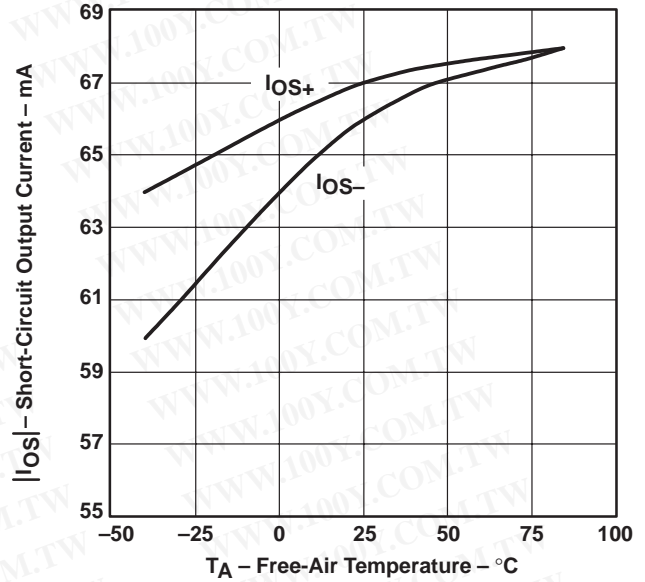


Figure 11

OUTPUT VOLTAGE
 vs
 DIFFERENTIAL INPUT VOLTAGE

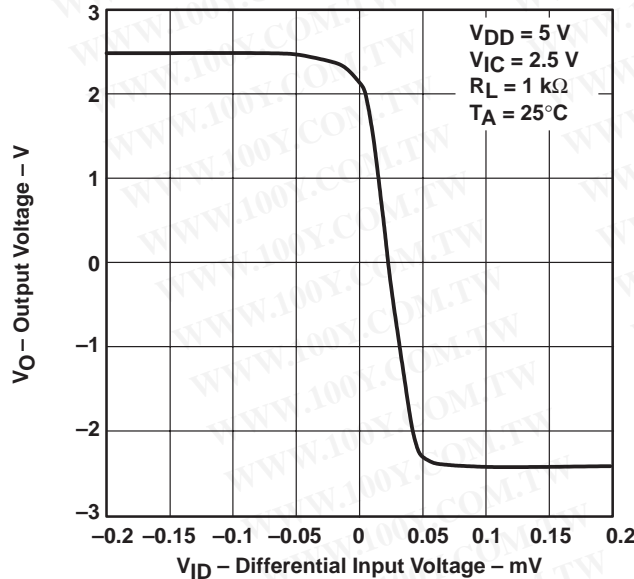


Figure 12

LARGE-SIGNAL DIFFERENTIAL
 VOLTAGE AMPLIFICATION
 vs
 FREE-AIR TEMPERATURE

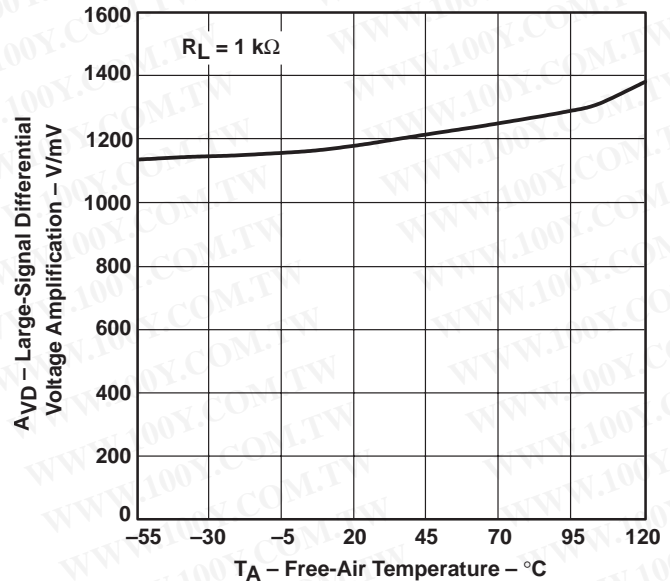


Figure 13

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TYPICAL CHARACTERISTICS

LARGE-SIGNAL DIFFERENTIAL VOLTAGE
 AMPLIFICATION AND PHASE MARGIN
 vs
 FREQUENCY

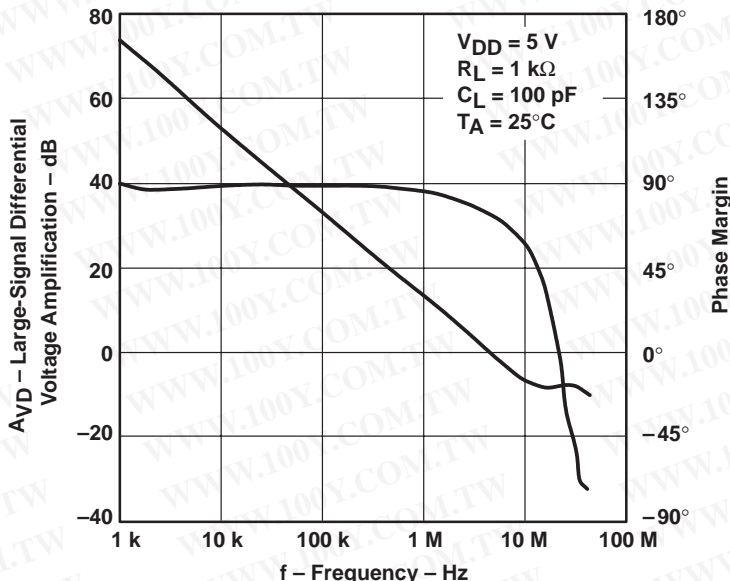


Figure 14

OUTPUT IMPEDANCE
 vs
 FREQUENCY

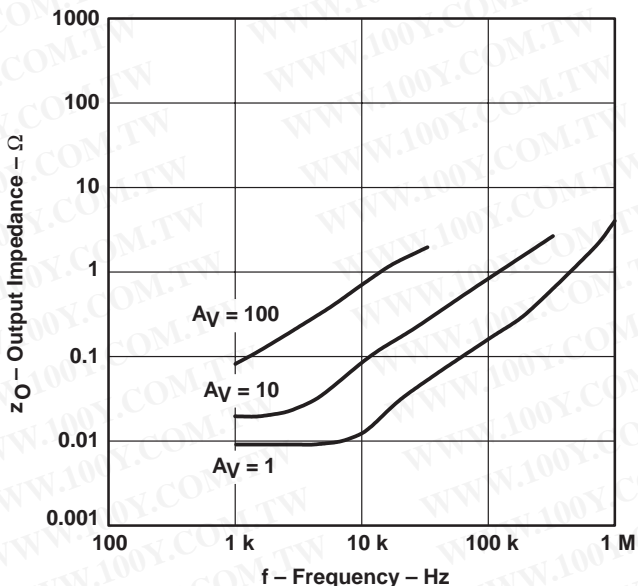


Figure 15

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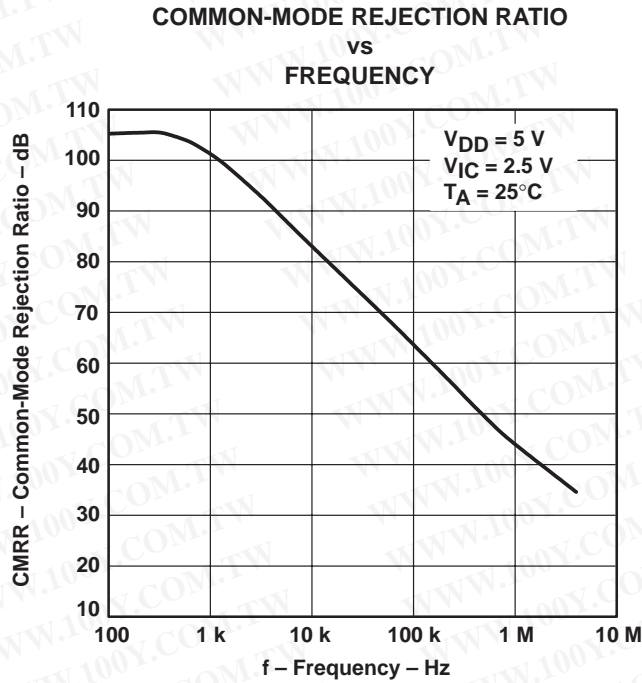


Figure 16

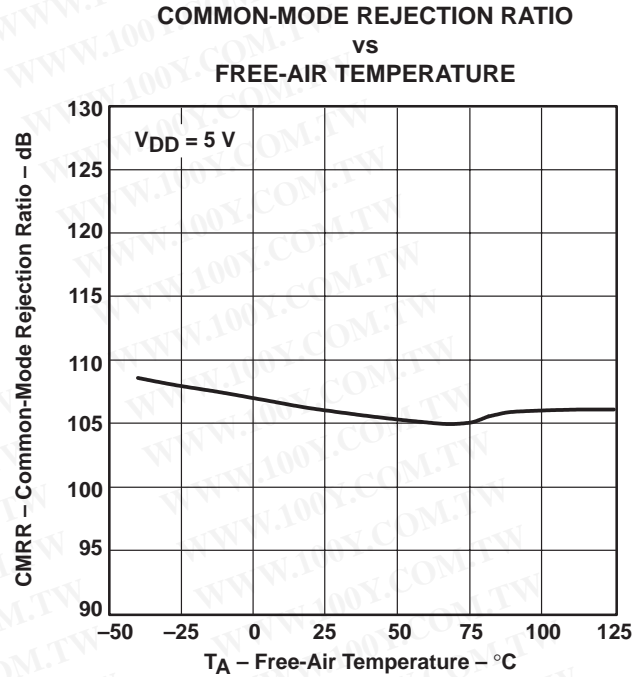


Figure 17

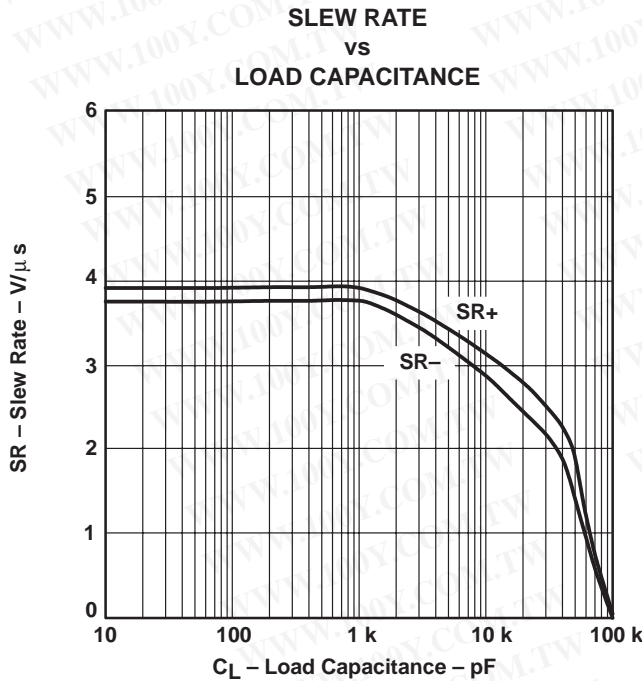


Figure 18

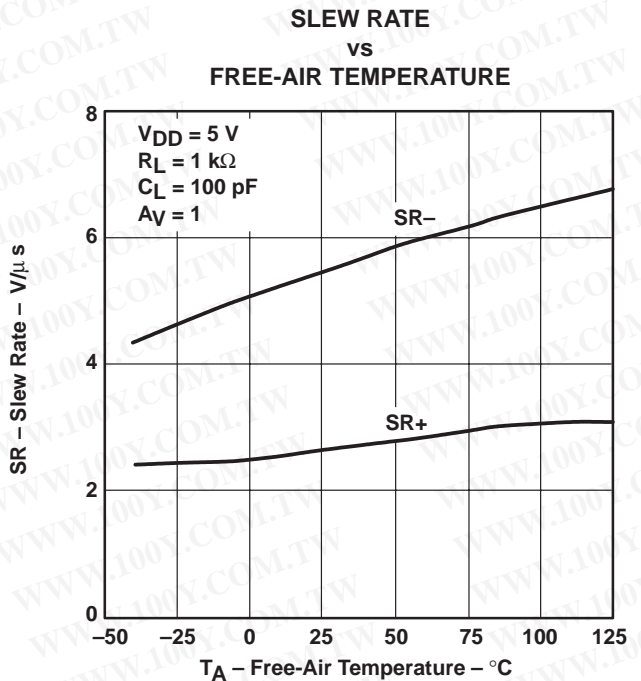


Figure 19

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TYPICAL CHARACTERISTICS

INVERTING LARGE-SIGNAL PULSE RESPONSE

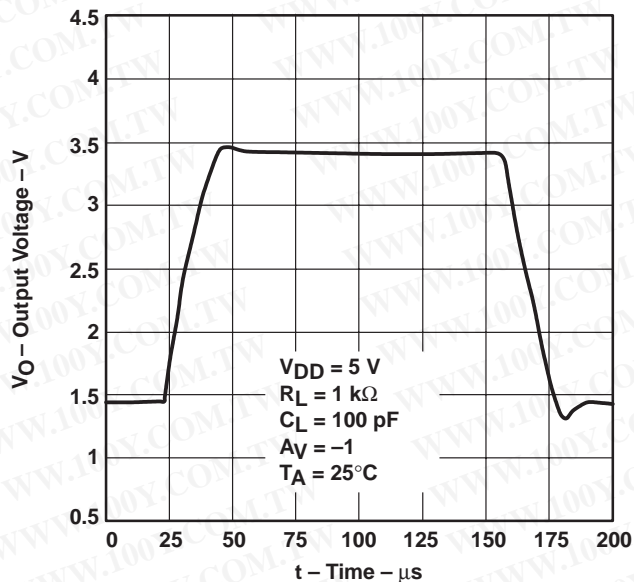


Figure 20

VOLTAGE-FOLLOWER LARGE-SIGNAL PULSE RESPONSE

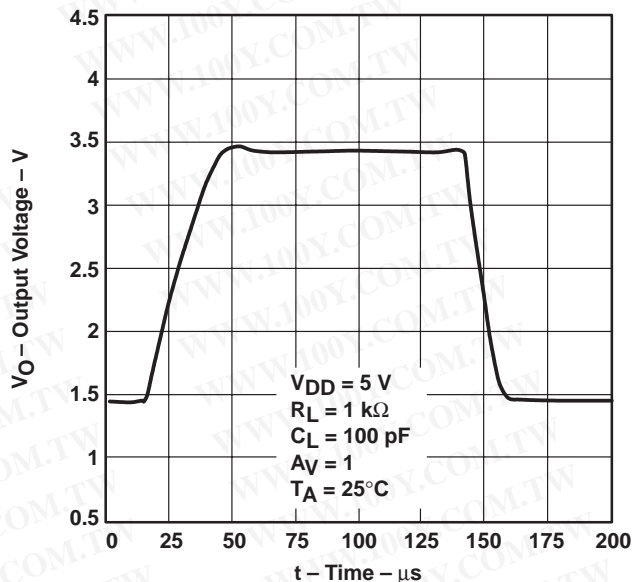


Figure 21

INVERTING SMALL-SIGNAL PULSE RESPONSE

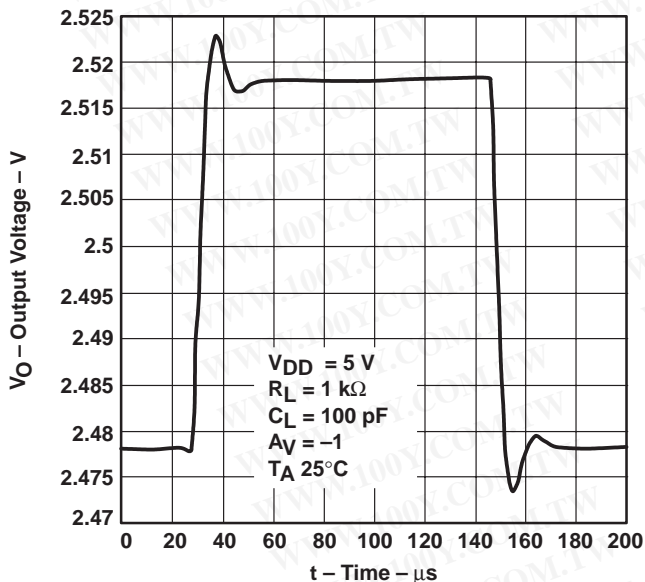


Figure 22

VOLTAGE-FOLLOWER SMALL-SIGNAL PULSE RESPONSE

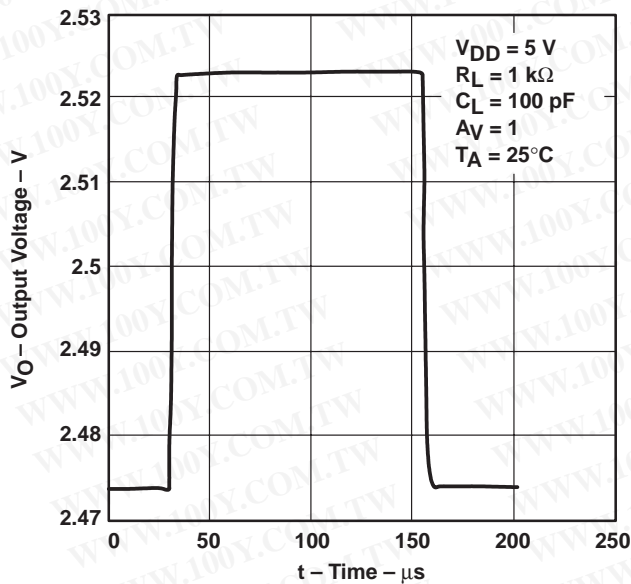


Figure 23

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TYPICAL CHARACTERISTICS

EQUIVALENT INPUT NOISE VOLTAGE
 vs
 FREQUENCY

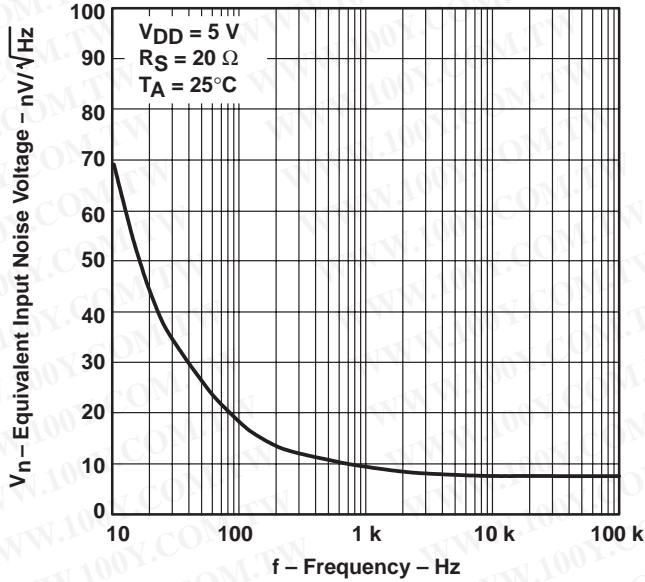


Figure 24

INPUT NOISE VOLTAGE OVER
 A 10-SECOND PERIOD

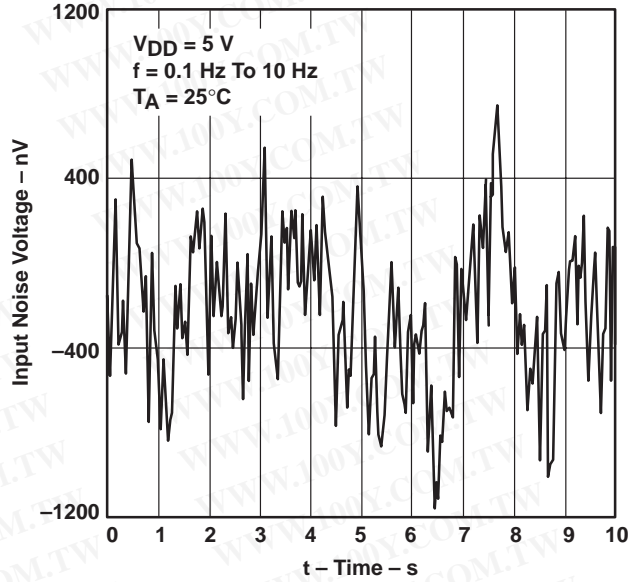


Figure 25

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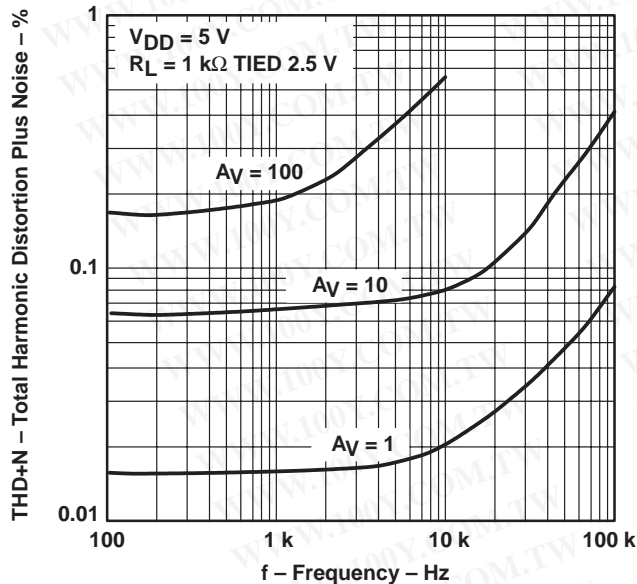


Figure 26

GAIN-BANDWIDTH PRODUCT
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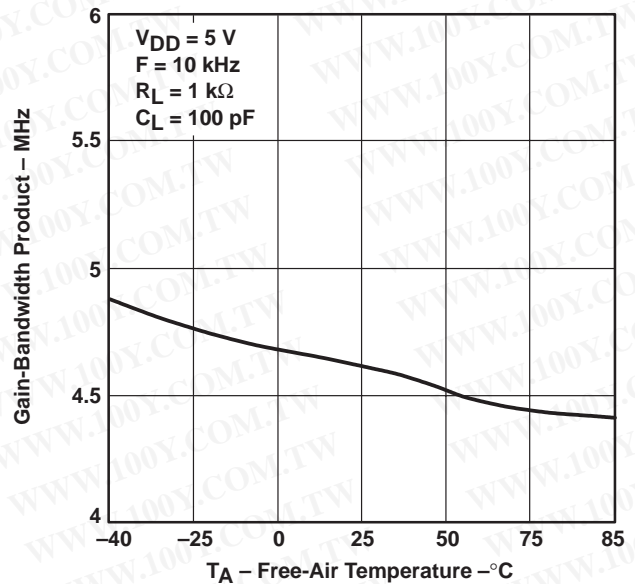


Figure 27

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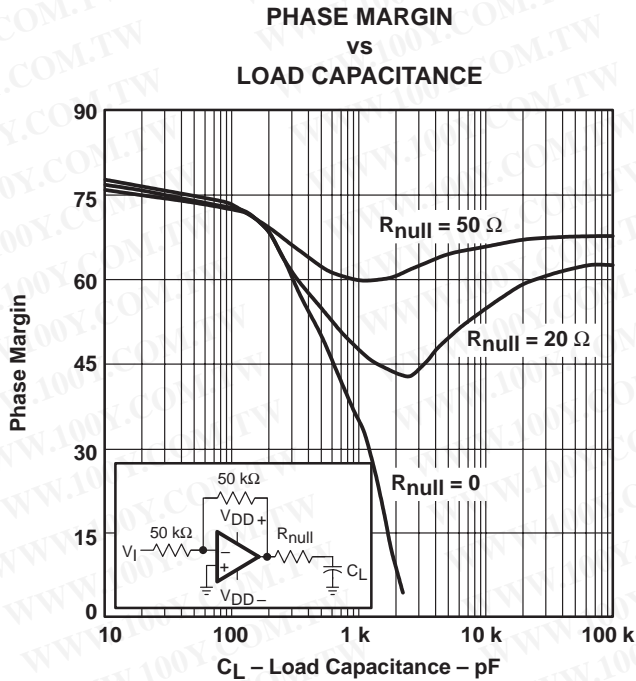


Figure 28

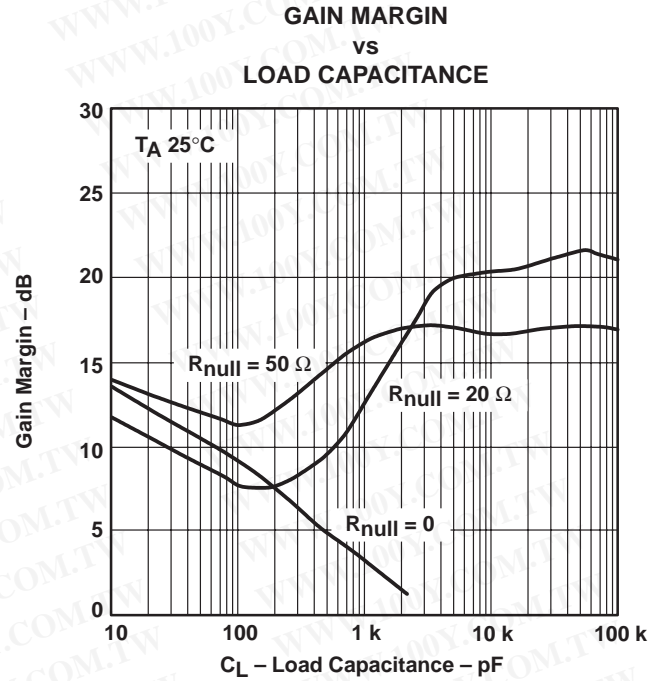


Figure 29

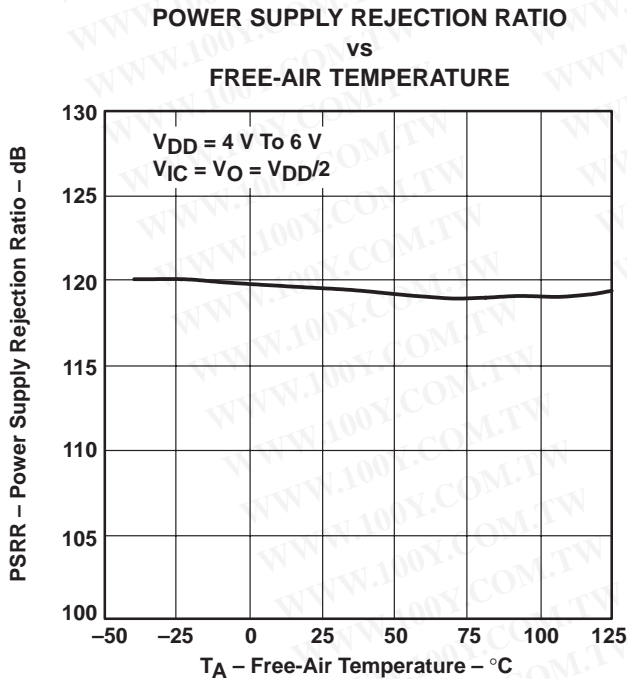


Figure 30

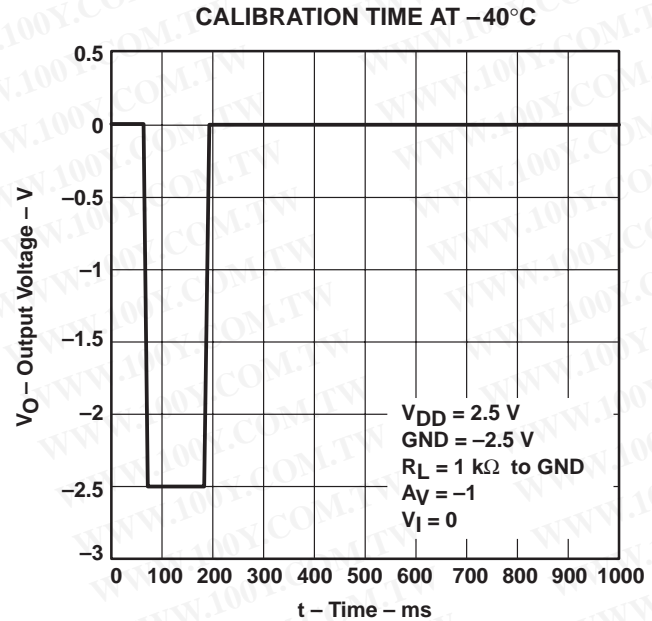


Figure 31

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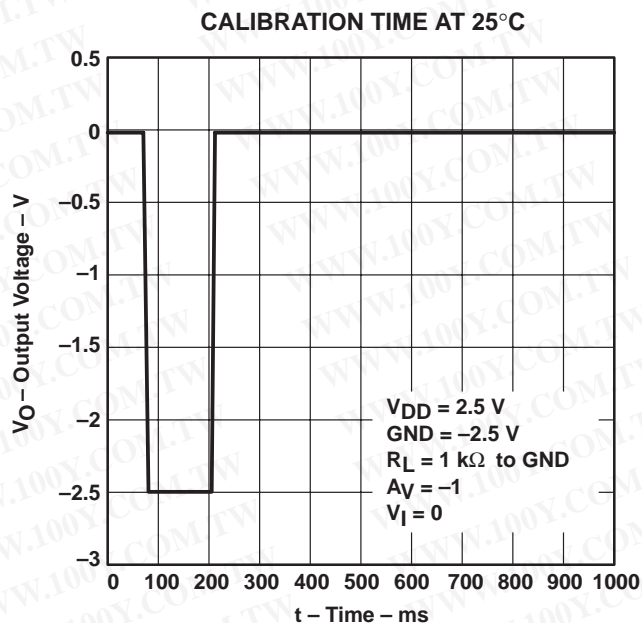


Figure 32

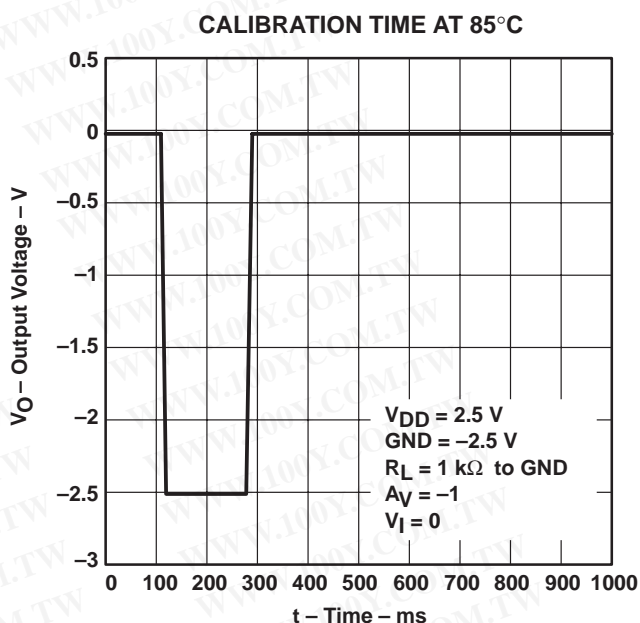


Figure 33

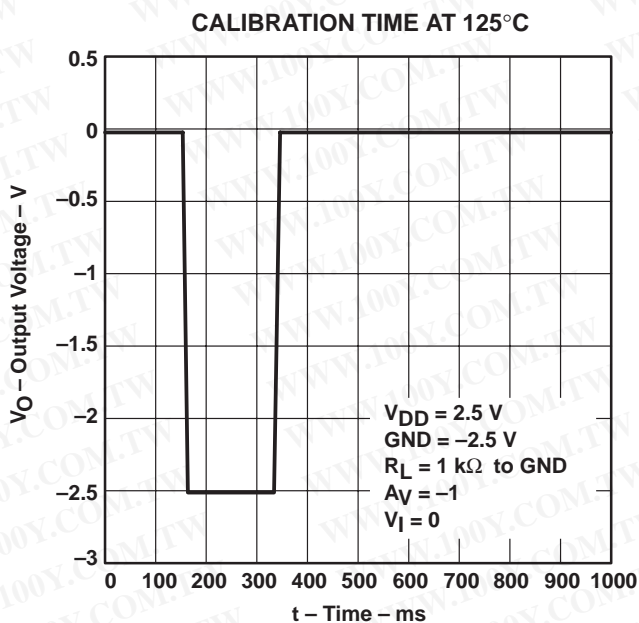


Figure 34

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APPLICATION INFORMATION

- The TLC4502 is designed to operate with only a single 5-V power supply, have true differential inputs, and remain in the linear mode with an input common-mode voltage of 0.
- The TLC4502 has a standard dual-amplifier pinout, allowing for easy design upgrades.
- Large differential input voltages can be easily accommodated and, as input differential-voltage protection diodes are not needed, no large input currents result from large differential input voltage. Protection should be provided to prevent the input voltages from going negative more than -0.3 V at 25°C . An input clamp diode with a resistor to the device input terminal can be used for this purpose.
- For ac applications, where the load is capacitively coupled to the output of the amplifier, a resistor can be used from the output of the amplifier to ground. This increases the class-A bias current and prevents crossover distortion. Where the load is directly coupled, for example in dc applications, there is no crossover distortion.
- Capacitive loads, which are applied directly to the output of the amplifier, reduce the loop stability margin. Values of 500 pF can be accommodated using the worst-case noninverting unity-gain connection. Resistive isolation should be considered when larger load capacitance must be driven by the amplifier.

The following typical application circuits emphasize operation on only a single power supply. When complementary power supplies are available, the TLC4502 can be used in all of the standard operational amplifier circuits. In general, introducing a pseudo-ground (a bias voltage of $V_I/2$ like that generated by the TLE2426) allows operation above and below this value in a single-supply system. Many application circuits shown take advantage of the wide common-mode input-voltage range of the TLC4502, which includes ground. In most cases, input biasing is not required and input voltages that range to ground can easily be accommodated.

description of calibration procedure

To achieve high dc gain, large bandwidth, high CMRR and PSRR, as well as good output drive capability, the TLC4502 is built around a 3-stage topology: two gain stages, one rail-to-rail, and a class-AB output stage. A nested Miller topology is used for frequency compensation.

During the calibration procedure, the operational amplifier is removed from the signal path and both inputs are tied to GND. Figure 35 shows a block diagram of the amplifier during calibration mode.

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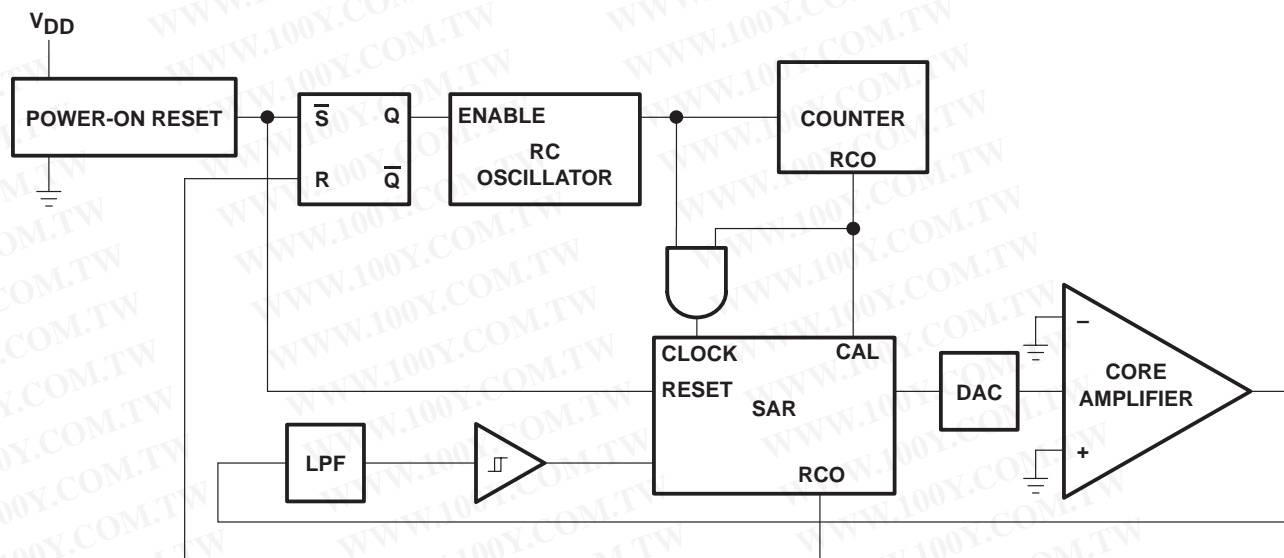


Figure 35. Block Diagram During Calibration Mode

The class AB output stage features rail-to-rail voltage swing and incorporates additional switches to put the output node into a high-impedance mode during the calibration cycle. Small-replica output transistors (matched to the main output transistors) provide the amplifier output signal for the calibration circuit. The TLC4502 also features built-in output short-circuit protection. The output current flowing through the main output transistors is continuously being sensed. If the current through either of these transistors exceeds the preset limit (60 mA – 70 mA) for more than about 1 μ s, the output transistors are shut down to approximately their quiescent operating point for approximately 5 ms. The device is then returned to normal operation. If the short circuit is still in place, it is detected in less than 1 μ s and the device is shut down for another 5 ms.

The offset cancellation uses a current-mode digital-to-analog converter (DAC), whose full-scale current allows for an adjustment of approximately ± 5 mV to the input offset voltage. The digital code producing the cancellation current is stored in the successive-approximation register (SAR).

During power up, when the offset cancellation procedure is initiated, an on-chip RC oscillator is activated to provide the timing of the successive-approximation algorithm. To prevent wide-band noise from interfering with the calibration procedure, an analog low-pass filter followed by a Schmitt trigger is used in the decision chain to implement an averaging process. Once the calibration procedure is complete, the RC oscillator is deactivated to reduce supply current and the associated noise.

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APPLICATION INFORMATION

The key operational-amplifier parameters CMRR, PSRR, and offset drift were optimized to achieve superior offset performance. The TLC4502 calibration DAC is implemented by a binary-weighted current array using a pseudo-R-2R MOSFET ladder architecture, which minimizes the silicon area required for the calibration circuitry, and thereby reduces the cost of the TLC4502.

Due to the performance (precision, PSRR, CMRR, gain, output drive, and ac performance) of the TLC4502, it is ideal for applications like:

- Data acquisition systems
- Medical equipment
- Portable digital scales
- Strain gauges
- Automotive sensors
- Digital audio circuits
- Industrial control applications

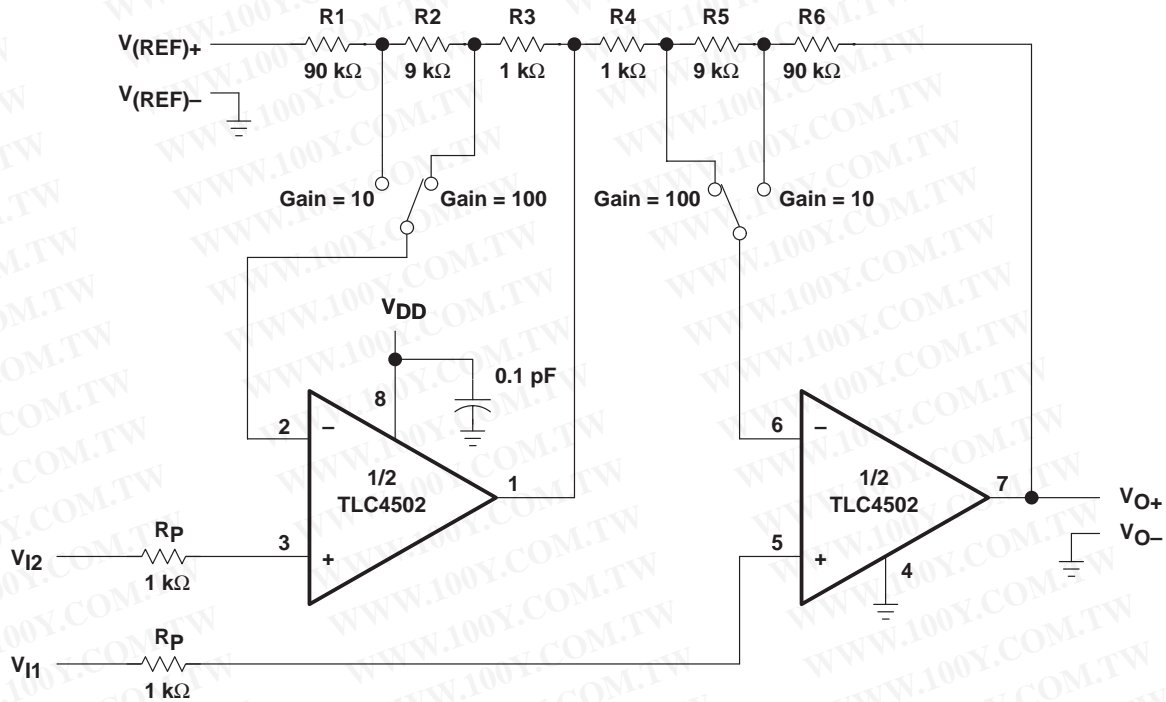
It is also ideal in circuits like:

- A precision buffer for current-to-voltage converters, a/d buffers, or bridge applications
- High-impedance buffers or preamplifiers
- Long term integration
- Sample-and-hold circuits
- Peak detectors

The TLC4502 self-calibrating operational amplifier is manufactured using Texas instruments LinEPIC process technology and is available in an 8-pin SOIC (D) Package. The C-suffix devices are characterized for operation from 0°C to 70°C. The I-suffix devices are characterized for operation from –40°C to 125°C. The M-suffix devices are characterized for operation from –55°C to 125°C.

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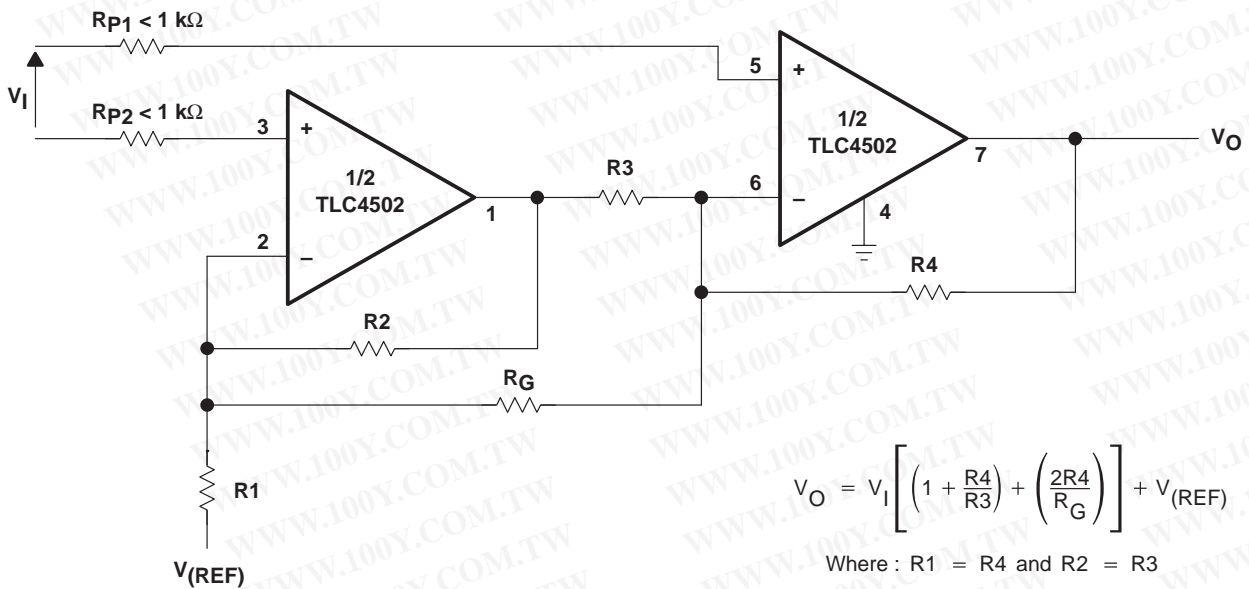
APPLICATION INFORMATION



$$\text{(Gain = 10)} \quad V_O = (V_{I1} - V_{I2}) \left(1 + \frac{R_6}{R_4 + R_5} \right) + V_{(REF)} \quad \text{Where } R_1 = R_6, R_2 = R_5, \text{ and } R_3 = R_4$$

$$\text{(Gain = 100)} \quad V_O = (V_{I1} - V_{I2}) \left(1 + \frac{R_5 + R_6}{R_4} \right) + V_{(REF)} \quad \text{Where } R_1 = R_6, R_2 = R_5, \text{ and } R_3 = R_4$$

Figure 36. Single-Supply Programmable Instrumentation Amplifier Circuit



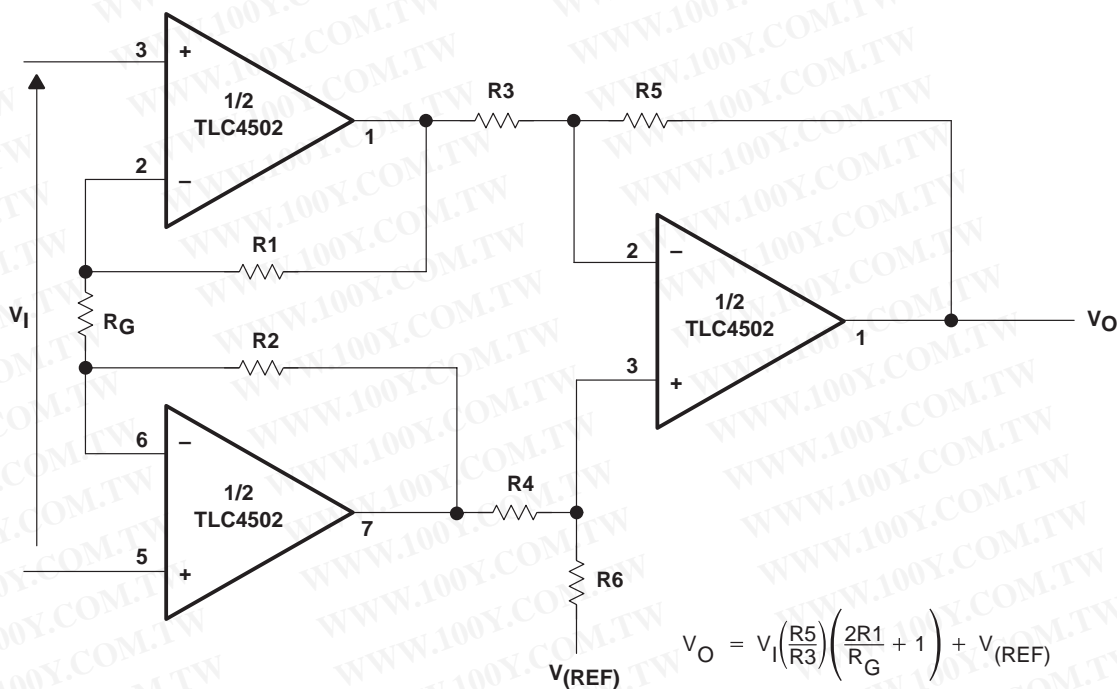
$$V_O = V_I \left[\left(1 + \frac{R_4}{R_3} \right) + \left(\frac{2R_4}{R_G} \right) \right] + V_{(REF)}$$

Where : R1 = R4 and R2 = R3

Figure 37. Two-Operational-Amplifier Instrumentation Amplifier Circuit

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Where : R1 = R2, R3 = R4, and R5 = R6

Figure 38. Three Operational-Amplifier Instrumentation Amplifier Circuit

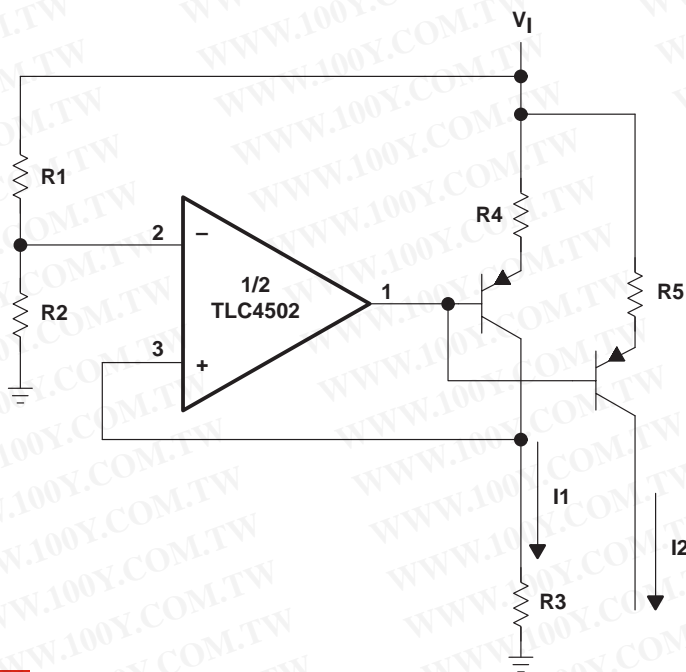


Figure 39. Fixed Current-Source Circuit

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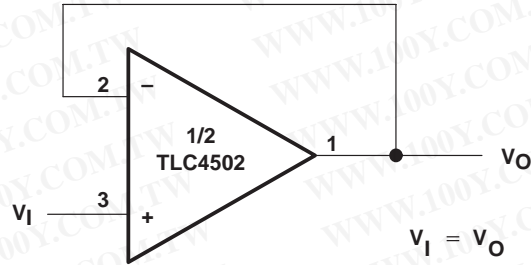


Figure 40. Voltage-Follower Circuit

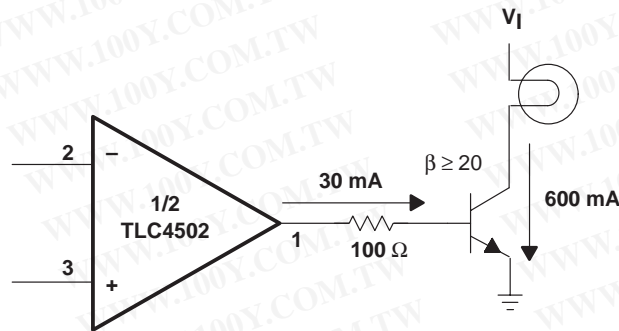


Figure 41. Lamp-Driver Circuit

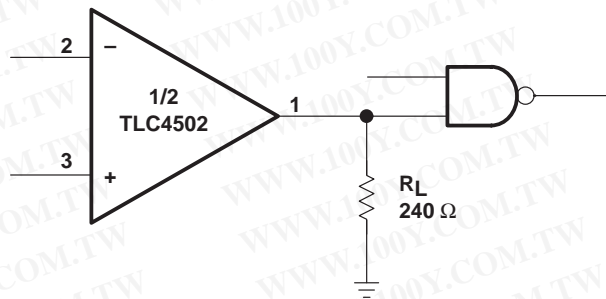


Figure 42. TTL-Driver Circuit

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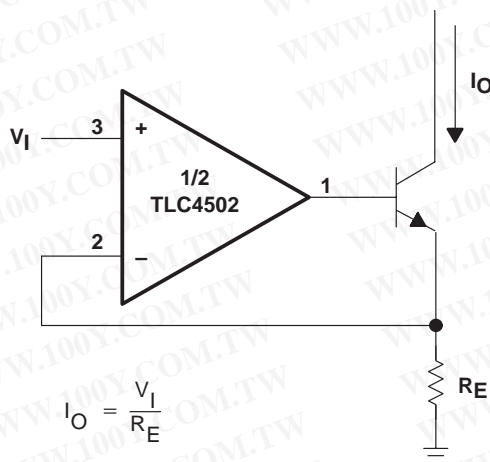


Figure 43. High-Compliance Current-Sink Circuit

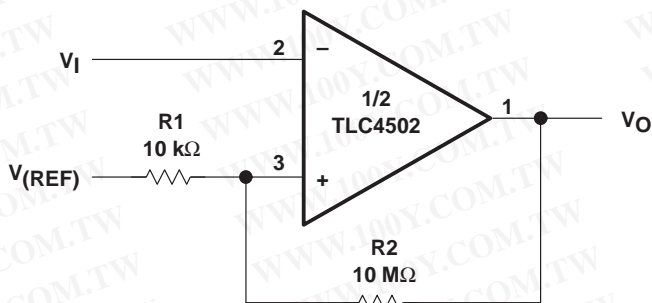


Figure 44. Comparator With Hysteresis Circuit

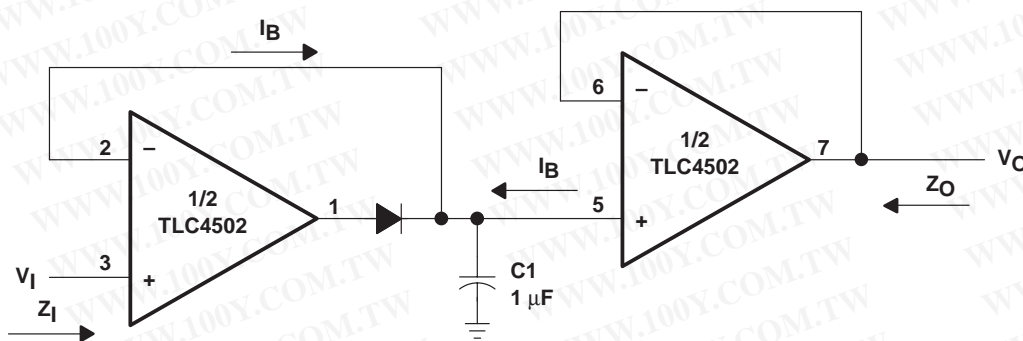


Figure 45. Low-Drift Detector Circuit

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APPLICATION INFORMATION

macromodel information

Macromodel information provided was derived using Microsim *Parts*™ Release 8, the model generation software used with Microsim *PSpice*™. The Boyle macromodel (see Note 4) and subcircuit in Figure 46 are generated using the TLC4501 typical electrical and operating characteristics at $T_A = 25^\circ\text{C}$. Using this information, output simulations of the following key parameters can be generated to a tolerance of 20% (in most cases):

- Maximum positive output voltage swing
- Maximum negative output voltage swing
- Slew rate
- Quiescent power dissipation
- Input bias current
- Open-loop voltage amplification
- Unity-gain frequency
- Common-mode rejection ratio
- Phase margin
- DC output resistance
- AC output resistance
- Short-circuit output current limit

NOTE 4: G. R. Boyle, B. M. Cohn, D. O. Pederson, and J. E. Solomon, "Macromodeling of Intergrated Circuit Operational Amplifiers", *IEEE Journal of Solid-State Circuits*, SC-9, 353 (1974).

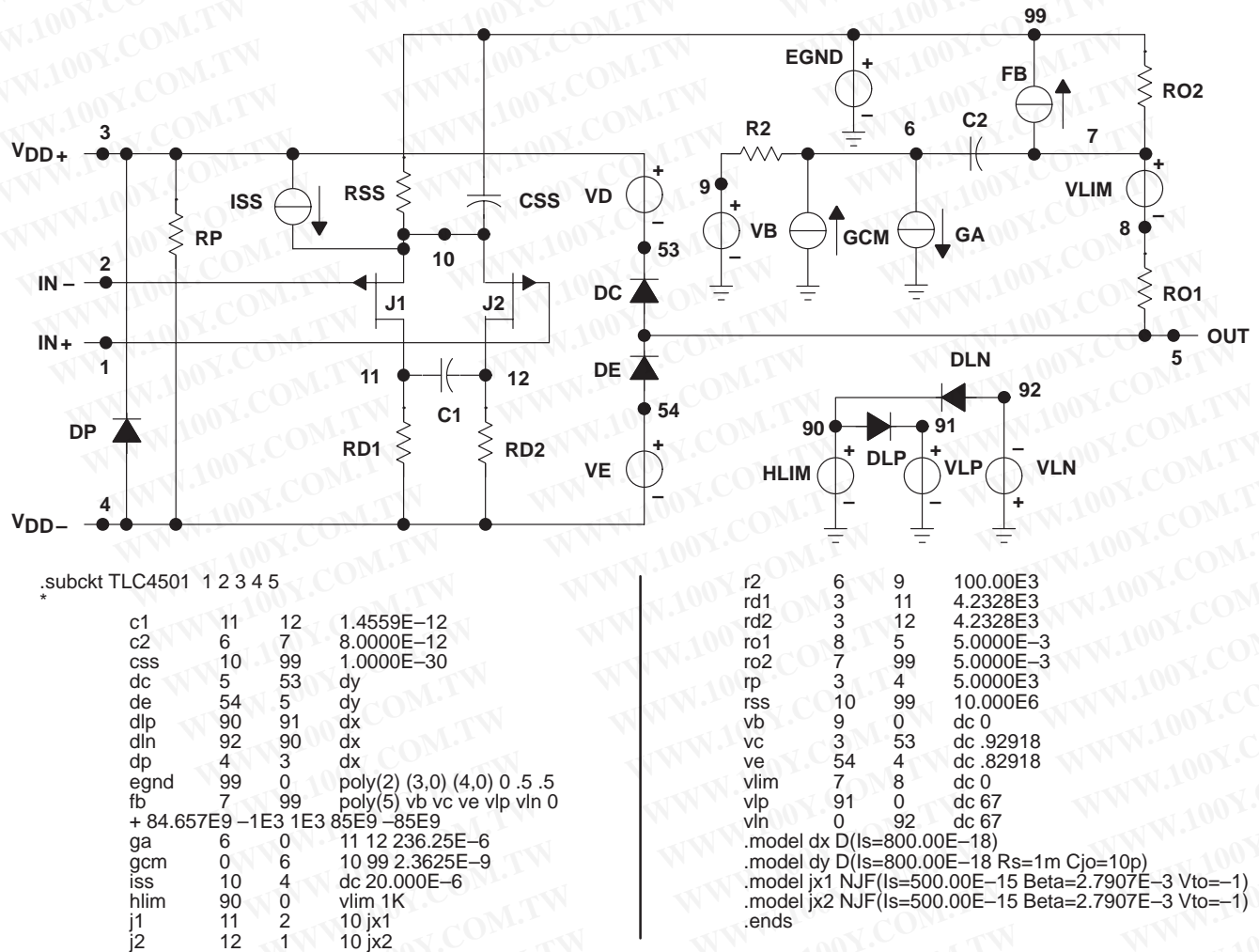


Figure 46. Boyle Macromodel and Subcircuit

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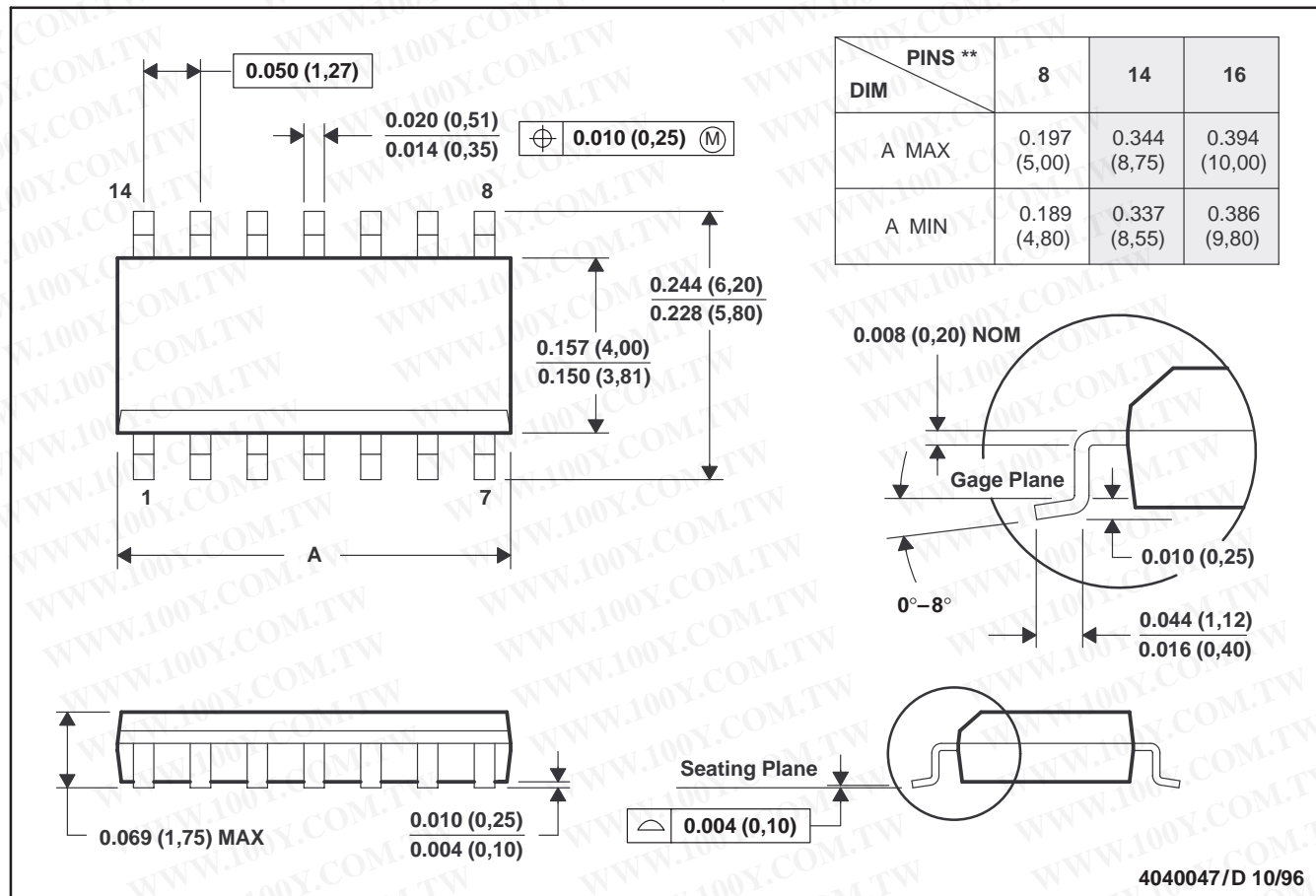
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MECHANICAL INFORMATION

D (R-PDSO-G**)

PLASTIC SMALL-OUTLINE PACKAGE

14 PIN SHOWN



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. Body dimensions do not include mold flash or protrusion, not to exceed 0.006 (0,15).
 D. Falls within JEDEC MS-012

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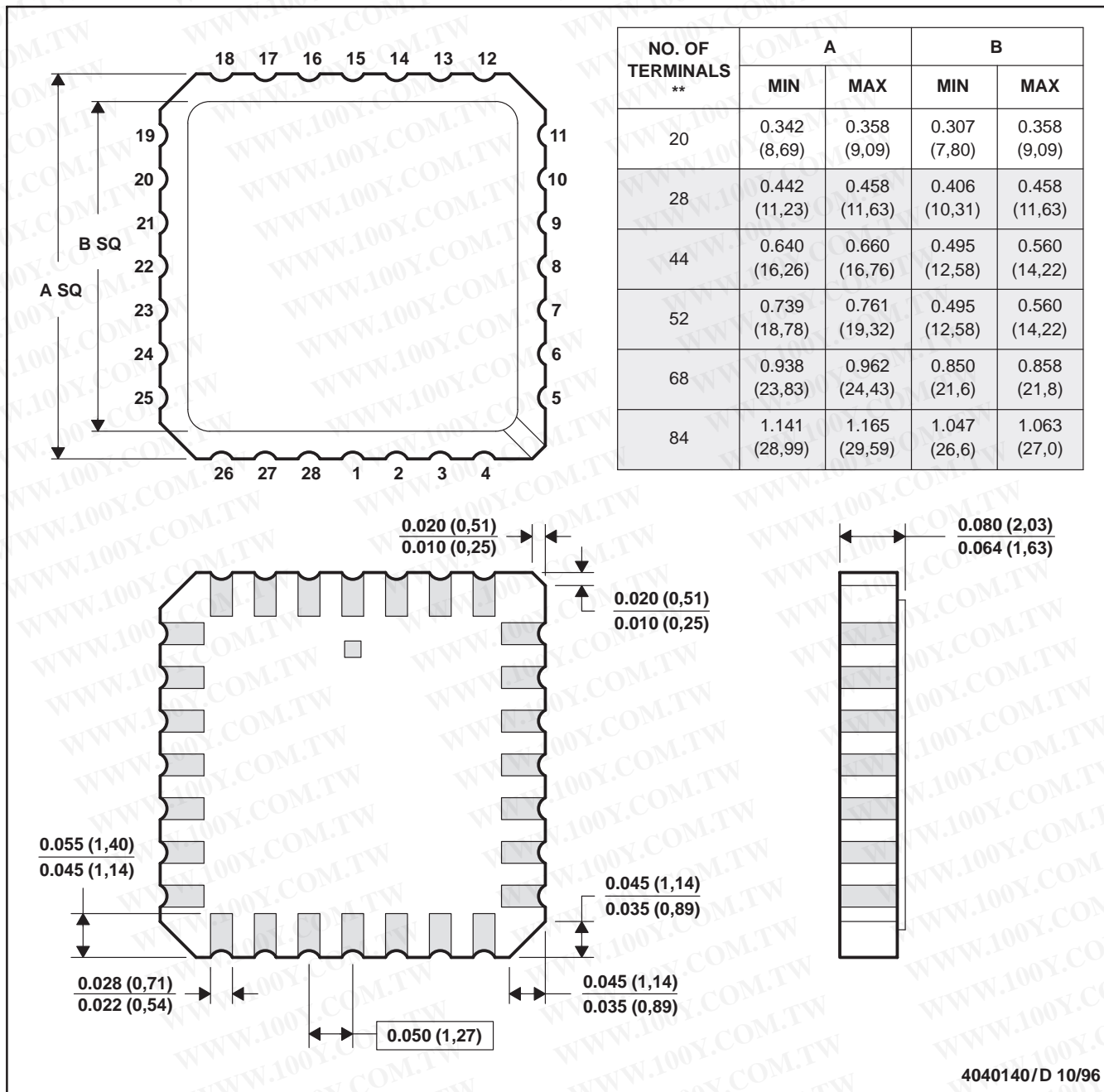
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MECHANICAL INFORMATION

FK (S-CQCC-N**)

LEADLESS CERAMIC CHIP CARRIER

28 TERMINAL SHOWN



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a metal lid.
 D. The terminals are gold plated.
 E. Falls within JEDEC MS-004

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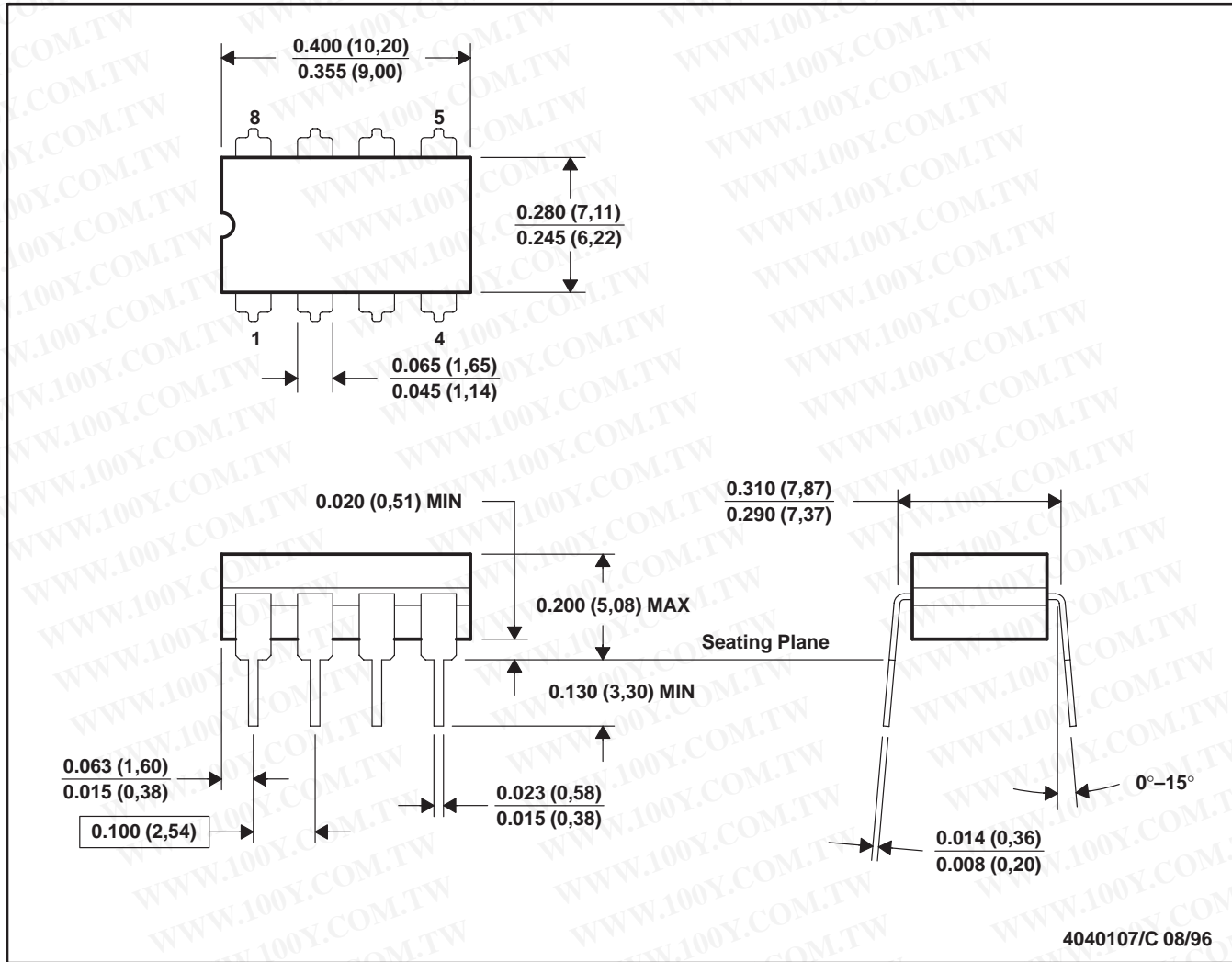
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MECHANICAL INFORMATION

JG (R-GDIP-T8)

CERAMIC DUAL-IN-LINE PACKAGE



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a ceramic lid using glass frit.
 D. Index point is provided on cap for terminal identification only on press ceramic glass frit seal only.
 E. Falls within MIL-STD-1835 GDIP1-T8

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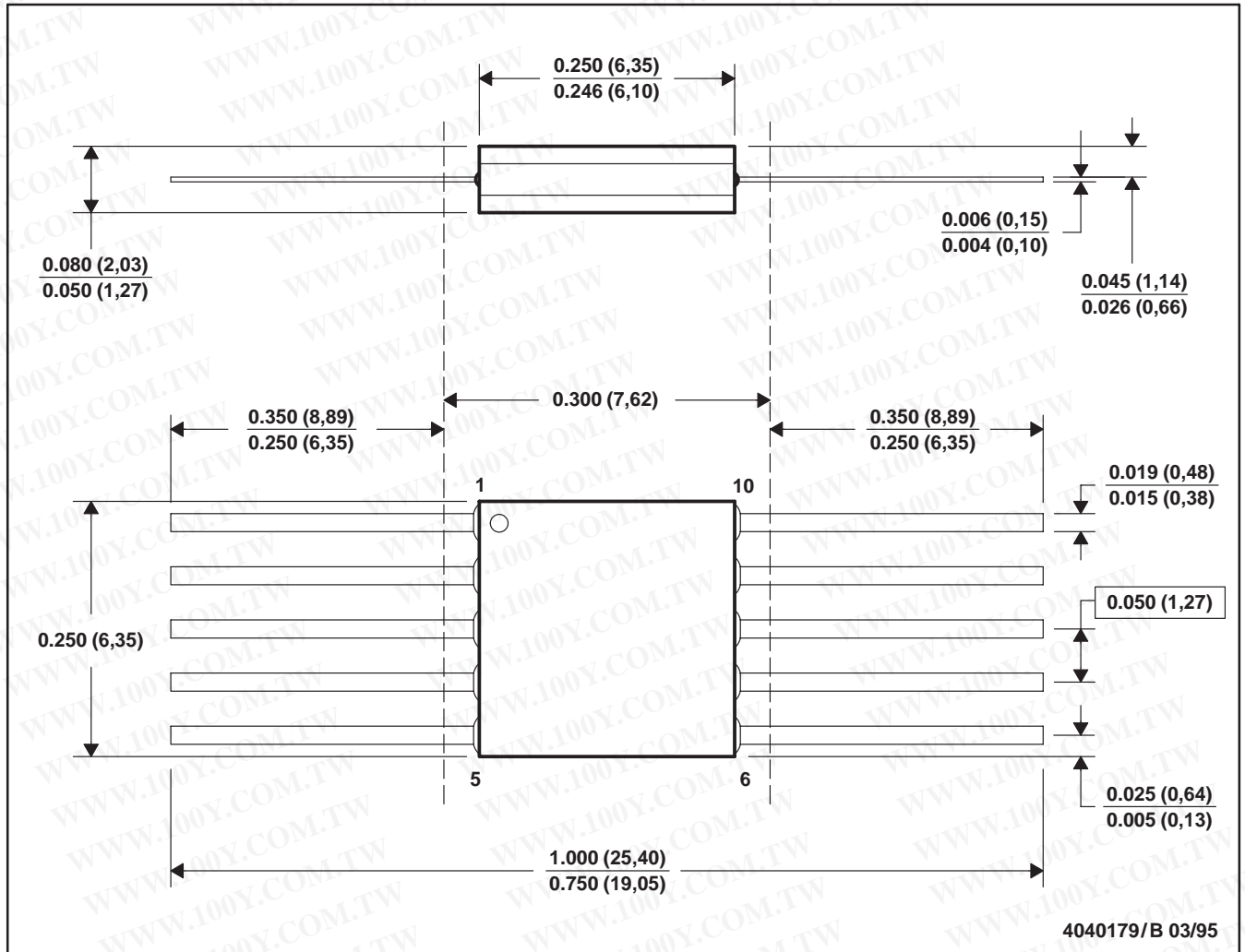
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MECHANICAL INFORMATION

U (S-GDFP-F10)

CERAMIC DUAL FLATPACK



- NOTES: A. All linear dimensions are in inches (millimeters).
 B. This drawing is subject to change without notice.
 C. This package can be hermetically sealed with a ceramic lid using glass frit.
 D. Index point is provided on cap for terminal identification only.
 E. Falls within MIL STD 1835 GDFP1-F10 and JEDEC MO-092AA

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